

Iron Loss Analysis of a Permanent Magnet Rotating Machine Taking Account of the Vector Hysteretic Properties of Electrical Steel Sheet

Heesung Yoon *, Seok-Myeong Jang ** and Chang Seop Koh *

Abstract – This paper presents the iron loss prediction of rotating electric machines taking account of the vector hysteretic properties of electrical steel sheet. The E&S vector hysteresis model is adopted to describe the vector hysteretic properties of a non-oriented electrical steel sheet, and incorporated into finite element analysis (FEA) for magnetic field analysis and iron loss prediction. A permanent magnet synchronous generator is taken as a numerical model, and the analyzed magnetic field distribution and predicted iron loss by using the proposed method is compared with those from a conventional method which employs an empirical iron loss formula with FEA based on a non-linear B - H curve. Through the comparison the effectiveness of the presented method for the iron loss prediction of the rotating machine is verified.

Keywords: E&S vector hysteresis model, Finite element method, Synchronous generator

1. Introduction

There is increasing demands to develop high-efficiency rotating electric machines to reduce energy consumption. In order to meet the demands, it is essential to predict the iron loss accurately at the design stage because the iron loss is of great importance in the total energy loss. The iron loss from rotating magnetic fields is in general significantly bigger than that from alternating ones even at the same magnitude of magnetic flux density, and accordingly its precise prediction is essential in the estimation of iron loss of rotating electrical machines [1]. The rotational iron loss, however, involves complicated vector hysteretic properties of an electrical steel sheet (ESS), and its precise prediction still remains complex and arduous works [2].

Conventionally the rotational iron loss has been estimated by using empirical iron loss formula based on Steinmetz equation [3]-[6]. In these researches, both the hysteresis and eddy current losses are calculated from the magnetic field distribution obtained by using FEA which utilizes a single non-linear B - H curve ignoring the vector hysteretic proper-

ties of ESS. Especially, the iron loss for an arbitrary rotating B -waveform is commonly estimated taking into account the ratio of minor to major axis of the approximated elliptic B -waveform with different coefficients of the Steinmetz equation from those for alternating B -waveform. Even if the empirical formula may estimate the rotational iron loss correctly, these methods have obvious error because of the exclusion of the vector hysteretic properties of ESS at the magnetic field analysis [7].

Recently vector hysteresis models such as Jiles-Atherton model [7], vector Preisach model [8], vector play model [9] and E&S (Enokizono&Soda) model [10]-[12] have been developed to directly describe the vector hysteretic properties of ESS, and combined with FEA to be applied for magnetic field and iron loss analyses of electric machines. These methods commonly calculate the iron loss directly from the B - and its corresponding H -waveforms instead of using an empirical formula. Among them, E&S model is an attractive engineering model because it gives more accurate iron losses under not only the alternating but also the rotating magnetic fields than the others [10]-[11]. For precise iron loss prediction under distorted magnetic field, recently, the dynamic E&S model is suggested by improving eddy current term and coefficient calculation algorithm [12], [13].

In this paper, the E&S vector hysteresis model is combined with FEA and applied to the analysis of the iron loss of a permanent magnet synchronous generator (PMSG).

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2. Vector Hysteretic Properties of Non-oriented Electrical Steel Sheet

In order to measure the vector hysteretic properties of a non-oriented ESS (35PN440), a two-dimensional single sheet tester (2-D SST) developed in [14] is used. The H -waveforms corresponding to arbitrary elliptic B -waveforms with geometric parameters, B_{\max} , α and φ , defined in Fig. 1 are measured at 50Hz adopting B -waveform control.

Fig. 2 shows the H -waveforms measured under rotating B -waveforms. Even though the B -waveforms are purely circular, the corresponding H -waveforms are far from circular. The magnitude of H along the transverse direction (TD) is significantly larger than that along the rolling direction (RD) due to the anisotropic property of the non-oriented ESS. It is found that, in Fig. 2(d), there are phase differences between B - and H -vectors, and moreover they do not remain a constant with respect to the rotation of the B -vector.

Fig. 3 compares the measured iron losses according to the variations of B_{\max} and α when φ is equal to zero. It can be seen that the iron loss by rotating magnetic field ($\alpha=1$) is considerably larger than that by alternating one ($\alpha=0$) at the same B_{\max} . For example, at 1.2T of maximum magnetic flux density, the rotating magnetic field gives 87% more iron loss (2.37W/kg) than alternating one (1.27W/kg).

The above magnetic properties and iron loss characteristics come from the vector hysteretic properties of the ESS, and cannot be described precisely via the conventional method using empirical iron loss formula and FEA with non-linear B - H curve [3]-[6].

3. Iron Loss Prediction using FEA combined with E&S Vector Hysteresis Model

In order to describe the vector hysteretic properties, in this paper, the E&S vector hysteresis model is adopted. The E&S model defines the relationship between B - and H -waveforms as follows [10]-[12]:

$$H_k(\tau) = v_k B_k(\tau) + h_k \frac{\partial B_k(\tau)}{\partial \tau} + \frac{\sigma \omega d^2}{12} \frac{\partial B_k(\tau)}{\partial \tau} \quad (1)$$

$$k = R, T \text{ and } \tau = \omega t \in [0, 2\pi]$$

where the subscripts R and T denote RD and TD of the ESS, respectively, v_k and h_k are the magnetic reluctivity and hysteresis coefficients, respectively, ω is the angular frequency of the fundamental component of the B -waveform, B_1 is the

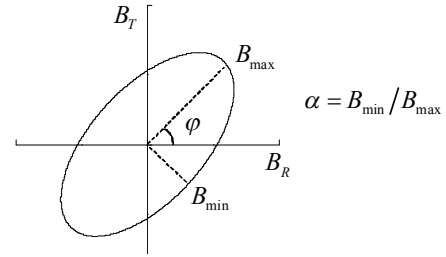


Fig. 1. Definition of B -waveform using B_{\max} , φ and α

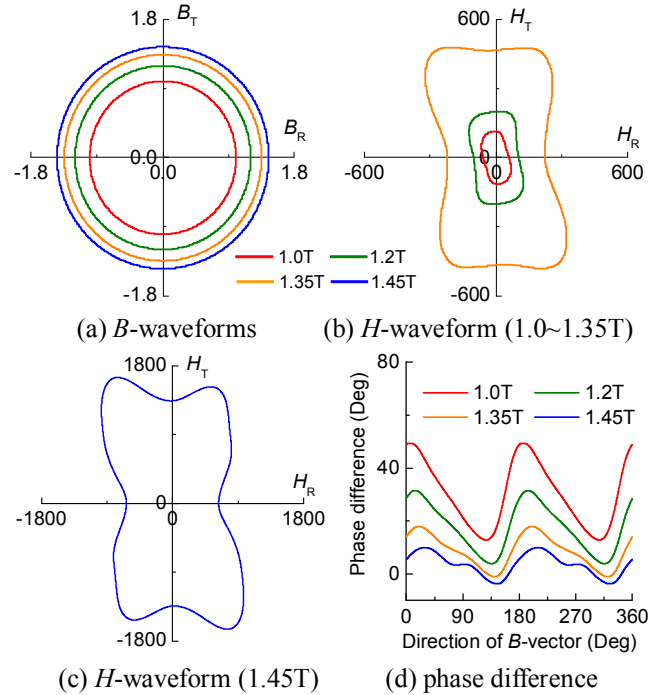


Fig. 2. Measured H -waveforms under rotating magnetic flux densities with variation of B_{\max}

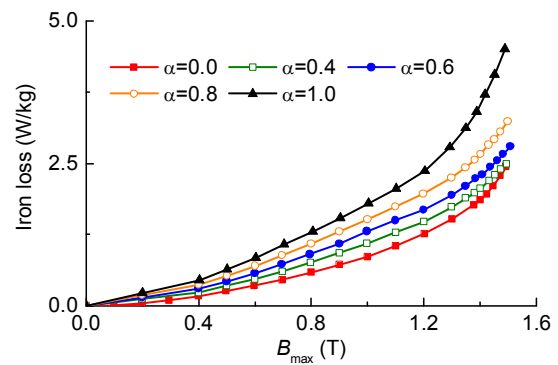


Fig. 3. Comparison of measured iron losses according to variations of B_{\max} and α when $\varphi=0^\circ$

fundamental component of the B -waveform, σ and d are the conductivity and thickness of the ESS, respectively.

The reluctivity and hysteresis coefficients in (1) are calculated by experimental data measured from the 2-D SST under various elliptic B -waveforms defined in Fig. 1 [11],

[14].

The governing equation incorporated with the E&S model is derived from Maxwell's equations as follows [12]:

$$\nabla \times \left(\mathbf{v} \nabla \times \mathbf{A} + \mathbf{h} \frac{\partial \mathbf{B}_1}{\partial \tau} + \frac{\sigma \omega d^2}{12} \frac{\partial (\nabla \times \mathbf{A})}{\partial \tau} \right) = \mathbf{J}_0 \quad (2)$$

where \mathbf{A} and \mathbf{J}_0 are vector magnetic potential and applied current density, respectively.

By solving the governing equation by means of finite element method (FEM), the local B -waveforms in the electrical machine are obtained, and then the corresponding H -waveforms can be calculated by using the E&S model of (1). The resultant iron loss is predicted directly from calculated B - and H -waveforms as follows [12]:

$$P = \frac{1}{\rho T} \int_0^T \left(H_x \frac{\partial B_x}{\partial \tau} + H_y \frac{\partial B_y}{\partial \tau} \right) d\tau \quad [W/kg] \quad (3)$$

where ρ is the mass density, T is the exciting period.

4. Numerical Results in Permanent Magnet Rotating Machine

A permanent magnet synchronous generator shown in Fig. 4 is taken as a numerical model. The synchronous generator is made of a non-oriented ESS (35PN440), and is assembled so that RD of the stator core coincides with x -axis. Table 1 shows the specification of the synchronous generator.

In order to investigate the influence of the vector hysteric properties of the ESS, the following conventional method considering the rotational iron loss is compared [5].

$$P = P_{hys-alt} + P_{eddy-alt} + P_{add-rot} \quad [W/kg]$$

$$P_{add-rot} = \gamma \frac{B_{min}}{B_{max}} (P_{hys-alt} + P_{eddy-alt}) \quad (4)$$

$$= \gamma \alpha (P_{hys-alt} + P_{eddy-alt})$$

where $P_{hys-alt}$ and $P_{eddy-alt}$ is the hysteresis and eddy current losses under alternating magnetic field, respectively, $P_{add-rot}$ is the additional iron loss caused by rotating magnetic field. The γ presents the rate of iron loss increment under rotating magnetic field to alternating iron loss, and may depend on (B_{max} , φ , α) of the B -waveform. For simplicity, in this paper, the γ is fixed to 0.87 obtained from the ratio of the rotational to alternating iron losses at $B_{max}=1.2T$ shown in Fig. 3.

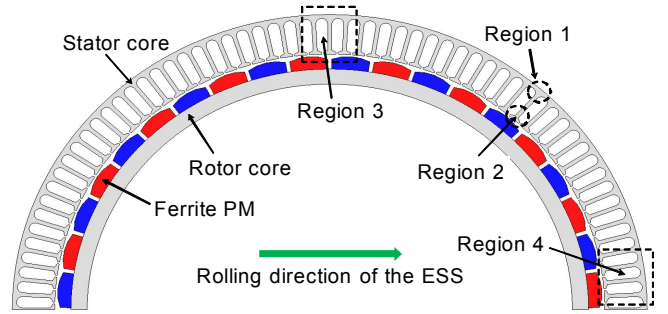


Fig. 4. A permanent magnet synchronous generator (half model)

Table 1. Specification of permanent magnet synchronous generator

Item		Value
Rated power (W)		400
Rated speed (rpm)		150
Number of phases and slots		3, 120
Electrical steel sheet		35PN440
Stator and rotor	Outer, inner radiuses of stator (mm)	157, 136
	Outer, inner radiuses of rotor (mm)	128, 120
	Thickness of return yoke (mm)	2
	Thickness of tooth (mm)	2
	Lamination (mm)	101
	Air-gap length (mm)	1
PM	Material	Ferrite
	Residual B (T)	0.4
	Coercive force (kA/m)	318.3
	Magnetization pattern	Parallel
	Width and thickness (mm)	18.5, 7
	Number of poles	40

The iron loss formula in (4) is expressed by considering both rotating magnetic field and flux density harmonics as follows [5]:

$$P_{hys-alt} + P_{eddy-alt} = \sum_n \left[k_h (nf) B_{mn}^\delta + k_e (nf)^2 B_{mn}^2 \right] \quad (5)$$

$$P = \sum_n \left[k_h (nf) B_{mn}^\delta + k_e (nf)^2 B_{mn}^2 \right] (1 + \gamma \alpha_n)$$

where δ is a Steinmetz constant, k_h and k_e are the hysteresis and eddy current coefficients fitted from measured alternating iron loss in Fig. 3, respectively, n is harmonic order of the B -waveform, B_{mn} and α_n are the magnitude and axis ratio of the n -th harmonic component of the B -waveform, respectively.

Fig. 5 shows the B - and H -loci calculated by the proposed method in the region 1 and 2 of the stator core defined in Fig. 4. It is found that the rotating magnetic fields

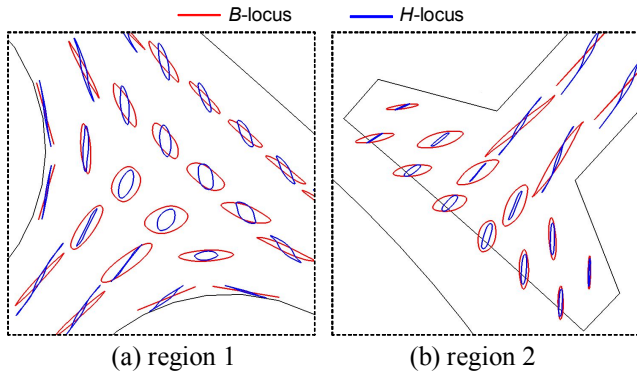


Fig. 5. B and H -loci, calculated by the proposed method, in the stator core

are distributed extensively in the return yoke and pole shoe of the stator core, and it is indicated that the rotational iron loss is of great importance to total iron loss in the stator core.

Fig. 6 and 7 compare the distributions of maximum magnetic flux densities (B_{max}) and maximum magnetic field intensities (H_{max}), calculated by the proposed and conventional methods. The conventional method gives same H_{max} distributions in the region 3 and 4 due to same B_{max} distributions. On the other hand, the proposed method gives higher H_{max} distributions in the region 3 than the region 4, especially at the stator teeth. This phenomenon results from the anisotropic magnetic property of the non-oriented ESS, and cannot be described precisely by the single non-linear B - H curve in the conventional method.

Fig. 8 compares the iron loss distributions calculated by the proposed and conventional methods. The following differences can be found in the iron loss distributions: 1) While the conventional method gives symmetrical loss distributions between the region 3 and 4, the proposed method gives asymmetric iron loss distributions, especially in the stator teeth. It is because that, the iron loss caused by the magnetic flux along the TD is considerably bigger than that cause by the magnetic flux along the RD due to the anisotropic magnetic properties [2]. 2) The rotational iron losses may be distributed in the re-turn yoke and pole shoe of the stator core where the rotating magnetic fields are existed as shown in Fig. 5. In the return yoke with the rotating fields of $\alpha > 0.8$, the conventional method gives similar rotational iron losses with the proposed one by the help of the γ . However, the large deviations of the rotational iron losses are observed in the pole shoe with the rotating fields of $\alpha < 0.8$. It means that, even if the conventional method adopts the additional term for the rotational iron loss, it still cannot predict precisely the rotational iron losses corresponding to various elliptic B -waveforms.

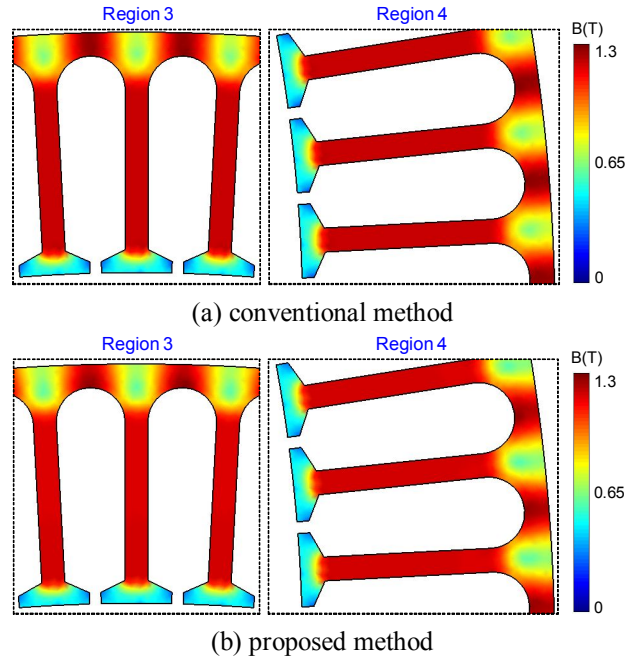


Fig. 6. Maximum magnetic flux densities calculated by the conventional and proposed methods in the stator core

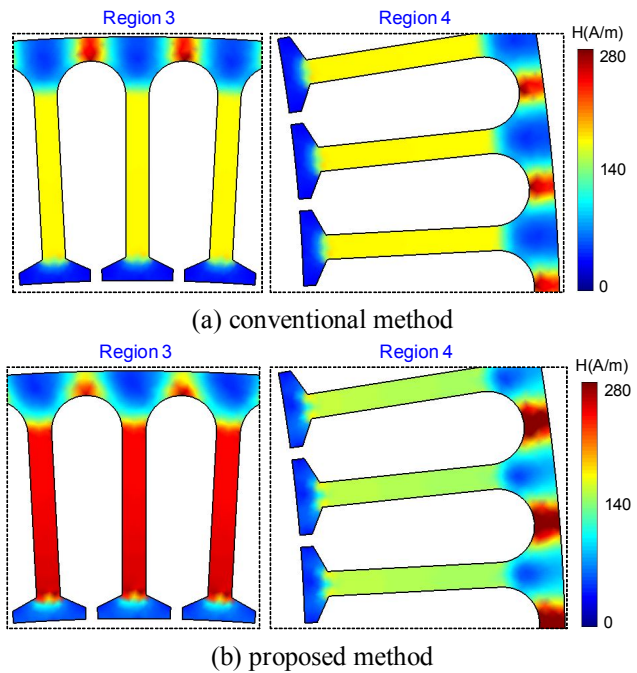


Fig. 7. Maximum magnetic field intensities calculated by the conventional and proposed methods in the stator core

For the total iron loss generated in the stator core, the pro-posed method gives 129% higher iron loss (7.35W) than the conventional one (3.2W). It is thought that this large deviation comes from that of the rotational iron loss in the pole shoe and the alternating iron loss at the stator teeth

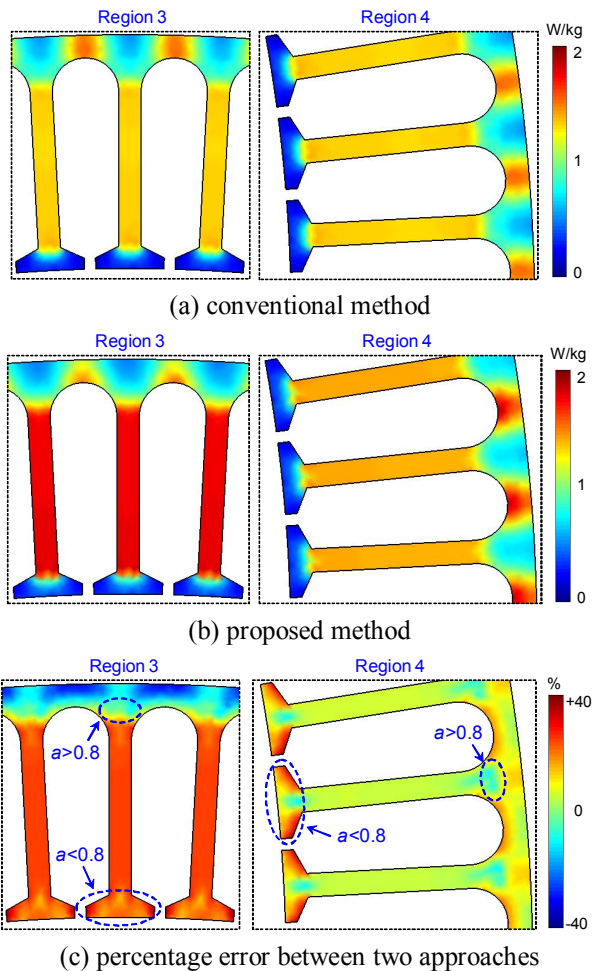


Fig. 8. Iron losses calculated by the conventional and proposed methods in the stator core

along the TD.

5. Conclusion

In this paper, the proposed method based on the E&S vector hysteresis model is presented for the iron loss prediction of the permanent magnet rotating machine. The proposed method gives relatively higher rotational iron loss in the pole shoe and alternating one in the stator teeth along the TD than the conventional method based on the empirical iron loss formula. In consideration of the iron loss characteristics measured by the 2-D SST, it is thought that the proposed method gives more reasonable result for both not only alternating but also rotational iron losses.

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