

지구관측위성 광대역 신호 발생기 구현

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Wideband Signal Generator Implementation for Earth Observation Satellite

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요 약

지구 관측 영상을 획득하는 합성개구레이더의 해상도를 향상시키기 위해서는 광대역 칩신호 발생이 필수적으로 요구된다. 본 논문에서는 고해상도를 얻기 위한 저궤도 관측위성용 합성개구레이더의 광대역 신호 발생기를 설계하고 시험 모델 제작 및 기능 시험 결과를 다루었다. 구현된 광대역 신호 발생기의 파형발생기는 위성에서 주로 적용되는 메모리맵 기반의 구조를 사용하였으며 내부는 파형 발생을 위한 디지털 모듈과 직교 변조를 위한 RF 모듈로 구성된다. 디지털 모듈의 메모리에 저장된 I/Q 신호는 D/A 변환기를 거쳐 RF 모듈로 전달되며 1275 MHz 기준 신호에 대해 직교 변조기를 거쳐 변조된다. 광대역 신호 발생기 검증을 위한 치구 및 GUI도 개발하였다. 시험 결과 대역폭 요구사항 144 MHz를 잘 만족하고 있음을 확인하였다. 또한 사전 왜곡 보상 기능을 구현하여 발생된 왜곡이 보상됨을 확인하였다.

Key Words : earth observation satellites; synthetic aperture radar; digital chirp signal generator; quadrature modulator; pre-distortion compensation.

ABSTRACT

The wideband chirp signal generator to enhance the resolution of synthetic aperture radar of obtaining the earth observation image is needed. This paper deals with designing, manufacturing and testing the wideband digital chirp signal generator having high resolution for LEO earth observation satellite. The wideband digital chirp signal generator is implemented with the memory-map based structure which is mostly applied in the satellite, and consists of the digital module to generate the digital chirp signal and the RF module to perform the quadrature modulation. The I/Q signals stored in the memory of the digital module are D/A converted and delivered to be quadrature modulated with the reference signal of 1275 MHz in the RF module. Furthermore, the test bench and GUI to validate the signal generator function are also developed. It is found that the requirement of 144 MHz bandwidth for the digital chirp signal generator is well met. Finally it is noteworthy that the distortion occurred in the chirp signal generator was compensated by the pre-distortion compensation.

I. Introduction

The synthetic aperture radar (SAR) basically requires a wideband chirp pulse signal generator to get high resolution images [1]. Typical methods for generating wideband chirp signals are DDS (direct digital synthesizer and memory map [2]. The DDS has advantage in generating various signal sources but has not yet fully space-qualified in space, so the memory mapped method

has been widely used for the space-borne SAR such as Terra-SAR and COSMO-Skymed program.

This paper focuses on developing the wideband chirp pulse generator having the bandwidth of 144 MHz at L-band 1275 MHz which will be expanded by frequency multiplier to generate X-band chirp signal with the bandwidth of 576 MHz.

In this paper, the wideband digital chirp signal generator is implemented with the memory-map based

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structure commonly used in a spaceborne SAR, and largely consists of the digital module to generate the digital chirp signal and the RF module to generate the quadrature modulated L-band signal of 1275 MHz. The I/Q signals stored in the memory of the digital module are D/A converted and delivered to be quadrature modulated with the reference LO signal of 1275 MHz in the RF module. Furthermore, PC GUI program and the test bench were also developed to validate the requirements for the chirp signals. PC GUI generates the chirp signals to be memorized in the memory of the digital module.

The amplitude and phase error coefficients of the received chirp signal are extracted and then the pre-distortion amplitude and phase coefficients are extracted [3]. With the extracted pre-distortion amplitude and phase coefficients, the pre-distorted chirp pulse is generated and applied to the chirp signal generator. The amplitude and phase error coefficients are recalculated and compared to the requirements. IRF (Impulse Response Function) of the chirp signal down-converted by the test bench is analyzed and the measured IRF is compared to the ideal one. It is found that test results of the chirp signal generator are all met with the requirements.

II. Design of Wideband Chirp Signal Generator

1. Required Design Parameters

The key requirements for the chirp signal generator are defined in the Table 1.

Table 1. Required Design Parameters

Parameters	Requirements
Bandwidth	≤ 144 MHz
PRF	2000 - 5000 Hz
Pulse Width	1 - 100 us
Output Frequency	1275MHz \pm 50kHz
Output Power	0 dBm \pm 0.5dB
VSWR	$\leq 2:1$

The general chirp signal equation is defined as follows.

$$x(t) = \text{rect}\left(\frac{t}{T}\right) A e^{j\pi\beta t^2} \quad (1)$$

$$x(t) = \text{rect}\left(\frac{t}{T}\right) A [\cos(\pi\beta t^2) + j \sin(\pi\beta t^2)]$$

where A is the amplitude, T is the pulse duration and β is the chirp rate in hertz per second and t is the time variable in seconds.

Table 2 and Table 3 show the amplitude and phase distortion requirements which will be used to check that the amplitude and phase error of the pre-distorted chirp signal can meet the requirements.

Table 2. Amplitude Distortion Requirements

Amplitude	Within Pulse		Pulse to Pulse
	absolute	relative	
linear [dB/us]	0.01	0.15	0.1
quadratic [dB/us]	0.01	0.06	0.05
random [dBrms]	0.7	0.3	0.15

Table 3. Phase Distortion Requirements

Phase	Within Pulse		Pulse to Pulse
	absolute	relative	
linear [deg/us]	0.125	1.2	5
quadratic [deg/us]	0.125	1.8	1.2
random [degrms]	5	2.5	2.8

2. Configuration of Wideband Chirp Signal Generator and Test Bench and PC GUI

Fig. 1 shows the functional block diagram of wideband chirp signal generator and the test bench as well as the PC GUI. The chirp signal generator largely consists of the digital module to generate the digital chirp signal and the RF module to perform the quadrature modulation and amplification and filtering. PC GUI sends I/Q chirp data to be stored in the memory in the digital module. The chirp digital pulse is generated with I/Q chirp data stored in the memory and then goes through DAC and LPF and entered the RF module to be I/Q modulated signal with 1275 MHz local oscillator, and the I/Q modulated signal is filtered by BPF (Band-Pass Filter) and amplified and lastly filtered by LPF (Low-Pass Filter).

Also Fig. 1 shows the functional block diagram of test bench. The test bench largely consists of the RF module to perform the frequency down-conversion from the digital chirp signal of 1275 MHz to 500 MHz and the digital module to change the down-converted 500 MHz signal into the IF digital signal by ADC with 400 MHz PLL signal. Finally the IF digital signal is digitally I/Q down-converted and image-rejected and decimated and analyzed in the PC GUI for the amplitude and phase error estimation and the results are displayed on PC GUI.

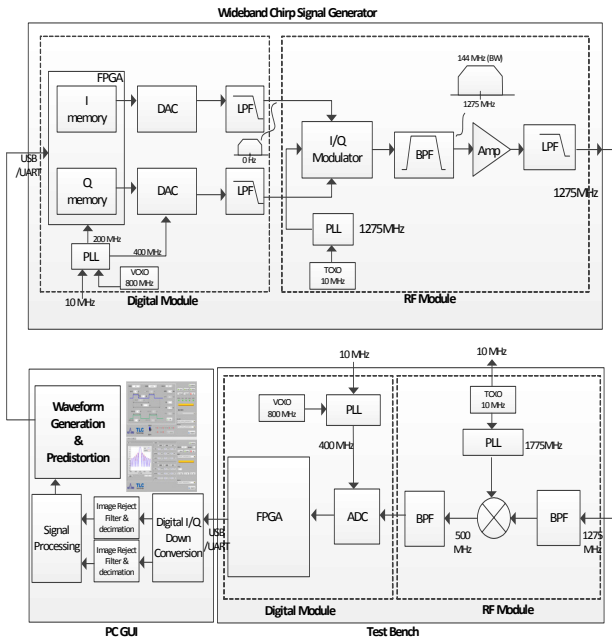


Fig. 1. Configurations of Wideband Chirp Signal Generator and Test Bench and PC GUI

3. Manufactured Board Configurations

Fig. 2 shows the manufactured digital board and RF board for the digital chirp signal. The FPGA and DAC used are V5-FX70T and AD9122, respectively. ADL5375 is used for I/Q modulator. Fig. 3 shows the manufactured digital board and RF board for the test bench.

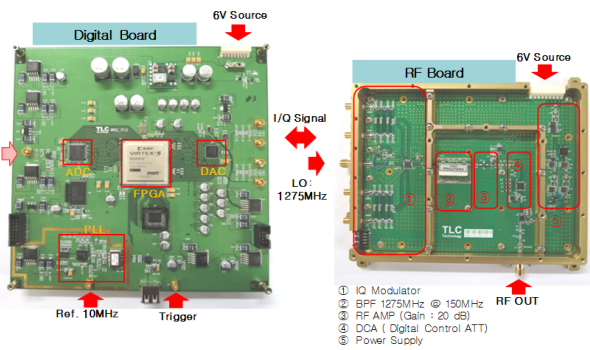


Fig. 2. Wideband Signal Generator Boards

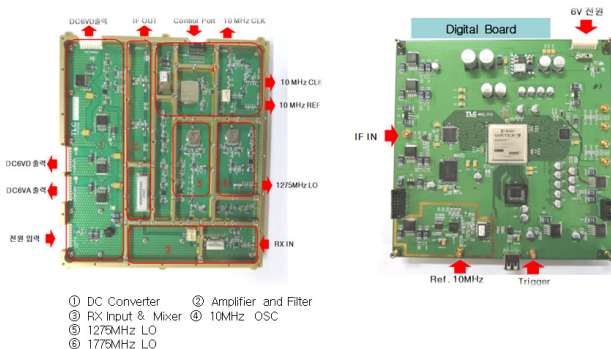


Fig. 3. Test Bench Boards

III. Test Results

In order to make sure that the chirp signal is well generated from I/Q memory in Fig. 1, the output signal of the DAC is measured. As shown in Fig. 4 as an example, it shows that the chirp signal is well matched with the shape of I/Q chirp signal at the bandwidth condition of 144 MHz, 100 μs pulse width, PRF 2000 Hz, and 200 MHz sampling rate.

The output spectrum is measured by using Agilent VSA to check the bandwidth requirement of 144 MHz. It is shown in Fig. 5 that the measured bandwidth meets the bandwidth requirement of 144 MHz, showing pretty good chirp frequency and phase variation.

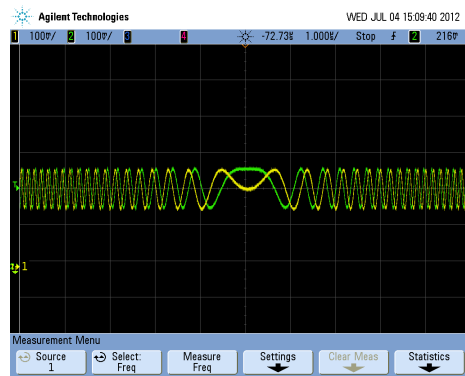


Fig. 4. I/Q Chirp Signals measured

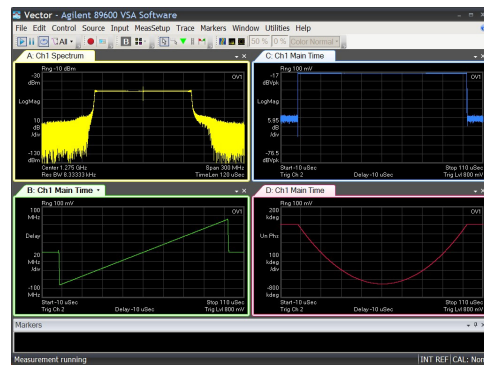


Fig. 5. Bandwidth Measurement

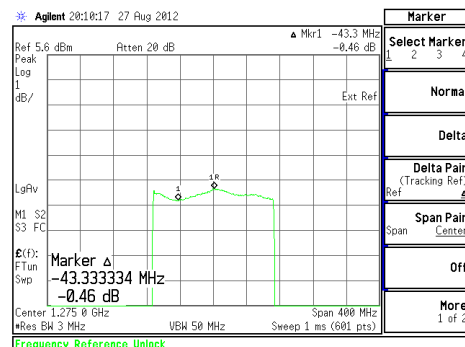


Fig. 6. Output Power Variation Measurement

Fig. 6 shows the output power variation of -0.46 dB. It meets the output power variation requirement of 0 dBm \pm 0.5 dB. Also, VSWR measured at the output of the chirp signal generator was 1.1891 which meets the VSWR requirement of less than 2:1.

Finally the pre-distortion compensation function was tested to see if the amplitude and phase error are within the requirements defined in Table 2 and Table 3.

First of all, Fig. 7 shows the received chirp pulse signal distorted by the chirp signal generator before the pre-distortion compensation.

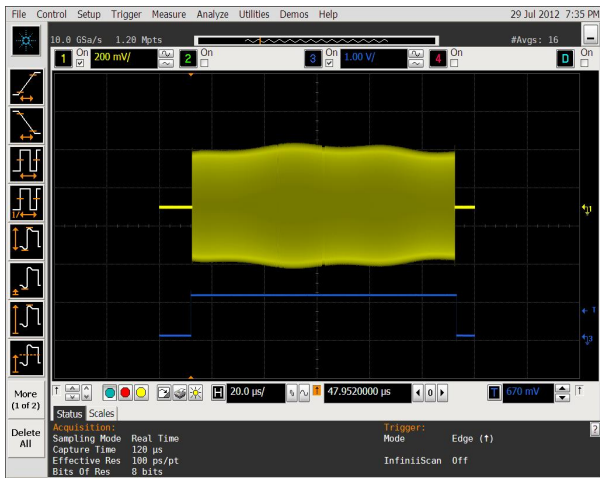


Fig. 7. I/Q Chirp Signals measured Before Pre- distortion

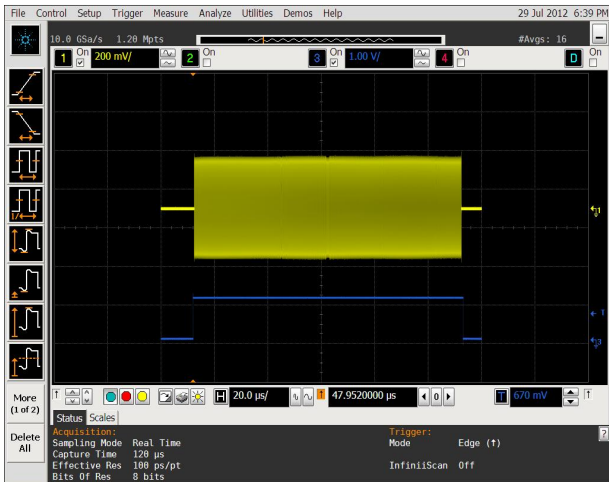


Fig. 8. I/Q Chirp Signals measured after pre-distortion

Fig. 8 shows the received chirp pulse signal after the pre-distortion compensation is applied.

Table 4. Amplitude and Phase Distortion Measured

Amplitude	Within Pulse (Absolute)		
	Req.	Measured (Before PD)	Measured (After PD)
linear [dB/us]	0.01	0.022236	0.000128
quadratic [dB/us]	0.01	0.000030	0.000000
random [dBrms]	0.7	0.629498	0.005798
Amplitude	Within Pulse (Relative)		
	Req.	Measured (Before PD)	Measured (After PD)
linear [dB/us]	0.15	0.000058	0.000032
quadratic [dB/us]	0.06	0.000000	0.000000
random [dBrms]	0.3	0.004419	0.002869
Phase	Within Pulse (Absolute)		
	Req.	Measured (Before PD)	Measured (After PD)
linear [deg/us]	0.125	0.677398	0.038318
quadratic [deg/us]	0.125	0.000069	0.000019
random [degms]	5	63.404558	3.720878
Phase	Within Pulse (Relative)		
	Req.	Measured (Before PD)	Measured (After PD)
linear [deg/us]	1.2	0.001446	0.000026
quadratic [deg/us]	1.8	0.000002	0.000000
random [degms]	2.5	0.070539	0.001145
Amplitude	Between Pulse (Absolute)		
	Req.	Measured (Before PD)	Measured (After PD)
linear [dB/us]	0.1	0.000000	0.000000
quadratic [dB/us]	0.05	0.000000	0.000000
random [dBrms]	0.15	0.000000	0.000000
Phase	Between Pulse (Relative)		
	Req.	Measured (Before PD)	Measured (After PD)
linear [deg/us]	5	2.160426	0.191203
quadratic [deg/us]	1.2	0.882302	0.000583
random [degms]	2.8	6.666101	1.866506

For the validation of the requirements defined in Table 2 and Table 3, the amplitude and phase errors of the chirp signal generator before pre-distortion and after pre-distortion are extracted. The random phase errors before pre-distortion did not meet the requirements, but after the predistortion is applied the random phase errors are well within the requirements.

Fig. 9 shows the IRF curve comparing the ideal one to

the measured one. The measured one before predistortion has higher sidelobe than the ideal one. After the pre-distortion compensation, it is shown in Fig. 10 that the measured IRF becomes almost the same as the ideal one.

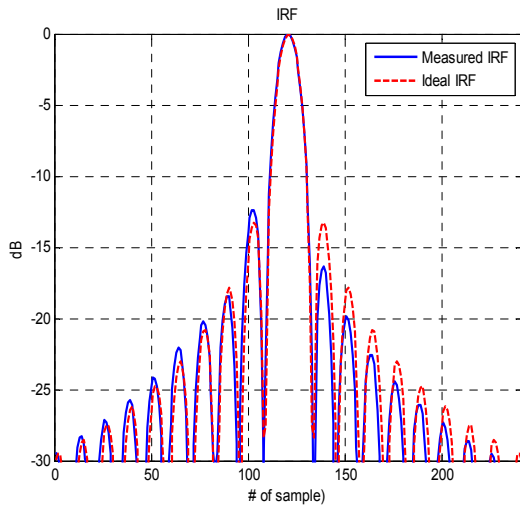


Fig. 9. IRF Before Pre-distortion

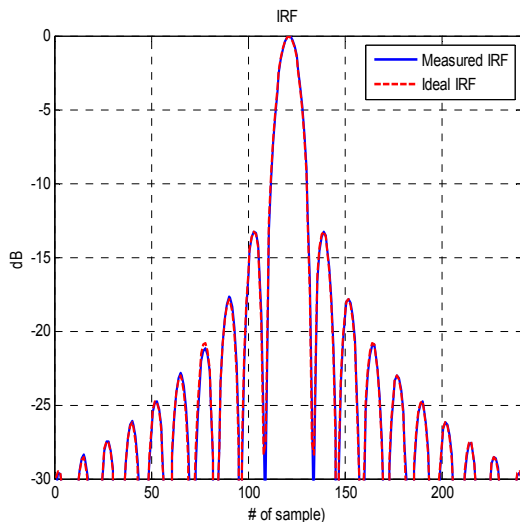


Fig. 10. IRF After Pre-distortion

IV. Conclusions

The wideband digital chirp signal generator is implemented with the memory-map based structure as well as I/Q modulator, and largely consists of the digital module to generate the digital chirp signal and the RF module to perform the quadrature modulation. Furthermore, the test bench to validate the signal generator function is also developed. After various tests, it is found that the key requirements are within the

requirements. Also the amplitude and phase errors of the chirp signal generator before pre-distortion and after pre-distortion are measured. Especially, before the pre-distortion, the random phase distortion errors not meeting the distortion requirements were reduced to be well within in the requirements after pre-distortion. All results obtained will be utilized for developing the flight model of the memory map based wideband chirp signal generator for a spaceborne SAR satellite.

References

- [1] Ian G. Cumming, "Digital processing of synthetic aperture radar data", Artech House, 2005.
- [2] M. Y. Chua and V. C. Koo, "FPGA-based chirp generator for high resolution UAV SAR", PIER 99, pp. 71-88, 2009.
- [3] S. Y. Kim, N. H. Myung, "Wideband linear frequency modulated waveform compensation using system predistortion and phase coefficients extraction method", IEEE Microw. Wireless Compon. Lett., vol 17, pp 808-810, Nov. 2007.

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