

정확한 모드해석방법을 이용한 차동모드 급전을 가지는 PCB 전력공급회로에서의 슬릿에 의한 전기잡음의 영향 연구

論 文
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On the Effects of Electric Noise due to the Slits in the PCB Power-Distribution Network with the Differential-Mode Signaling using a Rigorous Modal Analysis Method

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Abstract - This study investigates the performances of the signaling techniques including differential signals for the power-distribution network(PDN)s with and without the slit, using a rigorous evaluation method 'Modal Analysis', validated by being compared to the FDTD simulation up to 5 GHz.

Key Words : EMC, Multiple feeding in Power Distribution Network, Full-wave analysis

1. Introduction

The power- and ground planes in PCBs are called the power-distribution network(PDN) and are known for causing cavity-mode resonance and possibly noise in the related system. T. Okoshi uses a modal sum expressions to characterize the structure[1]. Expanding the circuit concept, M. Hampe et al examine the effect of loads like DeCaps on the power-bus resonance[2]. Lately, S. Kahng presented the performance of differential signaling in the PDN and the advantage in reducing the number of resonance frequencies and impedance level[3]. However, it used the feeding scheme for a regular geometry, and a question can be raised if the differential signaling will work well in the power-bus with a geometrical change like having slits shown in Z. Wang's work[4]. This study investigates the performances of the differential signal feeding between the power-bus with and without the slit, using a rigorous evaluation method, which is validated by the FDTD application of [3].

2. Theory

The slit power-bus structure can be modeled as a cavity having the PEC power- and ground planes and the PMC walls.

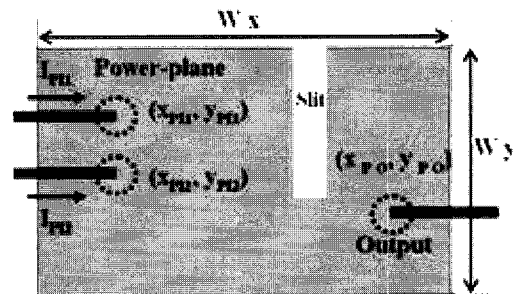


Fig. 1 Top view of a slit PDN

Fig. 1 is the top-view of the PDN structure where 2 feeds provide currents I_{P11} and I_{P12} , passing the structure through the holes at (X_{P11}, Y_{P11}) and (X_{P12}, Y_{P12}) . The output port is placed at (X_{PO}, Y_{PO}) . Excluding the slit, the size of the power-bus is $W_x * W_y * W_z$. The sandwiched substrate is featured by W_z , 4.2 and 0.02 given as its thickness, relative dielectric constant and loss tangent[1-4]. Regarding the feeds, when I_{P11} and I_{P12} are in-phase and the same in magnitude, it is the common-mode signaling. Out-of-phase, they are the differential-mode signals. Ahead of working on the differential signaling with 2 feeds, the 1-feed case needs to be addressed as the basics. For this, a rigorous evaluation method is adopted, shown as follows[1-3].

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$$Z(f, X_f, Y_f) = \sum_{m=0}^{\infty} \sum_{n=0}^{\infty} \frac{\gamma_{mn} \cdot c_{mni} \cdot c_{mnf} \cdot W_z / (W_x W_y)}{\epsilon \omega / Q + j(\epsilon \omega - \frac{k_{xm}^2 + k_{yn}^2}{\omega \mu})} \quad (1)$$

where

$$C_{mii} = \cos(k_{xm}X_i)\cos(k_{yn}Y_i)\text{sinc}(k_{xm}P_{xi}/2)\text{sinc}(k_{yn}P_{yi}/2)$$

$$k_{xm} = m/W_x, k_{yn} = n/W_y,$$

$$Q = [\tan \delta + \sqrt{2/\omega\mu_0\kappa W_z^2}]^{-1}$$

This 1-feed case can be expanded to the differential and common-mode signaling by the superposition principle[4]. Furthermore, the slit structure can be solved by the segmentation scheme as done in [3] and details are not repeated here.

3. Results for validation

Firstly, the impedance is evaluated on the power-bus structure with the differential signals so as to verify whether Eqn. (1) is numerically well-implemented. For the same environment as [3], Eqn. (1) and the FDTD approaches are used and compared. Stating again the structure, the geometry and frequency range are the same as [3], where 54mm33.5mm1.1mm, (27mm, 17.2mm), (27mm, 16.3mm), (41.8mm, 27.4mm) are given to $W_xW_yW_z$, (X_{PII}, Y_{PII}) , (X_{PI2}, Y_{PI2}) , and (X_{PO}, Y_{PO}) .

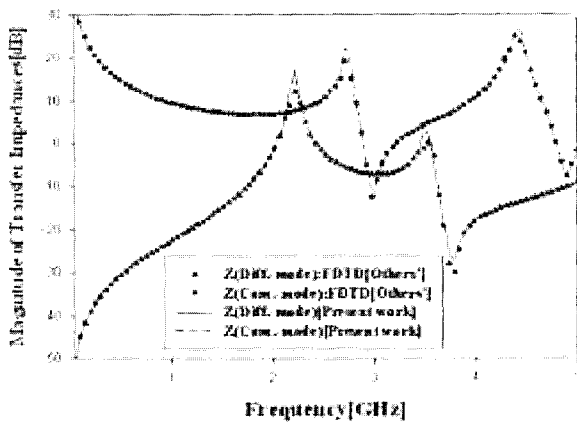
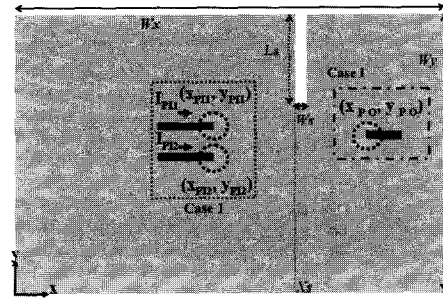
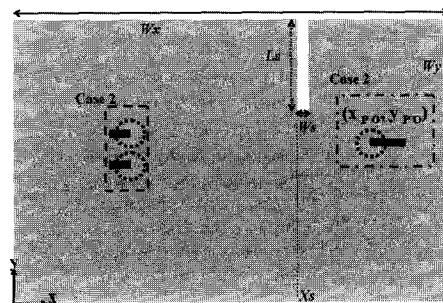


Fig. 2 Differential & Common-mode signaling for the PDN without a discontinuity

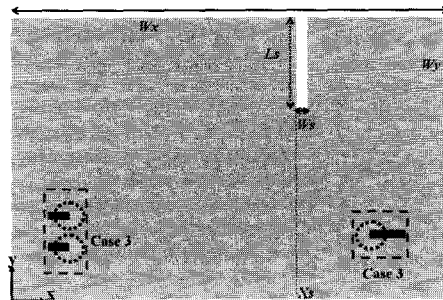
The results are in good agreement between the present method and FDTD[3]. It is noticed that the differential signals lower the impedance level and outperforms the common-mode signals. As of now, a slit is considered starting from Fig. 3.



(a) Port configuration case 1 with Differential & Common-mode signaling for the PDN



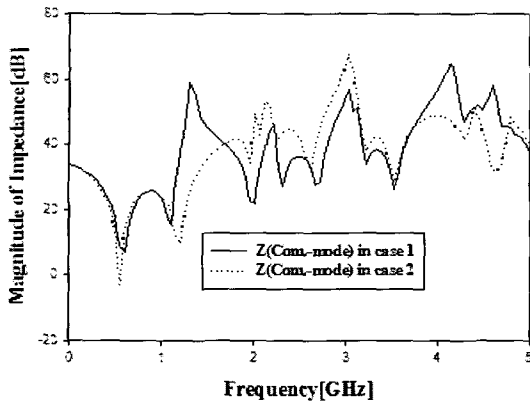
(b) Port configuration case 2 with Differential & Common-mode signaling for the PDN



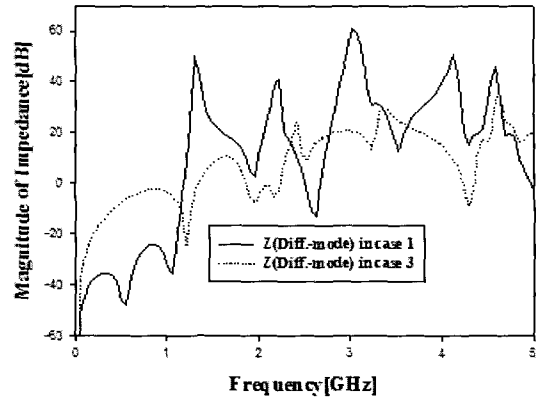
(c) Port configuration case 3 with Differential & Common-mode signaling for the PDN

Fig. 3 Three port configuration cases of Differential & Common-mode signaling for the PDN with a discontinuity

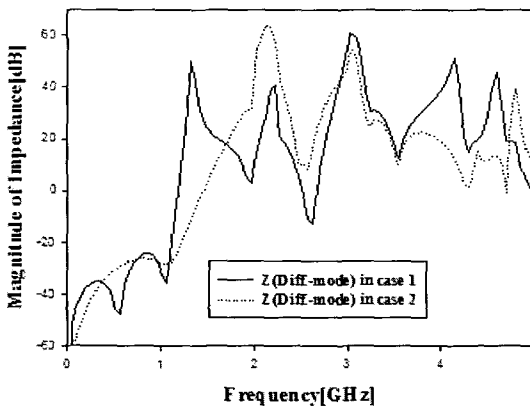
Case 1 has (27mm, 17.2mm)~(27mm, 16.3mm), and case 2 has (18mm, 17.2mm)~(18mm, 16.3mm) for feeding with (41.8mm, 27.4mm) as (X_{PO}, Y_{PO}) in common. ($X_s=36$ mm, $L_s=10$ mm) and $W_s= 2$ mm are given to the slit. Also test case 3 is with (14mm, 7.9mm), (14mm, 6.9mm) and (41.8mm, 7.5mm) as (X_{PII}, Y_{PII}) , (X_{PI2}, Y_{PI2}) , and (X_{PO}, Y_{PO}) in order.



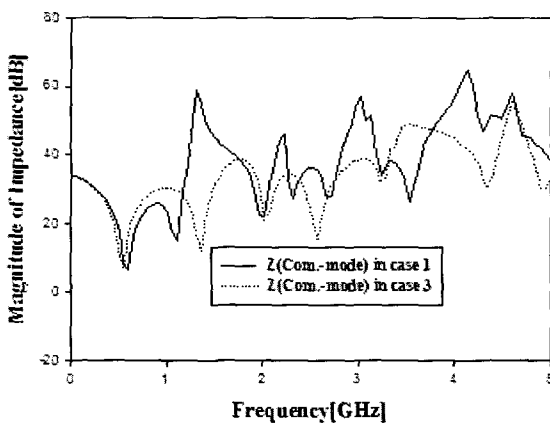
(a) Resultant Impedance profile of cases 1 & 2 with Common-mode signaling for the PDN



(d) Resultant Impedance profile of cases 2 & 3 with Differential-mode signaling for the PDN



(b) Resultant Impedance profile of cases 1 & 2 with Differential-mode signaling for the PDN



(c) Resultant Impedance profile of cases 1 & 3 with Common-mode signaling for the PDN

Fig. 4 Three cases of impedance profiles on Differential & Common-mode signaling for the PDN with a slit as a discontinuity

Compared to Fig. 2 (without the slit), cases 1 and 2 have an increased level of impedance with more resonance points despite the differential feeding, because the slit makes the current path longer and imbalance between 2 feeding paths. Lastly, differential feeding can be much improved by selecting case 3-scheme. Seeing Fig. 4(d), the performance is remarkably improved with (14mm, 7.9mm), (14mm, 6.9mm) and (41.8mm, 7.5mm) as (X_{PI1}, Y_{PI1}) , (X_{PI2}, Y_{PI2}) , and (X_{PO}, Y_{PO}) , because the current path is placed so that fed signals be not disturbed by the slit.

4. Conclusion

The discontinuity of a slit is considered and its influence is rigorously analyzed on the differential signaling in the power-distribution network. And an effective way has been suggested to improve the performance.

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