

# A Study on a Gain-Enhanced Antenna for Energy Harvesting using Adaptive Particle Swarm Optimization

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**Abstract** – In this paper, the adaptive particle swarm optimization (APSO) algorithm is employed to design a gain-enhanced antenna with a reflector for energy harvesting. We placed the reflector below the main radiating element. Its back-radiated field is reflected and added to the forward radiated field, which could increase the antenna gain. We adopt the adaptive particle swarm optimization (APSO) algorithm, which improves the speed of convergence with a high frequency solver. The result shows that performance of the optimized design successfully satisfied the design goal of the frequency band, gain and axial ratio.

**Keywords:** Energy harvesting, Reflector, Gain-enhanced antenna, Adaptive particle swarm optimization, Circular polarization

## 1. Introduction

Recently, the field of energy harvesting is drawing social and industrial attention, since energy is running out. It should be recycled and consumed efficiently. Especially, the infrastructure needs to be operated intelligently. Sensor nodes and their link play an important role in monitoring and maintaining the quality of energy such as electric power. The sensor node is an electronic device that consumes the electric power for tasking, and its battery should be substituted on a regular basis [1-3]. Changing the battery of a sensor node is difficult job, because it is placed high above the ground or a high wall. Therefore, Wireless charging technology using RF energy harvesting is recommended in this case.

Energy harvesting is to collect radio frequency (RF) energy propagating in the air. The ambient RF power is a good potential candidate for the energy supply as it is widely emitted from numerous reliable electromagnetic power resources [4, 5]. An energy harvesting equipment always needs one antenna or more as a core component which can catch the ambient RF power, when the optimal condition is met. For efficient energy harvesting, a good antenna with a circular polarization (CP), low profile and high gain is required.

[6, 7]. Regarding the necessity of the circular polarization, the RF energy flies in the air with an arbitrary polarization, and the circular polarization will increase the probability of catching the incoming energy [8].

Considering the easy installation of a sensor node on a high wall, a small area of structure is preferred, which will make a layout complicated and its optimization design time-consuming. To overcome the degraded gain caused by the installation on the wall at its back, its antenna-gain should be enhanced. When the reflector is placed below the main radiating element, its back-radiated field is reflected and added to the forward radiated field, which will possibly increase the antenna gain [9].

As is mentioned above, the optimization of the energy harvesting is a complicated problem, Because a small structure are more complicated and has more design parameters and it should overcome its originally low gain and low axial ratio(AR) of the circular polarization. So, we use and adaptive particle swarm optimization (APSO) algorithm to optimize the antenna design with numerous design parameters [10]. The APSO has a faster convergence in the process than the PSO which we are going to show in the paper.

The APSO algorithm is implemented in visual basic (VB), and connected to HFSS for finite element analysis through the script function to optimize the antenna in an accurate way [11]. The proposed design is also tested and verified to have the desired performance through simulations and experiments.

## 2. A Gain-Enhanced Antenna with Reflector

The efficiency of RF energy harvesting is estimated as the received RF power. According to the Friis transmission formula, the received RF power at the antenna can be expressed as

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$$P_r = P_t G_t \left( \frac{\lambda}{4\pi R} \right)^2 G_r \quad (1)$$

where,  $P_r$  is received power,  $P_t$  is transmitted power,  $G_t$  is transmitted antenna gain,  $G_r$  is received antenna gain,  $R$  is distance between transmitted and received antenna. Therefore, the received antenna with a high gain is a key factor to high received power. One of the ways to enhance the gain of the planar radiating element is to let the field interfere with each other. To do this, we suggest the geometry as in Fig. 1. It is composed of the ring-slot antenna backed by a reflector such as Fig. 1 (b).

The ring-slot patch antenna is operated in  $TM_{11}$  mode. Thus the circumference of the ring is the half wavelength of the operating frequency. Complying with the need to reduce the size of antenna, we insert slits as shown in Fig. 1 (a) to make the effective length of the current path along the ring increase. As an additional design element, the patch corners are cut to create a circular polarization for better reception of energy in an arbitrary angle position [12].

This circular polarized field is radiated forward and backward together. The backward wave hits the reflector and is reflected into the forward direction. With the distance between the radiating element and the reflector, the characteristics of the interference between the forward and reflected field change and can lower the gain. So, the optimal distance should be met. To explain the interference effect, Fig. 2 shows the radiation patterns with and without the reflector.

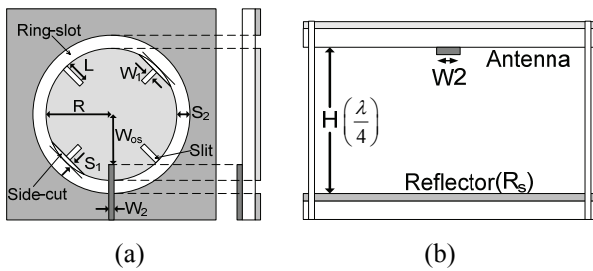


Fig. 1. The reflector ring-slot patch antenna with reflector and design parameters: (a) top; (b) side.

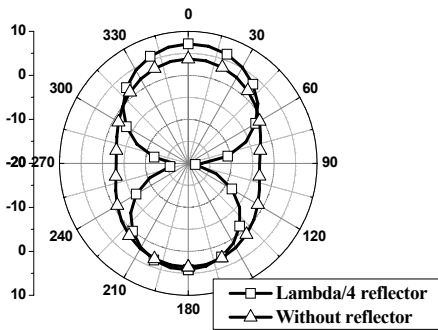


Fig. 2. The radiation pattern of antenna with and without reflector.

### 3. Optimal Design of Antenna

#### 3.1 Implementation of APSO algorithm

To find the optimal distance between the radiating element and reflector, we adopt the adaptive PSO (APSO). The APSO is a real-time evolutionary four state estimation procedure. Four state including exploration, exploitation, convergence, jumping-out in each iteration [10]. The flow chart of APSO algorithm shows in Fig. 3.

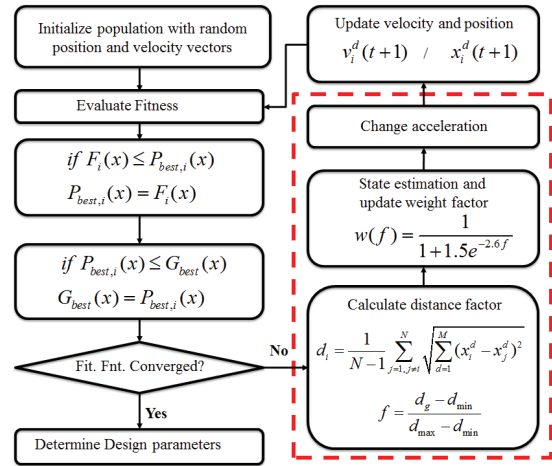


Fig. 3. The Flow chart of the APSO algorithm.

##### 3.1.1 Calculate distance factor

At the current position, the mean distance of each particle  $i$  is calculated to all the other particles. For example, this mean distance can be measured using an Euclidian metric

$$d_i = \frac{1}{N-1} \sum_{j=1, j \neq i}^N \sqrt{\sum_{d=1}^M (x_i^d - x_j^d)^2} \quad (2)$$

where  $N$  and  $M$  are the population size and the number of dimensions, respectively.

Denote  $d_i$  of the globally best particle as  $d_g$ . Compare all  $d_i$ 's, and determine the maximum and minimum distances  $d_{max}$  and  $d_{min}$ . Compute an "evolutionary factor"  $f$  as defined by

$$f = \frac{d_g - d_{min}}{d_{max} - d_{min}} \in [0,1] \quad (3)$$

##### 3.1.2 State estimation and update weight factor

Classify  $f$  into one of the four sets  $S_1$ ,  $S_2$ ,  $S_3$ , and  $S_4$ , which represent the states of exploration, exploitation, convergence, and jumping out, respectively. The key to fuzzy classification is overlap memberships. The formulation for numerical implementation of the classification is as follows.

Exploration : A medium to large value of  $f$  represents  $S_1$ ,  
 Exploitation : A shrunk value of  $f$  represents  $S_2$ ,  
 Convergence : A minimal value of  $f$  represents  $S_3$ ,  
 Jumping-out : When PSO is jumping out of a local optimum. The largest value of  $f$  reflect  $S_4$ ,

It would be beneficial to allow  $w$  to follow the evolutionary states using a sigmoid mapping  $w(f)$

$$\omega(f) = \frac{1}{1+1.5e^{-2.6f}} \quad \forall f \in [0,1] \quad (4)$$

### 3.1.3 Change acceleration

Parameter  $c_1$  represents the “self-cognition” that pulls the particle to its own historical best position. Parameter  $c_2$  represents the “social influence” that pushes the swarm to convergence to the current globally best region.

- Strategy 1- increasing  $c_1$  and decreasing  $c_2$
- Strategy 2- increasing  $c_1$  slightly and decreasing  $c_2$  slightly
- Strategy 3- increasing  $c_1$  slightly and increasing  $c_2$  slightly
- Strategy 4- decreasing  $c_1$  and increasing  $c_2$

### 3.1.4 Update velocity and position

The velocity of the particle is changed according to the relative locations of  $P_{best}$  and  $G_{best}$ . It is accelerated in the directions of these locations of greatest fitness according to the following equation:

$$v_i^d(t+1) = w(f) \times v_i^d(t) + c_1 \times r_1 \times (P_{best,i}(t) - x_i^d(t)) + c_2 \times r_2 \times (g_{best,i}(t) - x_i^d(t)) \quad (5)$$

where  $v_i^d(t)$  is the velocity of the particle and  $x_i^d(t)$  is the particle’s coordinate in the  $m$  dimension and  $w$  is the inertial weight.  $c_1$  and  $c_2$  are scaling factors of stochastic probability that determine the relative attraction to  $P_{best}$  and  $G_{best}$ .  $r_1$  and  $r_2$  are random variables whose value is provided between 0 and 1. The new position of each particle is updated according to the linear searching with an arbitrary interval.

## 3.2 Design goal and parameter of the antenna

We implemented the APSO program for the antenna problem such that the dimension of the system is 9 which is the number of the design parameters, and the number of particles is 15. For the part of Finite Element Analysis of the model, the HFSS is used [12]. The APSO is written in Visual Basic that is linked to the HFSS through the script function of the HFSS.

The design goal is to obtain a global optimum for making the antenna’s resonant frequency at 2.45GHz, CP, enhanced gain, and suppressing the harmonics radiation.

To improve the antenna gain, we minimize back-radiated

field and maximize forward radiated field. So, main lobe level of the radiation pattern is upper 7dBi and back lobe level is lower 0dBi. Also, the S11 of resonant band (2.2~2.6GHz) and stop band (3~8 GHz) is below -12 dB and above -8 dB, respectively. To realize circular polarization, the AR should be less than 3 dB in resonant frequency.

The fitness function of the APSO with constraints on the resonance frequency, return loss, and gain is defined as;

$$F_i(x) = \alpha_1 F_1(x) + \alpha_2 F_2(x) + \beta_1 G_1(x) + \beta_2 G_2(x) + A_f(x) \quad (6)$$

where,

$$F_1(x) = \int_{f_1}^{f_2} (s(x_i) - s_1)^2 df \quad \text{if } s > s_1$$

$$F_2(x) = \int_{f_3}^{f_4} (s(x_i) - s_2)^2 df \quad \text{if } s > s_2$$

$$G_1(x) = \int_{\theta_1}^{\theta_2} (g(\theta, f_{2.45GHz}) - g_1)^2 d\theta \quad \text{if } g < g_1$$

$$G_2(x) = \int_{\theta_3}^{\theta_4} (g(\theta, f_{2.45GHz}) - g_2)^2 d\theta \quad \text{if } g > g_2$$

$$A_f(x) = (a(x_i) - a_i)^2 \Big|_{f=2.45GHz} \quad \text{if } a_i > a_i$$

Functions  $F_{12}$ ,  $G_{12}$ , and  $A_f$  represent the resonant frequency, gain and axial ratio, respectively.  $s_{12}$ ,  $g_{12}$  and  $a_i$  denote the levels of return loss, the gain and the axial ratio of designated resonant frequency, respectively. We set the resonant frequency band from  $f_1=2.2$ GHz, to  $f_2=2.6$ GHz, and harmonic suppression band from  $f_3=3.0$ GHz to  $f_4=8.0$ GHz. Main lobe of the radiation is set from  $\theta_1=-20$ deg to  $\theta_2=20$ deg, and back lobe from  $\theta_1=-160$ deg to  $\theta_2=160$ deg. Upper and lower bound target values are  $s_1=-12$  dB,  $s_2=-8$ dB,  $g_1=7$ dBi,  $g_2=0$ dBi, respectively. Weight  $\alpha_1$ ,  $\alpha_2$ ,  $\beta_1$ ,  $\beta_2$  are the inverse values of the number of sampling points.

Some calculated values in the function which satisfied the requirement in the earlier stage may produce too large a value to the whole function. It should be adjusted to a proper value in minimizing the objective function. As a numerical manipulation, putting the function value 0 when it satisfied the required limit as (6) is much helpful.

## 3.3 Optimal design result

Initial and Optimal parameters are shown in Table 1 and the convergence of fitness function is illustrated in Fig. 4. It can be found that the band-basis objective function works very well for both the resonance and suppression bands and perform a global search with faster convergence

**Table 1.** The values of design parameters in antenna [mm]

| State   | W <sub>os</sub> | R     | S <sub>1</sub> | S <sub>2</sub> | L | W <sub>1</sub> | W <sub>2</sub> | H  | R <sub>s</sub> |
|---------|-----------------|-------|----------------|----------------|---|----------------|----------------|----|----------------|
| Initial | 1.71            | 14.46 | 2.62           | 0.53           | 3 | 1.53           | 1              | 31 | 50             |
| Max     | 4               | 16    | 3              | 1.5            | 8 | 2.5            | 2              | 33 | 70             |
| Min     | 1               | 13    | 1              | 0.3            | 2 | 1              | 1              | 29 | 50             |
| Optimal | 1               | 15.91 | 3              | 0.89           | 2 | 1              | 1              | 29 | 50             |

speed than PSO algorithm. The optimal results of APSO after 57 iterations are presented in Fig. 4 but the PSO algorithm not convergence.

The variations of the characteristics of the antenna are shown in Fig. 5~7. The optimal result of return loss is -12 dB under at resonant frequency band, AR goes to 1.91 dB

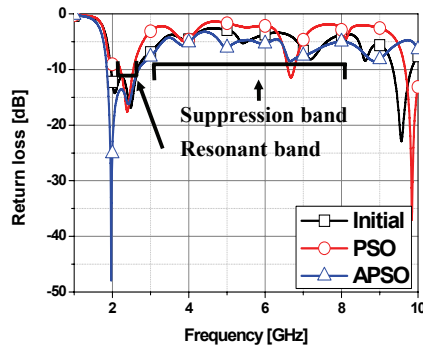


Fig. 4. The return loss distribution after optimization.

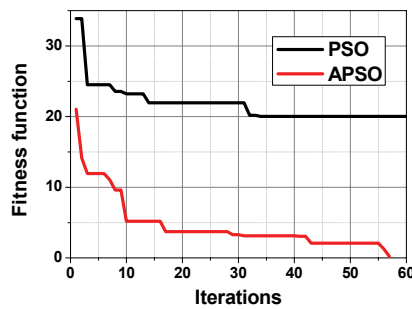


Fig. 5. Convergence of the fitness function APSO and PSO.

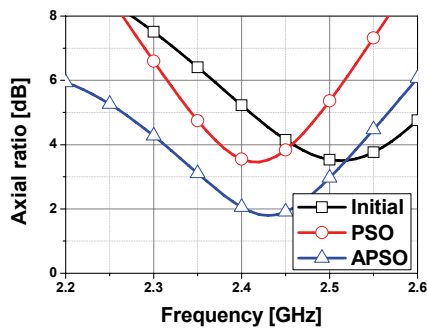


Fig. 6. The axial ratio after optimization.

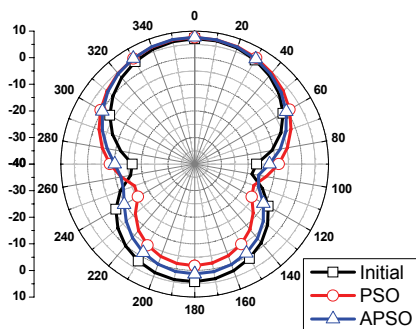


Fig. 7. The radiation pattern after optimization.

on the 2.45 GHz, and gain of antenna has main lobe 7.69 dBi, back lobe 1.32 dBi, which can be proved that it reaches the proposed design goal.

#### 4. Manufacturing and Measurement

The proposed design is fabricated on FR-4 substrate and measured as shown in Fig. 8~9. The predicted and the measured return loss are compared in Fig. 8, where the design goals are denoted with s1, and s2. The measured bandwidth is from 2.2~2.6 GHz and 3~8 GHz, which shows a good agreement with the simulation results and the design goal.

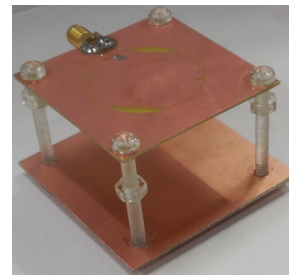


Fig. 8. The fabrication of the proposed reflector antenna.

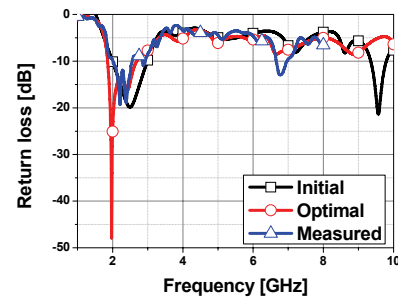


Fig. 9. Comparison of the measured and the simulated return loss for optimal design.

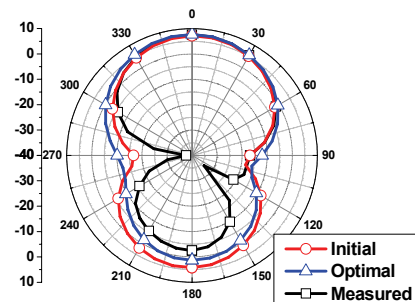


Fig. 10. Comparison of the measured and the simulated radiation pattern for optimal design.

#### 5. Conclusion

A circular polarized antenna with enhanced gain for energy harvesting was designed. We placed the reflector

below the main radiating element and its back-radiated field is reflected and added to the forward radiated field, which could increase the antenna gain.

We adopted the adaptive particle swarm optimization (APSO) algorithm, which improves the speed of convergence with a high frequency solver. The analysis reveal that the fitness function converges very well with APSO optimization algorithm and the performance of the optimized design successfully satisfied the design goal of the frequency band, gain and axial ratio.

The proposed optimized antenna was fabricated. We showed the simulated & measured results to verify the performance of the antenna. We expect that the optimized antenna can be used for a variety of applications.

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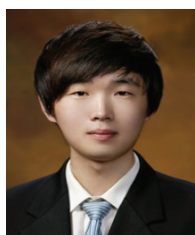
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