A Voltage-Lift DC-DC Converter with Large Conversion Ratio

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Abstract

A extension of the high boost voltage-lift DC-DC converter with large conversion ratio has been proposed in this paper. The proposed extension is combined the switched-inductor cell (SL-cell) and modular voltage cell (MV-cell). The proposed structure can achieve the large voltage conversion without high duty-cycle and the low voltage of the components. Moreover, the PID controller for novel SL-MV voltage-lift DC-DC converter also introduces. This technique a good-performance output voltage can kept constant with an good transient performance when the output load is suddenly changed. In order to prove the theoretical analysis, the experimental setup has been built for the DC load of $150[\Omega]$ and $300[\Omega]$. In addition, the transient of output voltage has been tested to determine the controller. Experimental results validate the effectiveness of the theoretical analysis proving the satisfactory converter performance.

Key words : tranformerless converter; high-voltage gain; switched-inductor; voltage-lift; duty ratio

I. Introduction

Nowadays, energy from renewable sources such as wind power, photovoltaic power, and fuel cells has become the best solution to generate clean electric energy. However, the demerit characteristics of these energies is the low-output voltage and instability. Thus, a DC-DC converters [1-2] with a high boost voltage conversion is used to boost the low output voltage of the clean energy to high voltage. Fig. 1 indicates a block diagram of the power-conversion system. A high gain DC-DC converter is used to boost a low voltage into a high voltage dc bus.



Fig. 1. Block diagram of power-conversion system.

Various DC-DC converter topologies have been researched and obtained a high-voltage gain in both isolated and non-isolated topologies. For the isolated DC-DC converter topologies [3-6], to boost low input voltage into high output voltage, a high frequency transformer is used. As a result, the cost, weight and volume of converter circuit are significantly risen. Furthermore, the leakage

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inductances can be caused a voltage spike on switches. Due to the transformer losses, leakage inductance and large volume/weight of the converter, the transformer-based topologies is difficult to obtain a high efficiency and low cost like nonisolated topologies. For conventional non-isolated DC-DC converters, an extremely high duty cycle is required to obtain a high voltage gain. Recently, many high boost DC-DC converters have been proposed in [7-20] to obtain a high voltage gain without using high duty cycle. In the high boost converters with non-isolated structure can obtain the high voltage gain; some techniques were proposed and used such as cascaded technique [10], switched-inductor technique [11], switchedcapacitor techniques [12], interleaved [13]-[14], and voltage-lift [15-16] techniques have been introduced. In switched-capacitor technique [8], by combining several switches and capacitors with minimum inductor achieving a higher output voltage can be achieved. The voltage-lift DC-DC converter and it's extension have been proposed in [21], which can increase the voltage gain. However, it is just verified with the simulation.

This paper presents the novel SL-MV voltagelift DC-DC converter with PID controller was determined with the voltage gain ratio will be improved and using the small duty cycle. Moreover, the PID controller with the experimental has been shown.



Fig. 2. Voltage-lift high boost DC-DC converter in [21].

II. Novel Voltage-lift DC-DC Converter

Fig. 2 shows the voltage-lift DC-DC converter in [21]. It consists of one SL cell L_a, L_b, D_a, D_b , and D_c , two power switches $S_1 - S_2$, modular voltage boost stage L_1, D_1, D_2 and C_1 , one diode D_0 , one capacitors C_0 . Fig. 3 shows the key waveforms of the voltage-lift converter operating in the CCM.



Fig. 3. Operating modes of the suggested converter: (a) state 1 and (b) state 2.

1. Derivation of novel voltage-lift DC-DC converter

Fig. 4 represents the operating state of the suggested converter. When S_1 is turned "ON" with DT. The D_a, D_c diodes are ON, while the D_b diode is OFF. The D_0 diode is forward-biased and D_1 is OFF. We have

$$\begin{cases} L_a \frac{di_{La}}{dt} = L_b \frac{di_{Lb}}{dt} = V_i \\ L_1 \frac{di_{L1}}{dt} = V_{C1} - V_o. \end{cases}$$

$$\tag{1}$$

When S_1 is switched "OFF" and S_2 is switched "ON"; the D_0 diode is OFF while the D_1 diode is ON. We have

$$\begin{cases} L_a \frac{di_{La}}{dt} + L_b \frac{di_{Lb}}{dt} = V_i - V_{C1} \\ L_1 \frac{di_{L1}}{dt} = V_{C1}. \end{cases}$$

$$(2)$$

The voltage gain of the converter is given

$$G = \frac{1+D}{D(1-D)},\tag{3}$$

2. Derivation of novel voltage-lift DC-DC converter

The novel voltage-lift DC-DC is shown in Fig. 4. It can be extended to produce a higher voltage conversion ratio by cascading more SL and MV cells, and the structure is shown in Fig. 5. It will be composed of SL-cell with one inductor and three diodes, and MV-cells with one inductor, one capacitor and two diodes.



Fig. 4. Novel voltage-lift DC-DC converter with with n SL-cells.



Fig. 5. Operating modes of the novel voltage-lift DC-DC converter with n SL-cells: (a) S_1 is ON, S_2 is OFF and (b) S_1 is OFF, S_2 is ON.

In the operating analysis, S_1 is ON and S_2 is

OFF, D_1 and D_2 are OFF. The diodes D_{bn} in SLells are OFF, diodes D_{an} and D_{cn} are ON. In this case, L_a and $L_{b1}, \dots L_{b(n+1)}$ are in series connection. When S_1 is OFF and S_2 is ON, D_1 and D_2 are ON. The diodes D_{bn} in SL-cells are ON, diodes D_{an} and D_{cn} are OFF. In this case, L_a and $L_{b1}, \dots L_{b(n+1)}$ are in parallel connection. The voltage gain of the novel voltage-lift DC-DC converter with n SL-cells is given

$$V_o = \frac{1+nD}{D(1-D)}V_i \tag{4}$$



Fig. 6. Voltage gain comparison when n = 1, 2, 3, 4.

3. The novel voltage-lift DC-DC converter with SL-MV cells

The voltage gain of the suggested converter can be improved by adding m-SL cells and m-stages. The switched-inductor cell can connect in series with nth times the voltage stress of diodes in SL-cell are decreased. Furthermore, the stage m_{th} can connect in series where the voltage gain can increase with n times compared to the voltage gain of the basic converter.



Fig. 7. Novel voltage-lift DC-DC converter with with SL-MV cells.



Fig. 8. Voltage gain comparison when m = 1, 2, 3, 4.

III. Control algorithm for novel voltage-lift DC-DC converter with SL-MV cells

Figure 9 demonstrates the block diagram controlled boost type DC-DC converter and the scheme of digital control circuit. In here, the power circuit of the DC-DC converter are included switches, diodes, inductors and capacitors. The output current and voltage is read by voltage and current sensor circuits. These analog signals are thrown the analog to digital block of the digital signal processor to convert the analog signal to the digital signal. Moreover, the pulse width modulation signal will be generated by the PWM generator of the digital signal processor with the switching frequency f_s . The switching frequency of the switching devices in power circuit will operate in the frequency of f_s . The control calculation is updated by each sampling point.



Fig. 9. Block diagram. (a) PID controlled DC-DC converter; and (b) Output voltage and current control algorithm.

IV. Experimental Verification

A laboratory prototype was set up to validate the experimental verification. Figure 9 presents a prototypal photograph and parameters for experiment were indicated in the Table 1. In the switchednductor cell, two inductors are 1 mH and three diodes STTH3002C were used. Two MOSFETs IXTR48P20P was used for switch S_1 and S_2 . One inductor of 2 mH, and one capacitor of 220 μ F used for L_1 and C_1 . Three diodes STTH3002C were used for D_0 , D_1 and D_2 . The PWM signals are generated by Digital Signal Processor of Texas Instrument TMS320F28027 for two MOSFET S_1 and S_2 .



Fig. 9. Prototypal photograph.

Table 1.	List of	parameters.
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Parameter		Values
Input DC voltage (Vi)		10 [V] - 20 [V]
Output voltage (Vo)		120 [V]
Resistor		150-300 [Ω]
Inductors	La= Lb	1 [mH]
	L1	2 [mH]
Capacitors	C1	220µF/200 [V]
	CO	330 µF/200 [V]
Operating frequency (fs)		10 [kHz]
MOSFETs (S1, S2)		IXTR48P20P
Diodes (Da-Db-Dc-D0-D1-D2)		STTH3002C



Fig. 10. Experiment waveforms when V_i =20[V], R=300[Ω].



Fig. 11. Experiment waveforms when V_i =20[V], R=150[Ω].

In the experimental verification, the input voltage is set to 20[V]. The operating frequency of semiconductor devices are 10 kHz. The output voltage of 120[V] was used for the resistive load is 150[Ω] and 300[Ω], as shown in Figures 9–10. In the Figures 9–10: (a) inductors La, Lb, L1 current, output current, (b) input voltage, capacitor C1 voltage, output voltage, diode D0 voltage; (c) diodes Da, Db, Dc and D1 voltage; and (d) diode D2, switches S1 and S2 voltages. The inductor L1 current is linearly increased at $D \cdot T$ when MOSFET S1 is ON. In addition, the inductor L1 current is gradually decreased during $(1-D) \cdot T$. The voltage of the experimental capacitor C1 is boosted to 58[V].



Fig. 12. Experimental waveform of the converter when load is suddenly changed. From top to bottom: input voltage, output voltage, load current.

Figure 12 Illustrates the dynamic response of the novel DC-DC converter with SL-MV cells when the load is suddenly changed. As shown in Figure 12, we can see that the load voltage is maintained constant at 120[V], even though the resistive load is changed from 150[Ω] to 300[Ω] and vice versa.

V. Conclusions

The novel high gain DC-DC boost converter based on the switched-inductor cell and modular voltage cell was presented in this paper. The novel converter can give the high voltage gain with combination of n switched-inductor cell and n-stage with a modular voltage cell. The operating mode of the converter adopting an SL cell and MV cell was analyzed, and the experimental results have been given to verify the analysis and merits of the converter. Moreover, the novel SL-MV voltage-lift DC-DC converter with PID controller was validated to determine the transient response for change in load conditions.

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