# **Characteristics Analysis of Short Flux-path 4/3 SRM**

### Jin-Woo Ahn\*

Abstract – This paper presents a design and characteristics analysis of novel 2-phase 4/3 switched reluctance motor (SRM) with short flux path for an air-blower application. The desired air-blower is unidirectional application, and requires a wide positive torque region without torque dead-zone. In order to get a wide positive torque region without torque dead-zone during phase commutation, asymmetric inductance characteristic with non-uniform air-gap is considered. The proposed motor could be started at any rotor position with high efficiency drive. The proposed 2-phase 4/3 SRM is verified by finite element method analysis.

Keywords: Switched reluctance motor, Short magnetic flux-path, Two-phase, 4/3 pole SRM

#### 1. Introduction

Recently, high speed drives are much interested in the industrial applications to reduce the system size and to increase the drive efficiency. Especially, blowers, compressors, pump and spindle drives are suitable for the high speed motor drives, and the demand for the high speed motor system is much increased according to the industrial market. Some types of electric motor such as DC, permanent magnet and reluctance motor have been researched.

SRM has simple structure and inherent mechanical strength without rotor winding and permanent magnet. These mechanical structures are suitable for harsh environments and high temperature and high speed applications [1-4].

This paper presents of a 2-phase 4/3 SRM for an airblower. In this application, the impeller rotates only one direction, and it requires a wide positive torque without torque dead-zone.

The proposed 4/3 SRM has self-starting ability without dead-zone. The 4/3 SRM is employed 4 stator poles and 3 rotor poles. In order to get a wide positive torque region without torque dead-zone during phase commutation, asymmetric inductance characteristic with non-uniform airgap is considered.

The proposed motor is to employ a short flux path. The advantages of short flux path are increasing the efficiency and torque production while decreasing the core loses.

In the paper, a novel 4/3 SRM is proposed and the

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characteristics are analyzed to optimize the magnetic circuit of the motor.

### 2. Design of 2-Phase 4/3 SRM

The design and the basic principles of the novel 2-phase 4/3 SRM with short magnetic flux is to be considered. And the FEA analyses of the proposed motor are presented. In order to show the advantages of the proposed motor, a comparison of characteristics of the 4/2 and 4/3 SRM is shown. Table I shows the specifications of the proposed 4/3 SRM.

**Table 1.** Specification of the Proposed 4/3 SRM

Parameters	Value
Output power	210 [W]
Rated torque	0.2 [Nm]
Rated speed	10,000 [rpm]
Number of stator poles	4
Number of rotor poles	3
Bore diameter	32 [mm]
Outer diameter	90 [mm]
Stack length	33 [mm]
Number of turns per phase	120 [turns]
Resistance of phase winding	0.4 [Ohm]
Unaligned inductance	2 [mH]

#### 2.1 Basic principles of 4/3 SRM

As known, high electrical frequency causes the core losses in the electric machine. In order to reduce the electrical frequency, reduction of the number of rotor poles is adopted. In this research, the rotor structure must be ensured that the motor can produce continuous

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torque with short magnetic flux and the electrical frequency as low as possible. Considering these two requirements, a rotor structure with three rotor poles was selected.

Fig. 1 shows the 2-phase 4/3 SRM structure. The proposed motor has 4 stator poles and 3 rotor poles. The stator core is separated into two c-cores with two poles each, and there are no steel lamination between these c-cores. The stator core is fixed by a plate. Each phase consists of windings on two poles of the c-core, and these windings are connected in serial. With this structure, the magnetic flux flows through the stator core are to be shortest.

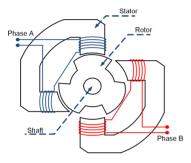


Fig. 1. 2-phase 4/3 SRM structure

The stator poles of the 4/3 SRM are placed at an angle of  $120^{0}$  that is equal to the rotor pole pitch. And two c-cores are shifted with rotation of  $180^{0}$ .

Electrical frequency of the 4/3 SRM is six times higher than rotation frequency. So, dwell angle is kept of  $60^0$  to produce the continuous torque. In order to get a high torque density, ideal stator pole arc is equal to dwell angle. But there are some manufacturing difficulties, so the stator pole arc was designed with  $50^0$ .

The disadvantage of the one-phase and 2-phase motor is self-starting ability, and the 4/3 SRM is not an exception. To overcome this problem, a wider rotor pole arc with non-uniform shape is used. This type of rotor shape is suitable for the applications which require only one rotational direction as air-blowers in this research. The proposed rotor pole arc is  $70^{\circ}$ . The non-uniform shape is applied for the first  $25^{\circ}$  of the rotor poles as shown in Fig. 2.

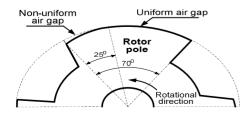


Fig. 2. Rotor shape

The air-gap is determined to develop a flat torque and the rotor shape is stepper type. With geometric structure, the inductance in the motor depends on the air-gap length and overlap area between rotor and stator poles. When the rotor and stator pole are at first overlap region, increasing overlap area results in increasing inductance. Otherwise, when the rotor and stator poles are at full aligned position, the overlap area is constant, reducing of average air-gap between rotor and stator pole results in increasing inductance. So, the motor can produce positive torque during 700 of rotor poles.

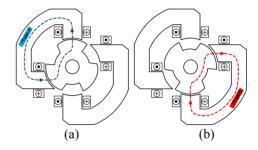


Fig. 3. Magnetic flux path (a) Phase A; (b) Phase B

Fig. 3 shows the magnetic flux path in the proposed motor at the aligned position. Because of separating of the c-cores, the magnetic flux flows only through the excited poles, and it can't flow through the other poles.

To keep the rotor poles not touching into the stator poles, the stator core is fixed by an aluminum plate. The plate has no influences on flux distribution of the proposed motor.

Fig. 4 shows the full assembly of the proposed 4/3 SRM.

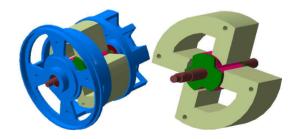


Fig. 4. Full assembly of the 4/3 SRM

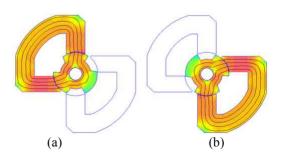


Fig.. 5. Flux distribution (a) Phase A; (b) Phase B

#### 2.2 FEA Analyses

Fig. 5 shows the flux distribution of the proposed motor. In this paper, the characteristic of the proposed motor is analyzed by using a 2D FEA analysis. By using the FEA software, the characteristic of the flux distribution that influences the performance and characteristic of the motor is analyzed. As shown in Fig. 5, the magnetic flux only flows through the respective excited c-core. Actually, small amount of flux flows through the other, but it is not appreciable.

Torque is one of the important performance parameters of the motor. The higher torque density that means the motor size and the cost can be reduced. Compared with the other motor which have a uniform rotor shape, the proposed motor has high average torque.

Torque characteristics depend on the relationship between flux linkages and rotor position as a function of phase current. Torque varies with rotor position and phase current. The torque production can be estimated by (1).

$$T_e = \frac{1}{2}i^2 \frac{dL(\theta, i)}{d\theta} \tag{1}$$

where  $T_e$  is the electromagnetic torque, i is the excited current, and  $L(\theta,i)$  is the inductance dependent on the rotor position and phase current.

By using the 2D FEA, the inductance and torque characteristics are analyzed. Fig. 6 shows the torque profile of the proposed motor. As shown, the overlap torque is about 100. This overlap torque ensures that the motor can be operated without the torque dead-zone and has self-starting ability.

Fig. 6 shows the inductance profile of the proposed motor. As the stator placement and rotor shape, the inductance profile has curved and symmetric shape.

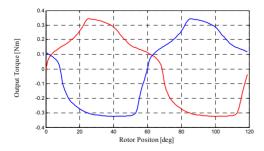


Fig. 6. Torque profile

Fig. 7 shows the torque comparison of the 4/2 SRM and the proposed 4/3 SRM. Both of 4/2 and 4/3 SRM use the

same dimension and input parameters. In this analysis, the 4/2 SRM has a stepper rotor shape [8]. In order to compare their torque, the torque profile of 4/2 and 4/3 SRM is analyzed in the same electrical position.

As shown in Fig. 8, the average torque of 4/3 SRM is 0.21[Nm] while average torque of 4/2 SRM is 0.16[Nm]. The average torque of the 4/3 SRM is 31% higher than the 4/2 SRM.

## 3. Simulation of proposed motor

Fig. 9 shows the simulation diagram of 4/3 SRM. The simulation diagram consists of the 4/3 SRM model, inverter part, speed and current control part.

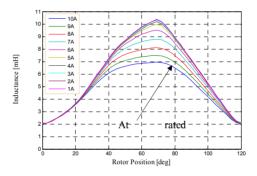


Fig. 7. Inductance profile

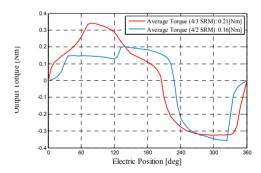


Fig. 8. Torque of the 4/2 and the proposed 4/3 SRM

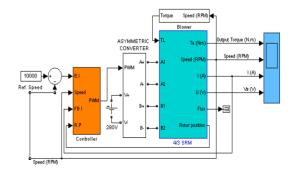


Fig. 9. Simulation diagram of the proposed motor

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After the proposed motor was analyzed by the FEA software, the full torque and inductance were analyzed. The 2D look-up tables of the current verse rotor position and flux linkage and of the torque verse rotor position and current were used for simulation.

The relationship between voltage, resistance and inductance of SRM is given by (2).

$$v = R_s i + \frac{d\lambda(\theta, i)}{dt} \tag{2}$$

where,  $R_s$  is the phase resistance, v is the voltage applied across the phase winding, and  $\lambda(\theta, i)$  is the phase flux linkages depending on rotor position and excited current.

The voltage drop of the SRM can be determined by using the phase flux linkage data and phase resistance. The speed of the SRM can be determined by using the load equation. The load for simulation consists of inertia, friction and load torque. The load equation is given by (3).

$$J\frac{d\omega_m}{dt} + B\omega_m = T_e - T_l \tag{3}$$

where J is the combined moment of inertia of the motor and load, B is the combined friction coefficient of the motor and load,  $T_I$  is load torque.

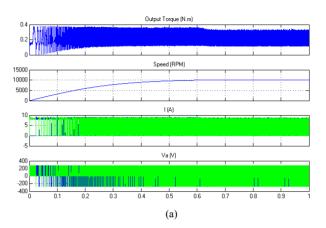
From (3) the speed of SRM is determined by (4).

$$\omega_m = \int \frac{T_e - T_l - B\omega_m}{J} dt \tag{4}$$

The number of turns per stator poles is only one in the FEA analysis. After that, the number of turns is determined by observing tail of current when it is turned off. High turn number results in low current/torque ratio and long tail of current and vice versa. Low current/torque ratio is good index for performance whereas long tail of current will produce negative torque. Balancing these two things, 50 turns/pole was selected.

As shown in Fig. 10 (a), the speed rising time is about 0.6s, it is fast enough for system response. The speed rising time depends on the unaligned inductance, current slope, and the system inertia moment. The lower unaligned inductance results in the lower speed rising time while rated current and voltage are kept constant. In order to reduce the unaligned inductance, the turn number of the windings can be reduced while increasing the excited current, and keeping the mmf is constant.

Fig. 10 (b) shows the simulation results of 4/3 SRM after adjusting the zoom. In this simulation, the operational parameters are the same with real system. The proposed motor is operated at 10,000rpm with excited current is 8A and the supplied dc voltage is 280V.



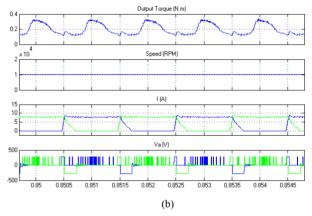


Fig. 10. Simulation results of the proposed motor

#### 4. Conclusion

This paper presents a novel 2-phase 4/3 SRM which has short flux path to reduce core losses and increase torque production. In order to produce a continuous output torque, the positive torque region is extended with asymmetric inductance characteristic. The air-gap is also determined to reduce torque ripple. The proposed motor is suitable for applications that require unidirectional rotation and is easy to manufacturing because of simple stator shape.

#### Acknowledgment

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