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# Wireless MIMO를 위한 Confluent Hypergeometric Function를 이용한 Closed Form Channel Capacity의 유도

S. M. Humayun Kabir\* · Van-Su Pham\* · Linh Mai\* · 윤기완\*

Derivation of Closed Form Channel Capacity Using Confluent Hypergeometric  
Function for Wireless MIMO

S. M. Humayun Kabir\* · Van-Su Pham\* · Linh Mai\* · Giwan Yoon\*

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## 요 약

MIMO기술은 대역폭과 전력에 의해 제한을 받는 무선네트워크의 데이터 전송률을 증대시켜주는 효과적인 기술이다. 송수신단 사이의 데이터 전송률은 채널용량에 의해 결정되며, MIMO기술은 이 채널용량을 적절히 이용함으로써 무선채널에서의 통신신뢰도를 높이는 이점을 가지고 있다. 본 논문은 Confluent Hypergeometric Function를 이용하여 새로운 공식인 closed form capacity formula를 유도한다.

## ABSTRACT

Multiple-input multiple-output (MIMO) is an efficient technology to increase data rate in wireless networks due to bandwidth and power limitations. Data transmission rate between transmitter and receiver is determined by channel capacity. MIMO has an advantage of reliable communication over wireless channel because of utilizing the channel capacity properly. In this letter, we drive a new formula, closed form capacity formula, using confluent hypergeometric function.

## 키워드

MIMO systems, channel capacity

## I. INTRODUCTION

Demand for capacity in wireless communications have been rapidly increasing worldwide, and as a solution to

fulfill such a demand, multiple transmitting and receiving antennas namely, multiple-input multiple-output (MIMO) systems have been introduced. The MIMO systems has replaced the single-input single-output (SISO) systems

mainly because the multiple antennas at the transmitter and the receiver can increase data rate and improve quality of service through diversity. Channel capacity is used to determine the consistent information transmission. The standard formula for the Shannon channel capacity is expressed in bps/Hz [1]

$$C = \log_2(1 + \rho |H|^2) \quad (1)$$

where,  $|H|^2$ , denotes the normalized channel power transfer characteristic. To achieve higher data rate research has been done based on this formula. The unequal number of antennas at both the transmitter and receiver increase the capacity logarithmically whereas the equal number of antennas increases the capacity linearly for any fixed signal-to-noise ratio [1] [2]. Multi-element array can effectively comprehend a MIMO wireless channel. Considering the performance of the two situations, the capacity of single user multi-element arrays systems can be explored as follows.

Firstly, the channel is known to the transmitter, thus the power allocation is the finest; secondly, the channel is totally unknown to the transmitter. So, equal power is allocated for each transmitting antenna [3].

In this letter using the confluent hypergeometric function we have introduced an analytical formula for the channel capacity of MIMO wireless systems. In section II we have describe the MIMO system model. In section III proposed channel capacity formula has derived, followed by simulation result. Finally, section IV concludes the paper.

## II. SYSTEM MODEL

We consider a single point-to-point MIMO system where the number of transmit antennas are expressed as  $n_t$  and the number of receive antennas as  $n_r$ . We also assume that  $n = \max(n_t, n_r)$  and  $m = \min(n_t, n_r)$ . The transmitted signals in each symbol period are represented by an  $n_t \times 1$  vectors  $\mathbf{s}$ , whose  $i^{\text{th}}$  entry,  $s_i$ , will be transmitted from antenna  $i$ . We consider the total transmitted power over a

symbol period to  $P$ , not considering the number of transmit antennas  $n_t$ . It is assumed that the covariance matrix of  $\mathbf{s}$ ,  $R_{ss} = E\{\mathbf{s}\mathbf{s}^H\}$ , satisfies the criterion of

$$\text{Tr}(R_{ss}) \leq P \quad (2)$$

where  $\text{Tr}(\cdot)$  denotes the trace of a matrix.

Under the assumption that the channel is frequency-flat fading, the input-output relation over a symbol period is given by

$$\mathbf{r} = \mathbf{H}\mathbf{s} + \mathbf{w} \quad (3)$$

where,

- $\mathbf{r}$  is the  $n_r \times 1$  receive vector
- $\mathbf{H}$  is the  $n_r \times n_t$  channel matrix whose entries are considered the corresponding path gain between the transmit and receive antennas.
- $\mathbf{w}$  represents the additive noise at  $n_r$  receive antennas.

The transmitter completely has no knowledge of the channel but the receiver has perfect knowledge of the channel and the transmit power from each antenna is  $P/n_t$ . The overall channel capacity can be determined by using Shannon capacity formula [4]

$$C = W \sum_{i=1}^r \log_2 \left( 1 + \frac{P_{ri}}{\sigma^2} \right) \quad (4)$$

where  $\sigma^2$  is the noise power,  $W$  is the bandwidth of each sub-channel and  $P_{ri}$  is the received signal power in the  $i^{\text{th}}$  sub-channel. It is given by

$$P_{ri} = \frac{\lambda_i P}{n_t} \quad (5)$$

where  $\sqrt{\lambda_i}$  is the singular value of the channel matrix  $\mathbf{H}$ . Thus the channel capacity can be written as [4]

$$C = W \log_2 \det \left( \mathbf{I}_m + \frac{P}{n_t \sigma^2} \mathbf{Q} \right) \quad (6)$$

when the transmitted signal vector  $\mathbf{s}$  is circularly symmetric complex Gaussian with zero-mean and

covariance  $(P/n_r)I_{n_r}$  and then the capacity is given by [2]

$$C = \mathbb{E}[\log_2 \det \left( I_{n_r} + \frac{P}{n_r \sigma^2} \mathbf{Q} \right)] \quad (7)$$

where the random matrix  $\mathbf{Q}$  is the Wishart matrix defined as

$$\mathbf{Q} = \begin{cases} \mathbf{H}\mathbf{H}^H, & n_r < n_t \\ \mathbf{H}^H\mathbf{H}, & n_r \geq n_t \end{cases} \quad (8)$$

Again in [2], the capacity of the channel with  $n_t$  transmitters and  $n_r$  receivers under power constraint  $P$  is expressed as

$$C = \int_0^\infty \log \left( 1 + \frac{P\lambda}{n_r} \right) \sum_{k=0}^{m-1} \frac{k!}{(k+n-m)!} \times [L_k^{n-m}(\lambda)]^2 \lambda^{n-m} e^{-\lambda} d\lambda \quad (9)$$

where  $L_k^l$  is the associated Laguerre polynomials defined as [5]

$$L_a^b(x) = \frac{e^x x^{-b} d^a (e^{-x} x^{a+b})}{a! dx^a} \quad (10)$$

Also in [5], the connection between Laguerre polynomials and confluent hypergeometric function of the first kind is given by

$$L_a^b(x) = \frac{\Gamma(a+b+1)}{a! \Gamma(b+1)} {}_1F_1(-a; b+1; x) \quad (11)$$

where the hypergeometric function has a hypergeometric series given by

$$\begin{aligned} {}_1F_1(a; b; z) &= 1 + \frac{a}{b}z + \frac{a(a+1)}{b(b+1)} \frac{z^2}{2} + \dots \\ &= \sum_{k=0}^{\infty} \frac{(a)_k}{(b)_k} \frac{z^k}{k!} \end{aligned} \quad (12)$$

where  $(a)_k$  and  $(b)_k$  are Pochhammer symbols. If  $a$  and  $b$  are integers,  $a < 0$ , and either  $b > 0$  or  $b < a$ , then the series yields a polynomial with a finite number of terms, i.e. it turns to a closed form.

### III. DERIVED FORMULA FOR CAPACITY

We derive the formula for the channel capacity. The Laguerre polynomial is defined as [6]

$$\begin{aligned} [L_a^b(x)]^2 &= \frac{\Gamma(a+b+1)}{2^{2a} a!} \sum_{k=0}^a \binom{2a-2k}{a-k} \frac{(2k)!}{k!} \\ &\times \frac{1}{\Gamma(b+k+1)} L_{2k}^{2b}(2x) \end{aligned} \quad (13)$$

where  $\Gamma(\cdot)$  is the gamma function.

According to (13), the Laguerre polynomial of (9) can be expressed as

$$\begin{aligned} [L_k^{n-m}(\lambda)]^2 &= \frac{\Gamma(k+n-m+1)}{2^{2k} k!} \sum_{l=0}^k \binom{2k-2l}{k-l} \frac{(2l)!}{l!} \\ &\times \frac{1}{\Gamma(n-m+l+1)} L_{2l}^{2n-2m}(2\lambda) \end{aligned} \quad (14)$$

Expressing  $L_{2l}^{2n-2m}(2\lambda)$  in terms of (11), we get

$$\begin{aligned} L_{2l}^{2n-2m}(2\lambda) &= \frac{\Gamma(2n-2m+2l+1)}{(2l)! \Gamma(2n-2m+1)} \\ &\times {}_1F_1(-2l; 2n-2m+1; 2\lambda) \end{aligned} \quad (15)$$

According to (12), the hypergeometric function  ${}_1F_1(-2l; 2n-2m+1; 2\lambda)$  will be

$$\begin{aligned} &{}_1F_1(-2l; 2n-2m+1; 2\lambda) \\ &= \sum_{f=0}^{\infty} \frac{(-2l)_f}{(2n-2m+1)_f} \frac{(2\lambda)^f}{f!} \end{aligned} \quad (16)$$

Replacing (16) into (15), we get

$$\begin{aligned} L_{2l}^{2n-2m}(2\lambda) &= \frac{\Gamma(2n-2m+2l+1)}{(2l)! \Gamma(2n-2m+1)} \\ &\times \sum_{f=0}^{\infty} \frac{(-2l)_f}{(2n-2m+1)_f} \frac{(2\lambda)^f}{f!} \end{aligned} \quad (17)$$

Substituting (17) into (14), we get

$$\begin{aligned}
 [L_k^{n-m}(\lambda)]^2 &= \frac{\Gamma(k+n-m+1)}{2^{2k}k!} \sum_{l=0}^k \sum_{f=0}^{\infty} \binom{2k-2l}{k-l} \\
 &\times \frac{1}{\Gamma(n-m+l+1)} \frac{\Gamma(2n-2m+2l+1)}{l!\Gamma(2n-2m+1)} \\
 &\times \frac{(-2l)_f}{(2m-2l+1)_f} \frac{(2\lambda)^f}{f!} \quad (18)
 \end{aligned}$$

Considering the relation  $\Gamma(n) = (n-1)!$ , and substituting (18) into (9) we get the associated channel capacity that can be expressed as

$$\begin{aligned}
 C &= \log_2(e) \sum_{k=0}^{m-1} \sum_{l=0}^k \sum_{f=0}^{\infty} \binom{2k-2l}{k-l} \\
 &\times \frac{1}{(n-m+l)!} \frac{(2n-2m+2l)!}{(2n-2m)!!!} \frac{(-2l)_f}{(2m-2l+1)_f} \\
 &\times \frac{2^f}{2^{2k}f!} \int_0^{\infty} \ln\left(1 + \frac{P\lambda}{n_r}\right) \lambda^{n-m+f} e^{-\lambda} d\lambda \quad (19)
 \end{aligned}$$

For computing the integral of the above equation, we use the result of appendix B of [5]

$$\begin{aligned}
 I_m(\alpha) &= \int_0^{\infty} \ln(1+y) x^{m-1} e^{-\alpha y} dy, \quad \alpha > 0, 1, 2, \dots \\
 &= (m-1)! e^{\alpha} \sum_{j=1}^m \frac{\Gamma(-m+j, \alpha)}{\alpha^j} \quad (20)
 \end{aligned}$$

where the complementary incomplete gamma function is defined in [6] as

$$\Gamma(\alpha, z) = \int_z^{\infty} e^{-y} y^{\alpha-1} dy \quad (21)$$

Computing the integral of (19), according to (20)

$$\begin{aligned}
 I &= (n-m+f)! \left(\frac{n_r}{P}\right)^{n-m+f+1} e^{\left(\frac{n_r}{P}\right)} \\
 &\times \sum_{i=1}^{n-m+f+1} \frac{\Gamma(-n+m-f-1+i, n_r/P)}{(n_r/P)^{-i}} \quad (22)
 \end{aligned}$$

In [7], the relation between the exponential integral of order  $n$  and the complementary incomplete gamma function is defined as

$$E_p(x) = x^{p-1} \Gamma(1-p, x) \quad (23)$$

By employing (23) to (22), we get

$$I = (n-m+f)! e^{(n_r/P)} \sum_{i=0}^{n-m+f} E_{i+1}(n_r/P) \quad (24)$$

Inserting (24) into (19), and also applying the condition of (12), we get the following equation:

$$\begin{aligned}
 C &= e^{(n_r/P)} \log_2(e) \sum_{k=0}^{m-1} \sum_{l=0}^k \sum_{f=0}^{\infty} \binom{2k-2l}{k-l} \\
 &\times \frac{(n-m+f)!}{(n-m+l)!} \frac{(2n-2m+2l)!}{(2n-2m)!!!} \frac{(-2l)_f}{(2m-2l+1)_f} \\
 &\times \frac{1}{2^{2k}} \frac{2^f}{f!} \sum_{i=0}^{n-m+f} E_{i+1}(n_r/P) \quad (25)
 \end{aligned}$$

Figure 1 shows simulation result of the derived capacity. Simulation is done for the different number of antennas in both transmitter and receiver side. Simulation result shows that the capacity increased logarithmically because of unequal number of antennas are deployed in both the transmitter and receiver.

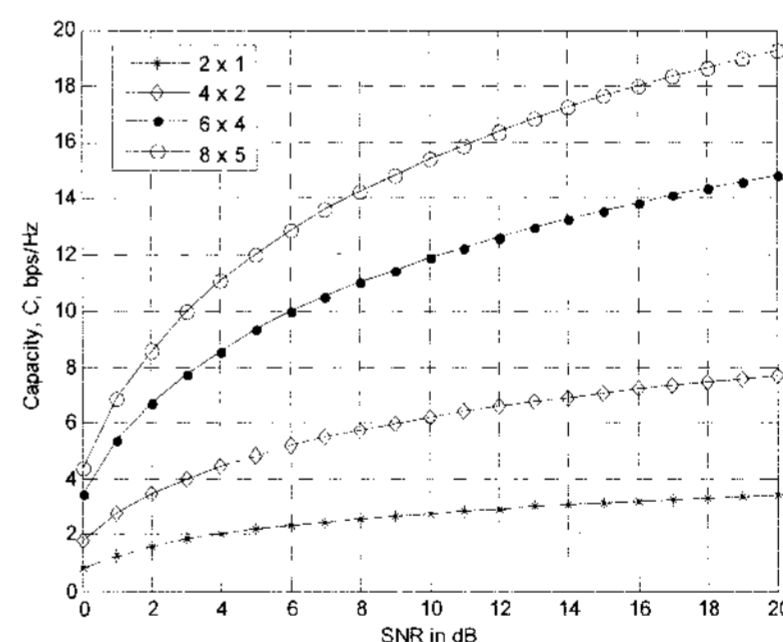


그림 1. 평균SNR에 대한 분석채널용량 (사용부호는 송수신 안테나의 숫자를 나타낸다.)  
 Fig. 1. Analytical channel capacity versus average signal to noise ratio. The legend denotes the number of transmit and receive antennas.

#### IV. CONCLUSION

In this paper, a closed form channel capacity formula has been derived using confluent hypergeometric function.

Simulation was performed to get analytical channel capacity versus average signal to noise ratio. The derived capacity formula allows us to calculate the channel capacity in a closed form. Consequently, the derived formula can be applied for calculating the channel capacity of practical MIMO system where large numbers of antennas are used to transmit and receive data through wireless media.

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저자소개



S. M. Humayun Kabir

2007 Member KIMICS. Received B.S. degree in Computer Science and Engineering, Khulna University, Bangladesh

2007~ M.S. student in Communication Electronics Lab, ICU, Korea.

※ 관심분야 : intelligent algorithms, space-time coding and MIMO systems.



Linh Mai

1996 Member KIMICS Received B. S. degree in Natural Science, Hanoi University, Vietnam

1998 M.S. Degree in ITIMS, Hanoi, Vietnam

1998~2001 Teacher of Posts and Telecommunications Institute of Technology, Vietnam.

2001~ Ph.D. student in Communication and Electronics Lab, ICU, Korea.

※ 관심분야 : rf device



Van-Su Pham

1999 Member KIMICS. Received B.S. degree in Electronic Engineering, Hanoi University, Vietnam

2003 Received M.S. degree in Electrical Engineering, ICU, Korea

2004~ Ph.D. student in Communication Electronics Lab, ICU, Korea.

※ 관심분야 : intelligent algorithms, space-time coding and MIMO systems.



윤기완(Giwan Yoon)

1983 Member KIMICS. Received B.S. degree in Seoul National University

1985 M.S. degree in KAIST both Seoul, Korea.

1985~1990 LG Group, Seoul, Korea

1994 Received Ph.D. degree in The University of Texas at Austin, Texas, USA. .

1994~1997 Digital equipment corporation, MA, USA.

1997~ Professor of school of engineering, ICU

※ 관심분야 : multifunctional intelligent devices & their technologies for RF and wireless applications.