



double-talk detector, is based on a comparison of the variances of the estimated and measured microphone signals  $\hat{y}(t)$  and  $y(t)$ . In its standard form, the NCR algorithm is computationally infeasible, but fortunately one may form a computationally cheap version of the algorithm by assuming that the adaptive filter has converged to the true echo path. Here, we use the forgetting factor version of the cheap NCR algorithm, forming the decision variable [7].

TABLE I  
SUMMARY OF THE MODIFIED AP ALGORITHM

$\mathbf{e}_n = \mathbf{d}_n - \mathbf{X}_n^t \mathbf{w}_n$	(1)
$\mathbf{w}_{n+1} = \mathbf{w}_n + \mu \mathbf{X}_n [\mathbf{X}_n^t \mathbf{X}_n + P \cdot L \cdot \sigma_{e,n}^2 \mathbf{I}]^{-1} \mathbf{e}_n$	(2)
where $L$ is the adaptive filter length, $P$ is the projection order of modified AP algorithm, and the following definitions are made:	
$x_n$ : excitation signal of $n^{\text{th}}$ instant	
$\mathbf{x}_n = [x_n \ x_{n-1} \ \cdots \ x_{n-L+1}]^t$ : excitation vector	(3)
$\mathbf{X}_n = [\mathbf{x}_n \ \mathbf{x}_{n-1} \ \cdots \ \mathbf{x}_{n-P+1}]$ : $P$ excitation matrix	(4)
$\mathbf{w}_n = [w_{0,n} \ w_{1,n} \ \cdots \ w_{L-1,n}]^t$ : adaptive coefficient vector, $w_{i,n}$ : $i^{\text{th}}$ adaptive coefficient at time $n$	(5)
$y_n$ : measurement noise signal.	
$\mathbf{y}_n = [y_n \ y_{n-1} \ \cdots \ y_{n-P+1}]^t$ : measurement noise vector	(6)
$\mathbf{d}_n = [d_n \ d_{n-1} \ \cdots \ d_{n-P+1}]^t$ : system output vector	(7)
$e_n = d_n - \mathbf{x}_n^t \mathbf{w}_n$ : <i>a priori</i> error signal	(8)
$\mathbf{e}_n = [e_n \ e_{n-1} \ \cdots \ e_{n-P+1}]^t$ : <i>a priori</i> error vector	(9)
$\mu$ : step size parameter	
$\sigma_{e,n}^2 = \beta \sigma_{e,n-1}^2 + (1 - \beta) e_n^2$ : running power estimate of a <i>priori</i> error signal	(10)

$$d_{CN}(k) = \sqrt{\frac{1}{\sigma_y^2} \hat{\mathbf{r}}_{xy}^T(k) \hat{\mathbf{H}}(k)} \quad (11)$$

$$\text{where } \hat{\mathbf{r}}_{xy}^T(k) = \lambda \hat{\mathbf{r}}_{xy}^T(k-1) + (1 - \lambda) x(k) y(k) \quad (12)$$

$$\hat{\sigma}_y^2(k) = \lambda \hat{\sigma}_y^2(k-1) + (1 - \lambda) y^2(k) \quad (13)$$

with  $\lambda$  denoting a forgetting factor. Double-talk is deemed to occur when  $d_{CN}(k)$  is below some predetermined threshold,  $T$ , i.e.,

$$\text{Decision} = \begin{cases} d_{CN} \geq T, & DT \text{ not present.} \\ d_{CN} < T, & DT \text{ present.} \end{cases} \quad (14)$$

### 2) Linear prediction error filter

The residual echo signal is whitened by using linear prediction error filter with  $P^{\text{th}}$  order. During non-double-talk states, estimated error signal  $e(k)$  in adaptive filter includes residual echo  $r(k)$  and background noise  $n(k)$ . Residual echo  $r(k)$  which has speech characteristics removed by whitening process using  $P^{\text{th}}$  order linear prediction error filter, just like (15).

$$r_e(k) = e(k) - \sum_{i=1}^P a_i(k) e(k-i) \quad (15)$$

In equation (15)  $a_i(k)$  and  $r_e(k)$  denote coefficients of linear predictor and error signal from whitened acoustic echo canceller respectively. The coefficient of the linear predictor calculated from the solution of Wiener-Hopf equation, which is earned from Levinson-Durbin algorithm [8].

## III. COMPUTER SIMULATION AND RESULTS

We apply to acoustic echo canceller with the proposed algorithm in hands-free environments. Far-end signal recorded with 8 kHz sampling rate, 16 bits quantization-level, and 10 second-long man and woman alternate pronounced English sentences. For considering double-talk situation, near-end signal recorded with 2 second-long man pronounced the other English sentences. Far-end signal to background noise ratio set to 30 dB and 20dB by assuming low level and high level white-Gaussian background noise, respectively. And acoustic echo path impulse response measured in small size office room with 512<sup>th</sup> order lengths. Adaptive filter has the same length, step-size,  $\mu = 0.125$  and projection order  $P = 2$ . For performance evaluation with proposed algorithm, AIC(acoustic interference cancellation) were used[4].

(2.25)

$$AIC(k) = 10 \log_{10} \frac{E\{y^2(k)\}}{E\{\hat{z}^2(k)\}} \quad (16)$$

$$(2.27) = 10 \log_{10} \frac{E\{y^2(k)\}}{E\{y^2(k) - \hat{i}^2(k)\}} \quad [dB]$$

Where,  $\hat{i}(k) = \hat{d}(k) + \hat{n}(k)$  means estimated interference signal including estimated echo and background noise. Equation (16) means power ratio of microphone input signal  $y(k)$  (including acoustic echo and background noise) and transmitted signal  $\hat{z}(k)$  (or residual error signal). Therefore as acoustical echo is eliminated more by acoustic echo canceller, AIC has larger value.

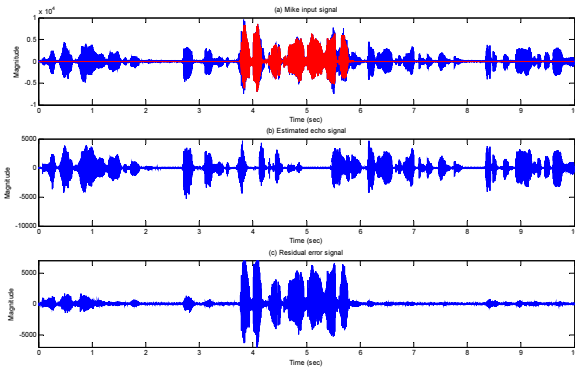


Fig. 2. A comparison of the echoes before and after cancellation (NLMS, 30dB white Gaussian noise).

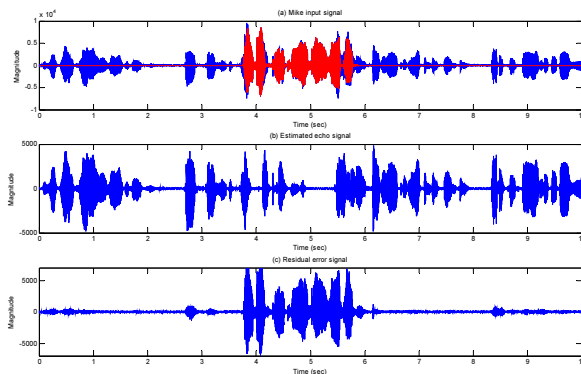


Fig. 3. A comparison of the echoes before and after cancellation (AP (P=2), 30dB white Gaussian noise).

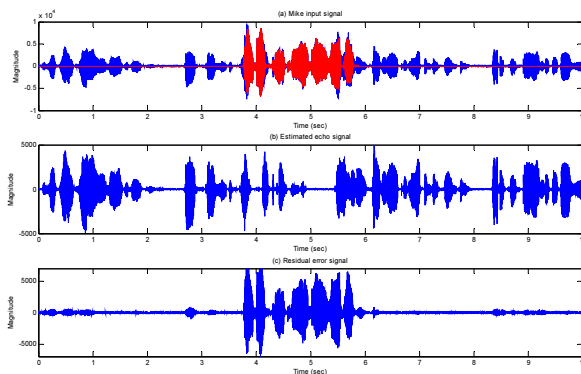


Fig. 4. A comparison of the echoes before and after cancellation (Proposed modified AP (P=2), without Post-processing, 30dB white Gaussian noise).

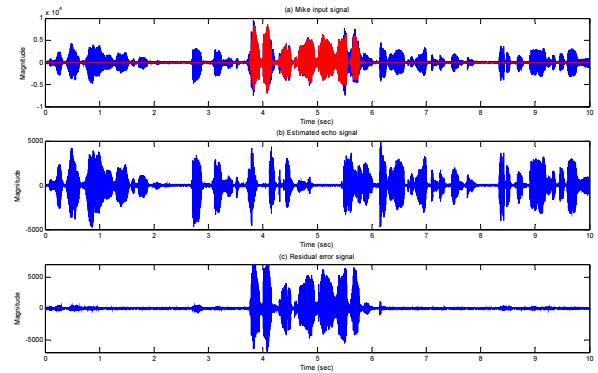


Fig. 5. A comparison of the echoes before and after cancellation (Proposed modified AP (P=2), 30dB white Gaussian noise).

Fig. 2 ~ 6 show the results for relatively high SNR, 30dB including double-talk intervals. In fig. 2 ~ 5 (a), red color signal means near-end signal for double-talk intervals. As shown in Fig. 5 (c), the echo and residual signal well removed by the proposed adaptive filter before and after double-talk intervals.

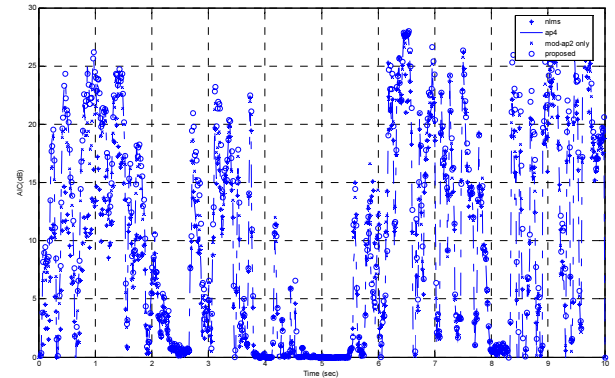


Fig. 6. AIC comparison with 30dB white Gaussian noise (NLMS, AP, proposed modified AP without post-processing, and proposed modified AP).

The results at relatively high SNR, based on AIC with NLMS, AP, proposed modified AP and proposed modified AP without post-processing are shown in Fig. 6. The Proposed algorithm has about 3 dB gain over the existing method at relatively high SNR.

Fig. 7 ~ 11 show the results for relatively low SNR, 20dB including double-talk intervals. In fig. 7 ~ 10 (a), red color signal means near-end signal for double-talk intervals. As shown in Fig. 10 (c), the echo and residual signal well removed by the proposed adaptive filter before and after double-talk intervals at relatively low SNR.

The results at relatively low SNR, based on AIC with NLMS, AP, proposed modified AP and proposed modified AP without post-processing are shown in Fig. 11.

The Proposed algorithm has more than 3 ~ 5dB gain over the existing method at relatively low SNR.

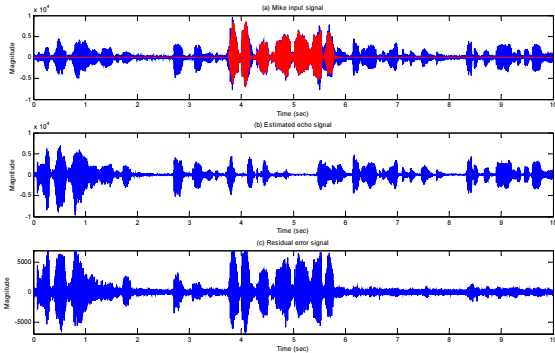


Fig. 7. A comparison of the echoes before and after cancellation (NLMS, 20dB white Gaussian noise).

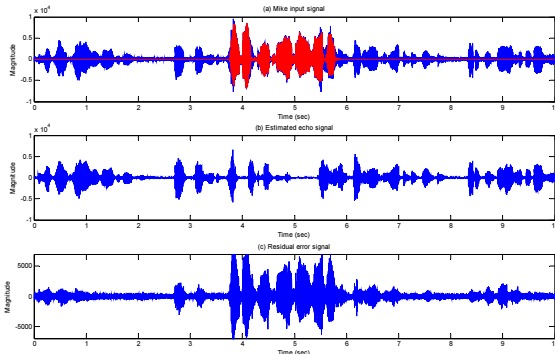


Fig. 8. A comparison of the echoes before and after cancellation (AP (P=2), 20dB white Gaussian noise).

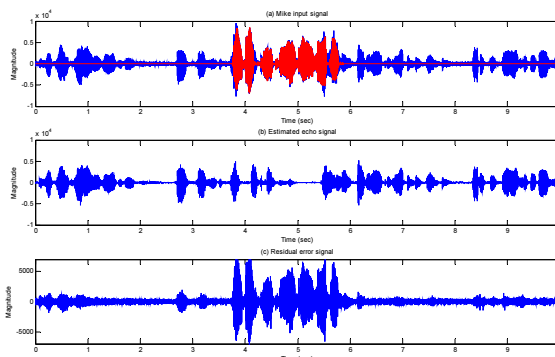


Fig. 9. A comparison of the echoes before and after cancellation (Proposed modified AP (P=2), without Post-processing, 20dB white Gaussian noise).

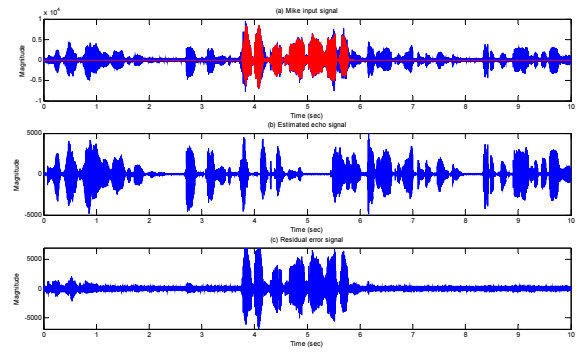


Fig. 10. A comparison of the echoes before and after cancellation (Proposed modified AP (P=2), 20dB white Gaussian noise).

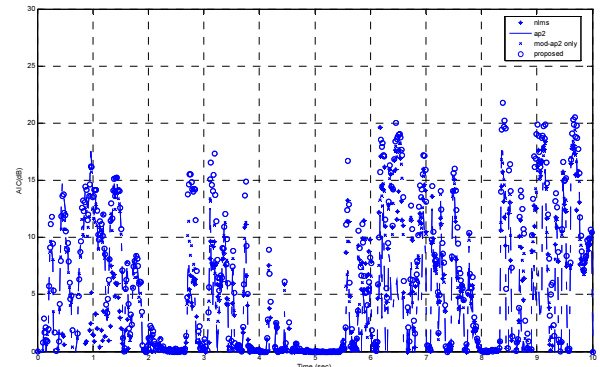


Fig. 11. AIC comparison with 20dB white Gaussian noise (NLMS, AP, proposed modified AP without post-processing, proposed modified AP).

#### IV. CONCLUSIONS

By computer simulation, the proposed algorithm has better performance over the existing method at relatively low SNR. Consequently, the proposed algorithm is more efficient to large background noise and residual echo.

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