

# A Simple Method to Reduce the Splitting Error in the LOD-FDTD Method

Ki-Bok Kong<sup>1</sup> · Myung-Hun Jeong<sup>1</sup> · Hyung Soo Lee<sup>2</sup> · Seong-Ook Park<sup>1</sup>

## Abstract

This paper presents a new iterative locally one-dimensional finite-difference time-domain(LOD-FDTD) method that has a simpler formula than the original iterative LOD-FDTD formula<sup>[1]</sup>. There are fewer arithmetic operations than in the original LOD-FDTD scheme. This leads to a reduction of CPU time compared to the original LOD-FDTD method while the new method exhibits the same numerical accuracy as the iterative ADI-FDTD scheme. The number of arithmetic operations shows that the efficiency of this method has been improved approximately 20 % over the original iterative LOD-FDTD method.

**Key words** : Alternating-Direction-Implicit Finite-Difference Time-Domain(ADI-FDTD), Iterative Method, Locally-One-Dimensional Finite-Difference Time-Domain(LOD-FDTD) Technique, Maxwell's Equations.

## I. Introduction

The finite-difference time-domain(FDTD) method is widely used to solve electromagnetic problems due to its low computational complexity and easy implementation<sup>[2]</sup>. However, certain types of problems require a small grid size, bringing about a large Courant-Friedrichs-Lewy(CFL) number. To overcome the CFL condition, the ADI-FDTD method has been applied to the discrete scheme of Maxwell's equations<sup>[3]</sup>. More recently, the LOD-FDTD scheme has been applied to another unconditionally stable method, which provides simple implementation and reduces the computational time<sup>[4]</sup>. However, the ADI-FDTD and LOD-FDTD schemes generate a large numerical dispersion error due to the truncation term in their two-step factorization<sup>[5]</sup>. The iterative scheme has been used to reduce the splitting error associated with the ADI and LOD-FDTD discrete formulas<sup>[1],[6]</sup>. In [7], the error vector between the Crank-Nicolson FDTD (CN-FDTD) and ADI-FDTD schemes has been applied to obtain a more efficient iterative version of the ADI-FDTD method.

In this paper, we propose a new 2-D iterative LOD-FDTD scheme using the error vector between the CN-FDTD and LOD-FDTD methods. The proposed scheme has a simpler discrete formulation and reduces the computational time while maintaining the same accuracy of the numerical results.

## II. Derivation of New LOD-FDTD Method

The 2-D Maxwell's equations of the TE<sub>z</sub> case can be written as the following matrix system divided into two matrices of  $x$  and  $y$ :

$$\frac{\partial}{\partial t} \vec{u} = \mathbf{A} \vec{u} + \mathbf{B} \vec{u}, \quad (1)$$

where

$$\mathbf{A} = \begin{bmatrix} -\sigma/\varepsilon & 0 & \frac{\partial}{\varepsilon \partial y} \\ 0 & 0 & 0 \\ \frac{\partial}{\mu \partial y} & 0 & 0 \end{bmatrix}, \quad \mathbf{B} = \begin{bmatrix} 0 & 0 & 0 \\ 0 & -\sigma/\varepsilon & -\frac{\partial}{\varepsilon \partial x} \\ 0 & -\frac{\partial}{\mu \partial x} & 0 \end{bmatrix}$$

and  $\vec{u} = [E_x \quad E_y \quad H_z]^T$ .

Replacing the time derivative of equation (1) with its corresponding difference operation generates the following CN-FDTD equation:

$$\left( \mathbf{I} - \frac{\Delta t}{2} \mathbf{A} - \frac{\Delta t}{2} \mathbf{B} \right) \vec{u}^{n+1} = \left( \mathbf{I} + \frac{\Delta t}{2} \mathbf{A} + \frac{\Delta t}{2} \mathbf{B} \right) \vec{u}^n, \quad (2)$$

where  $\mathbf{I}$  denotes an identity matrix. The two sub-step LOD-FDTD procedures are given by

$$\left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{A} \right] \vec{u}_{LOD}^{n+1/2} = \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{A} \right] \vec{u}^n \quad (3a)$$

$$\left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{B} \right] \vec{u}_{LOD}^{n+1} = \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{B} \right] \vec{u}_{LOD}^{n+1/2}, \quad (3b)$$

where  $\vec{u}_{LOD}^{n+1/2}$  is an intermediate solution. To reduce the

splitting error, the iterative LOD-FDTD scheme has been introduced [1], and it has two iterative processing steps with respect to  $k$ :

$$\left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{A} \right] \bar{\mathbf{u}}^{n+1/2} = \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{A} \right] \bar{\mathbf{u}}^n + \frac{\Delta t^2}{4} \mathbf{AB}(\bar{\mathbf{u}}_k^{n+1} + \bar{\mathbf{u}}^{n+1/2}) \quad (4a)$$

$$\left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{B} \right] \bar{\mathbf{u}}_{k+1}^{n+1} = \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{B} \right] \bar{\mathbf{u}}^{n+1/2} - \frac{\Delta t^2}{4} \mathbf{BA}(\bar{\mathbf{u}}^{n+1/2} + \bar{\mathbf{u}}^n). \quad (4b)$$

This method is less costly than the iterative ADI-FDTD method due to the reduced number of arithmetic operations, and the same numerical accuracy is maintained [1]. In this paper, a new iterative LOD-FDTD scheme for 2D Maxwell's equations is introduced and shown to be more efficient than the original iterative LOD-FDTD scheme. For each  $(n+1)$ th time step, we first solve  $\bar{\mathbf{u}}_{LOD}^{n+1}$  from the LOD-FDTD equation, and the equations of (3) can be written equivalently as

$$\begin{aligned} & \left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{A} \right] \left[ \left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{B} \right] \bar{\mathbf{u}}_{LOD}^{n+1} \right] \\ &= \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{B} \right] \left[ \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{A} \right] \bar{\mathbf{u}}^n + \frac{\Delta t^2}{4} (\mathbf{BA} - \mathbf{AB}) \bar{\mathbf{u}}_{LOD}^{n+1/2} \right], \end{aligned} \quad (5)$$

where  $\bar{\mathbf{u}}^n$  is the solution in the previous time step. To investigate the error between the LOD-FDTD and CN-FDTD schemes, we redescribe the linear system CN-FDTD (2), including the LOD-FDTD equation and the temporal second-order spatial derivative terms, giving us the following:

$$\begin{aligned} & \left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{A} \right] \left[ \left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{B} \right] \bar{\mathbf{u}}^{n+1} \right] \\ &= \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{B} \right] \left[ \left[ \mathbf{I} + \frac{\Delta t}{2} \mathbf{A} \right] \bar{\mathbf{u}}^n + \frac{\Delta t^2}{4} \mathbf{AB} \bar{\mathbf{u}}^{n+1/2} - \frac{\Delta t^2}{4} \mathbf{BA} \bar{\mathbf{u}}^n \right]. \end{aligned} \quad (6)$$

Since LOD-FDTD solves (3) instead of (6), a splitting error depending on the truncation error terms occurs. The truncation error is related to the terms of the spatial derivatives with the second-order time step size factor  $\Delta t^2$ , i.e.,  $\frac{\Delta t^2}{4} (\mathbf{BA} - \mathbf{AB}) \bar{\mathbf{u}}_{LOD}^{n+1/2}$ ,  $\frac{\Delta t^2}{4} \mathbf{AB} \bar{\mathbf{u}}^{n+1/2}$ , and  $\frac{\Delta t^2}{4} \mathbf{BA} \bar{\mathbf{u}}^n$ . The iterative scheme efficiently reduces the splitting error in ADI-FDTD and LOD-FDTD<sup>[1],[6]</sup>. To apply the iterative scheme, we first define a new vector  $\bar{\mathbf{e}}^{n+1} = \bar{\mathbf{u}}^{n+1} - \bar{\mathbf{u}}_{LOD}^{n+1}$  that describes the error difference between the LOD-FDTD and CN-FDTD schemes. After subtracting (5) from (6), we obtain the following equation related to  $\bar{\mathbf{e}}^{n+1}$ :

$$\mathbf{M} \bar{\mathbf{e}}^{n+1} = \mathbf{N} \bar{\mathbf{e}}^{n+1} + \bar{\mathbf{b}}^{n+1}, \quad (7)$$

where  $\mathbf{M} = [\mathbf{I} - (\Delta t/2) \mathbf{A}] [\mathbf{I} - (\Delta t/2) \mathbf{B}]$ ,  $\mathbf{N} = (\Delta t^2/4) \mathbf{AB}$ , and  $\bar{\mathbf{b}}^{n+1} =$

$(\Delta t^2/4) \mathbf{AB}(\bar{\mathbf{u}}_{LOD}^{n+1} + \bar{\mathbf{u}}_{LOD}^{n+1/2}) - (\Delta t^2/4) \mathbf{BA}(\bar{\mathbf{u}}_{LOD}^{n+1/2} + \bar{\mathbf{u}}^n)$ . Here,  $\mathbf{M} - \mathbf{N}$  is a splitting of the matrix  $(\mathbf{I} - (\Delta t/2) \mathbf{A} - (\Delta t/2) \mathbf{B})$ . To apply the iterative scheme to error vector equation (7), we highlight the iterative nature of equation (7):

$$\mathbf{M} \bar{\mathbf{e}}_{k+1}^{n+1} = \mathbf{N} \bar{\mathbf{e}}_k^{n+1} + \bar{\mathbf{b}}^{n+1}, \quad (8)$$

where the subscript  $k$  denotes the  $k$ th iterative solution<sup>[7]</sup>. If  $\mathbf{M}$  is nonsingular and the spectral radius of  $\mathbf{M}^{-1} \mathbf{N}$  is less than one, then iterate  $\bar{\mathbf{e}}_k^{n+1}$  converges into  $\bar{\mathbf{e}}^{n+1}$  for any initial guess<sup>[8]</sup>. The convergence of iterative equation (8) can be proved by a method similar to that of [7]. The solution  $\bar{\mathbf{u}}^{n+1}$  for each time step  $n+1$  is given by  $\bar{\mathbf{u}}^{n+1} \approx \bar{\mathbf{e}}_{final\ k}^{n+1} + \bar{\mathbf{u}}_{LOD}^{n+1}$  after the iterative execution of (8). If the LOD concept is applied to implement this iterative solver, the system in (8) can be divided into two steps,

$$\left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{A} \right] \bar{\mathbf{e}}^{tmp} = \frac{\Delta t^2}{4} \mathbf{AB} \bar{\mathbf{e}}_k^{n+1} + \frac{\Delta t^2}{4} \mathbf{AB}(\bar{\mathbf{u}}_{LOD}^{n+1} + \bar{\mathbf{u}}_{LOD}^{n+1/2}) \quad (9a)$$

$$\left[ \mathbf{I} - \frac{\Delta t}{2} \mathbf{B} \right] \bar{\mathbf{e}}_{k+1}^{n+1} = \bar{\mathbf{e}}^{tmp} - \frac{\Delta t^2}{4} \mathbf{BA}(\bar{\mathbf{u}}_{LOD}^{n+1/2} + \bar{\mathbf{u}}^n), \quad (9b)$$

where  $\bar{\mathbf{e}}^{tmp} = (e_x^{tmp}, e_y^{tmp}, e_z^{tmp})$  denotes an intermediate solution. Equations (9a) and (9b) resemble the original iterative scheme because they contain the common spatial second-order terms in each time sub-step. To recover (8) from the two-step implementation in (9), we use the matrix multiplication of  $\mathbf{ABA} = \mathbf{0}$  in the 2D TE<sub>z</sub> case.

In the first sub-step  $(n+1/2)$ , (9a) can be expanded into the following:

$$\begin{aligned} \left( 1 + \frac{\sigma \Delta t}{2\epsilon} \right) e_x^{tmp} &= \frac{\Delta t}{2\epsilon} \frac{\partial}{\partial y} e_y^{tmp} + \frac{\Delta t^2}{4\epsilon\mu} \frac{\partial^2}{\partial x \partial y} e_{y,k}^{n+1} \\ &\quad - \frac{\Delta t^2}{4\epsilon\mu} \frac{\partial^2}{\partial x \partial y} (E_{y,LOD}^{n+1} + E_{y,LOD}^{n+1/2}) \end{aligned} \quad (10a)$$

$$e_y^{tmp} = 0 \quad (10b)$$

$$e_z^{tmp} = \frac{\Delta t}{2\mu} \frac{\partial}{\partial y} e_x^{tmp}, \quad (10c)$$

and in the second sub-step  $(n+1)$ , we obtain the following equations from (9b):

$$e_{x,k+1}^{n+1} = e_x^{tmp} \quad (11a)$$

$$\begin{aligned} \left( 1 + \frac{\sigma \Delta t}{2\epsilon} \right) e_{y,k+1}^{n+1} &= e_y^{tmp} - \frac{\Delta t}{2\epsilon} \frac{\partial}{\partial x} e_{z,k+1}^{n+1} \\ &\quad + \frac{\Delta t^2}{4\epsilon\mu} \frac{\partial^2}{\partial x \partial y} (E_{x,LOD}^{n+1/2} + E_x^n) \end{aligned} \quad (11b)$$

$$e_{z,k+1}^{n+1} = e_z^{tmp} - \frac{\Delta t}{2\mu} \frac{\partial}{\partial x} e_{y,k+1}^{n+1}. \quad (11c)$$

In each iterative sub-step, the performance of solving the equation requires one implicit tridiagonal matrix solver and one explicit update.

The whole procedure of this scheme is illustrated in Fig. 1. Note that  $e_y^{tmp}$  is always zero and  $\vec{e}^{n+1}$  is zero on the boundary of a domain, which simplifies the tridiagonal matrix in comparison to the original iterative scheme.

Moreover, Table 1 illustrates that the number of arithmetic operations on the right side of each sub-step equation is less than the original iterative LOD-FDTD, leading to a reduction of CPU time for computational executions.

### III. Numerical Implementations

As a comparative study of numerical error over the ADI-FDTD method, two 2-m-long parallel conducting plates with a separation distance of 0.02 m in free space surrounded by a perfect magnetic conductor (PMC) are investigated. The grid sizes are  $\Delta h = \Delta x = \Delta y = 0.02$  m in the x and y directions. The numerical experiment is conducted with a 750-kHz raised cosine, which held constant after reaching its maximum of 1 V. In the region between the plates ( $4 \text{ m} \leq x \leq 6 \text{ m}$ ), the field am-

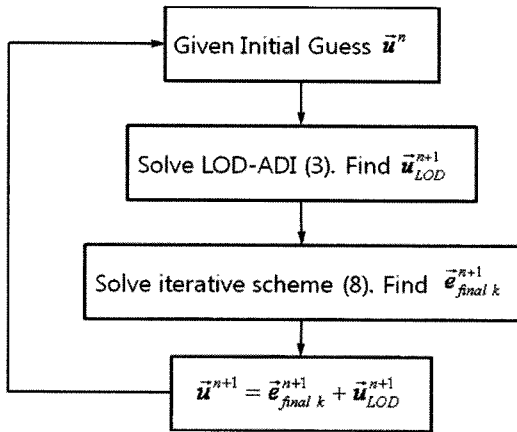


Fig. 1. The proposed iterative method.

Table 1. Number of arithmetic operations.

Method	Implicit		Explicit		Total	
	M/D	A/S	M/D	A/S	M/D	A/S
Original LOD	4	12	2	8	6	20
New LOD	2	14	2	3	4	17

M/D: multiplication/division, A/S: addition/subtraction.

plitude is constant. Unlike the original iterative LOD-FDTD, execution of the iterative process will be conducted in a different way because of the convergent speed of iterative equation (9). If the iteration number is small, then solution  $\vec{e}_k^{n+1}$  of (9) is large, leading to a divergence of  $\vec{u}^{n+1}$ . Thus, the iteration number needs to be sufficiently large. Since good agreement between the iterative LOD-FDTD and iterative ADI-FDTD has been previously observed and described by K. Jung<sup>[1]</sup>, we investigate the agreement between the proposed iterative LOD-FDTD and iterative ADI-FDTD.

The  $L^2$  norm  $\|\eta_k^n\|$  of  $e_{y,k}^n$  for each time step  $n$  and iterative step  $k$  is defined as

$$\|\eta_k^n\| = \sqrt{\sum_x |e_{y,k}^n - e_{y,k-1}^n|^2}, \quad (12)$$

where  $e_{y,0}^n$  is an initial guess. Since the iterative equation of (7) converges, the norm  $\|\eta_k^n\|$  goes to zero as  $k \rightarrow \infty$ <sup>[8]</sup>. We use  $\|\eta_k^n\|$  to determine the iteration number for convergence of the iterative scheme. The iterative step is executed until the norm  $\|\eta_k^n\|$  arrives at a sufficiently small value. The numerical implementation of (9) in Fortran can be simply expressed as follows:

```

do while ( $\|\eta_k^n\|$ .gt. LIMIT(n))
    execute (9)
    calculate  $\|\eta_k^n\|$ 
    k=k+1
end while.

```

Here  $LIMIT(n)$  is a lower limit for interrupting the *while* loop, and is related to the previous limit value  $\|\eta_{final}^{n-1}\|$  because of the stability of the iterative scheme. Thus, a numerical example of this chapter takes  $LIMIT(n)$  as a value similar to  $\|\eta_{final}^{n-1}\|$ . The total iteration number is obtained by summing every final  $k$  for each time step  $n$ . The iteration number of iterative ADI-FDTD is calculated by dividing the total iteration number of the new iterative LOD-FDTD by the final  $n$ .

The CFL number is given as  $s = c_0 \Delta t / \Delta h$  where  $c_0$  is the speed of light in a vacuum. The  $L_2$  relative error norm for  $E_y$  along the x-axis is defined as follows:

$$\sqrt{\frac{\sum_x |E_y - E_y^{ref}|^2}{\sum_x |E_y^{ref}|^2}} \times 100, \quad (13)$$

where  $E_y$  is a measured field at  $y=2$  where the upper plate is located, and  $E_y^{ref}$  is a reference field with a value of 1 at the feeding plate and 0 at the other area.

Fig. 2 shows the  $E_y$  field along the x-axis by the new iterative LOD-FDTD and iterative ADI-FDTD method with the same total iteration number. This illustrates that the new iterative LOD-FDTD reduces the splitting error in a manner similar to the original LOD-FDTD case.

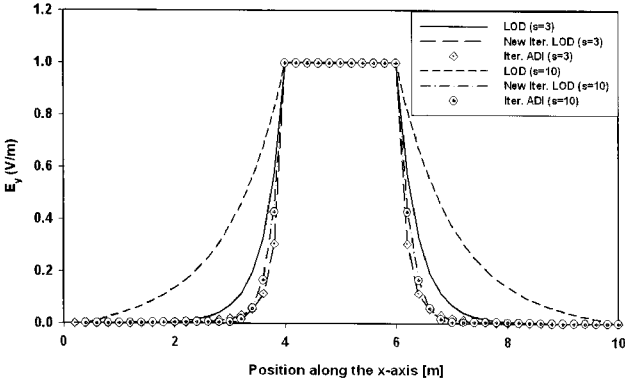


Fig. 2.  $E_y$  along the x-axis calculated by LOD-FDTD, the proposed iterative LOD-FDTD method, and iterative ADI-FDTD method with the same total iteration number.

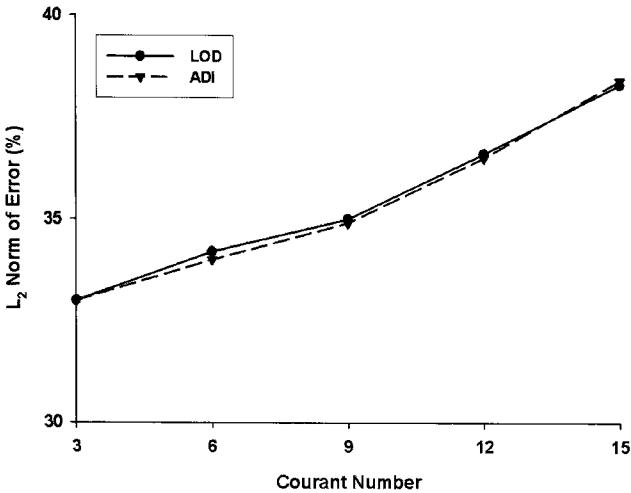


Fig. 3.  $L_2$  norm of the error of the new iterative LOD-FDTD method and the iterative ADI-FDTD method for different courant number  $s$ .

Fig. 3 shows that the  $L_2$  relative error norm for  $E_y$  along the x-axis of the new iterative LOD-FDTD method agrees with the iterative ADI-FDTD method with the same total iteration number.

#### IV. Conclusions

A new iterative LOD-FDTD method has been introduced using an error vector between the CN-FDTD and LOD-FDTD. A numerical example shows that the new iterative LOD-FDTD scheme efficiently reduces the splitting error in a manner similar to that of the original iterative LOD-FDTD scheme. The new iterative LOD-FDTD scheme uses a simpler numerical discrete formula compared with the original iterative LOD-FDTD, and due to the reduced number of arithmetic operations, the computational time is lower than that of the original iterative LOD-FDTD.

This work was supported by the IT R&D program of MKE/IITA, Korea, under contract No. 2008-F-050-01.

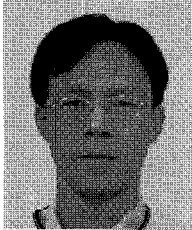
#### References

- [1] K. Y. Jung, F. L. Teixeira, "An iterative unconditionally stable LOD-FDTD method", *IEEE Microwave and Wireless Components Lett.*, vol. 18, no. 2, pp. 76-78, Feb. 2008.
- [2] Kane Yee, "Numerical solution of initial boundary value problems involving Maxwell's equations in isotropic media", *IEEE Trans. Ant. Propag.*, vol. 14, Issue 3, pp. 302-307, May 1966.
- [3] T. Namiki, "A new FDTD algorithm based on alternating-direction implicit method", *IEEE Trans. Microwave Theory Tech.*, vol. 47, no. 10, pp. 2003-2007, Oct. 1999.
- [4] J. Shibayama, M. Muraki, J. Yamauchi, and H. Nakno, "Efficient implicit FDTD algorithm based on locally one-dimensional scheme", *Electron. Lett.*, vol. 41, no. 19, pp. 1046-1047, Sep. 2005.
- [5] S. G. Garcia, T. W. Lee, and S. C. Hagness, "On the accuracy of the ADI-FDTD method", *IEEE Antennas and Wireless Propagation Lett.*, vol. 1, pp. 31-34, 2002.
- [6] S. Wang, F. L. Teixeira, and J. Chen, "An iterative ADI-FDTD with reduced splitting error", *IEEE Microwave and Wireless Components Lett.*, vol. 15, no. 2, pp. 92-94, Feb. 2005.
- [7] K. Kong, J. Kim, and S. Park, "Reduced splitting error in the ADI-FDTD method using iterative me-

thod", *Microw. Opt. Tech. Lett.*, vol. 50, no. 8, pp. 2200-2203, Aug. 2008.

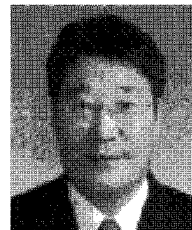
[8] G. H. Golub, C. F. Van Loan, *Matrix Computations*, 3<sup>rd</sup> ed., The Johns Hopkins Univ. Press, 1996.

#### Ki-Bok Kong



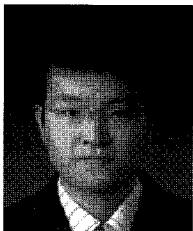
received the B.S. degree from Kyung-Pook National University, in 1992, and the M.S and Ph.D. degree from Korea Advanced Institute of Science and Technology, Daejeon, in 1999 and 2003, respectively, all in applied mathematics. Since 2004, he has been a research professor with IT Convergence Campus of Korea Advanced Institute of Science and Technology, Daejeon, Korea. His research interests are the partial differential equations and the computational electromagnetics using finite difference time domain method.

#### Hyung Soo Lee



received the B.E. degree in electronics from Kyungpook National University, Taegu, Korea, in 1980, the M.S. degree in electronic computation engineering from the Yonsei University, Seoul, Korea, in 1986, and the Ph.D. degree in information and communication engineering from Sungkyunkwan University, Seoul, Korea, in 1996. From 1983, he is a Research Engineer with Electronics and Telecommunications Research Institute, Daejeon, Korea. His main research interest is RF technology for WBAN, WPAN applications.

#### Myung-Hun Jeong



received the B.E. degree in electronics from Chungnam National University, Daejeon, Korea, in 2008 and is currently working toward the M.S. degree in electrical engineering from IT Convergence Campus of Korea Advanced Institute of Science and Technology, Daejeon, Korea. His main research interest is RF technology for wireless communications and biomedical applications.

#### Seong-Ook Park



received the B.S. degree from Kyung-Pook National University, in 1987, the M.S. degree from Korea Advanced Institute of Science and Technology, Seoul, in 1989, and the Ph.D. degree from Arizona State University, Tempe, in 1997, all in electrical engineering. From March 1989 to August 1993, he was a Research Engineer with Korea Telecom, Daejeon, working with microwave systems and networks. He later joined the Telecommunication Research Center, Arizona State University, until September 1997. Since October 1997, he has been with IT Convergence Campus of Korea Advanced Institute of Science and Technology, Daejeon, as an Professor. His research interests include mobile handset antenna, and analytical and numerical techniques in the area of electromagnetics.