

Novel Phase Noise Reduction Method for CPW-Based Microwave Oscillator Circuit Utilizing a Compact Planar Helical Resonator

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ABSTRACT—This letter describes a compact printed helical resonator and its application to a microwave oscillator circuit implemented in coplanar waveguide (CPW) technology. The high quality (Q)-factor and spurious-free characteristic of the resonator contribute to the phase noise reduction and the harmonic suppression of the resulting oscillator circuit, respectively. The designed resonator showed a loaded Q -factor of 180 in a chip area of only 40% of the corresponding miniaturized hairpin resonator without any spurious resonances. The fully planar oscillator incorporated with this resonator showed an additional phase noise reduction of 10.5 dB at a 1 MHz offset and a second harmonic suppression enhancement of 6 dB when compared to those of a conventional CPW oscillator without the planar helical resonator structure.

Keywords—Helical resonator; oscillator; phase noise.

I. Introduction

The quality (Q)-factor of a resonator is one of the most important parameters in the design of oscillator circuits because it determines the phase noise performance of an oscillator [1], [2]. Although a high- Q dielectric resonator has been widely used to reduce the phase noise of an oscillator circuit, researchers have recently endeavored to replace this type of dielectric resonator with a novel high- Q planar component because of the inherently bulky size and little compatibility with monolithic microwave

integrated circuits (MMICs) of a dielectric resonator [1], [2]. Miniaturization of the chip also has been another important design criteria in planar resonator design due to the high cost of the gallium arsenide substrate used in MMIC fabrication. Y.T. Lee and others achieved a size reduction of a planar resonator by introducing a small-sized spiral resonator instead of a conventional half wavelength resonator [3]. In their study, however, very little increment in Q -factor was observed when compared to that of the conventional resonator due to the similar operation principles between them. Although J. Lee and others reported a high- Q planar resonator incorporating an active filter into the conventional miniaturized hairpin resonator, the increased circuit size and complexity of the circuit remained as disadvantages to overcome [4]. In addition, because these results are based on microstrip technology, the direct application of these results to the CPW-based oscillator circuit is inherently limited.

In this letter, we present a novel planar high- Q resonator and its application to a microwave oscillator circuit with reduced phase noise and superior harmonic suppression characteristics in CPW technology. For the resonator component of the oscillator circuit, we modified a planar printed helical resonator, which was originally suggested by C. Broomfield and J. Everard in 2003, for the application of a planar filter implemented using microstrip technology [5]. From the introduction of this resonator, we could achieve the high- Q and spurious-free resonance characteristic of a resonator and the resulting low phase noise and superior harmonic suppression characteristics of an oscillator circuit in a very small chip area. Although the fabrication processes were done using a hybrid method, the circuit can be directly applied to an MMIC because of the perfect planar characteristics of the components.

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II. Design of a Planar Helical Resonator

Figure 1 shows our planar helical resonator (PHR) in CPW technology, which consists of two helices coupled to a transmission line with a characteristic impedance of 50Ω . Each helix has a three-quarter turn metal strip on each side of the substrate connected through a metal via hole. To make a metal strip pattern on both sides of a substrate, instead of a microstrip line we used a CPW line without a metal plane on the bottom as a transmission line.

The structure was fabricated on an RT/Duroid 6010 substrate with a dielectric constant of 10.2 and a thickness of 25 mil. We calculated the width of the center line and the gap between the center line and each ground plane of the 50Ω CPW transmission line to be 0.9 mm and 0.55 mm, respectively. To accommodate the helix cells and to improve the impedance matching, we inserted tapered lines, as well as slots that were wider than the external diameter of the unit cell. Then, after considering the minimum allowable resolution of the etching process and miniaturization of the unit cell, we set the width of each helix to 0.2 mm, and designed the radius of the helix to be 0.9 mm in order to provide a resonant frequency at 5.5 GHz. The size of the unit resonator was only $2 \text{ mm} \times 2 \text{ mm}$, which corresponds to $0.094 \lambda_g \times 0.094 \lambda_g$, where λ_g is the guided wavelength at a resonance frequency of 5.5 GHz. As can be seen in Fig. 2, this circuit area is measured to be only 40% of that of the conventional miniaturized hairpin resonator, resonating at the same frequency of 5.5 GHz. This size reduction is due to the equal distribution of the standing wave inside the resonator between the upper and lower metal rings.

Figure 3 shows the simulated and measured S-parameter characteristics of the resonator. For comparison, the measured S-parameter characteristic of the corresponding miniaturized hairpin resonator is also given in the figure. The loaded Q-factor of the helical resonator was calculated as 180, whereas that of the conventional hairpin resonator was only 32.9.

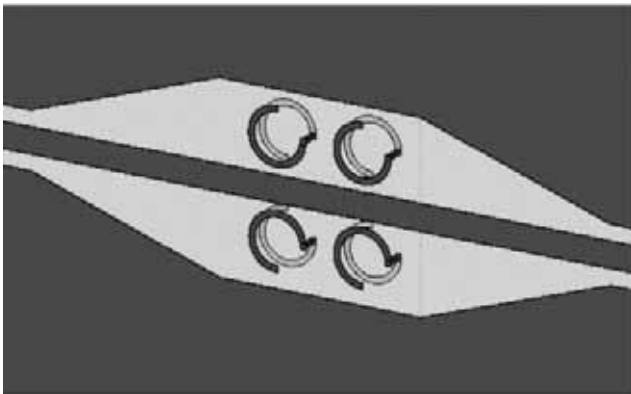


Fig. 1. Planar helical resonator coupled to CPW line.

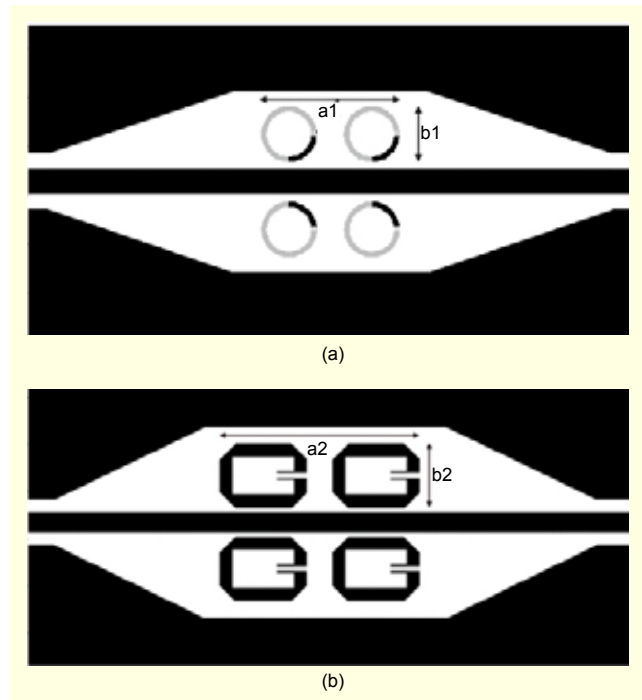


Fig. 2. (a) layout of the proposed planar helical resonator ($a_1 = 5 \text{ mm}$, $b_1 = 2 \text{ mm}$) and (b) layout of the conventional miniaturized hairpin resonator ($a_2 = 8.87 \text{ mm}$, $b_2 = 2.84 \text{ mm}$).

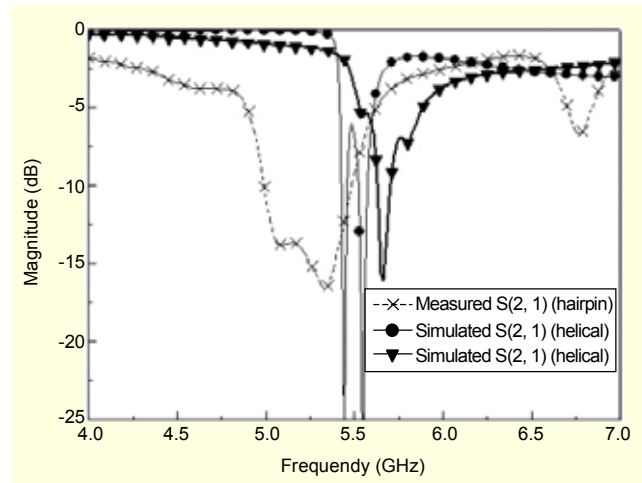


Fig. 3. Simulated and measured S-parameter characteristics of the resonator.

This high Q-factor was reported to originate from the increased energy storage per unit volume of the resonator due to the 3D helical nature of the structure. Additionally, the loss reduction in the via hole of the resonator by ensuring the exact location of the null point of the standing wave current at this point also contributes to the increase in Q-factor [3].

Another advantage of the PHR is its harmonic-free characteristic. The spurious resonances that are observed in most conventional quarter- or half-wavelength resonators are

not observed in the helical resonator. This phenomenon is due to the fact that the inductance and the capacitance of the helical resonator are generated from the helical ring and the gap between the rings on each side of the substrate, respectively. This LC resonance characteristic of the helical resonator generates no spurious resonances, whereas most planar resonators are vulnerable to the harmonic generation. The parasitic attenuation of the insertion loss characteristic over 10 GHz is believed to originate from the impedance mismatch of the tapered line. However, this problem does not affect the resonance generation at 5.5 GHz and can be easily solved by a more careful matching circuit design. These spurious-free resonance characteristics of the PHR contribute to the superior harmonic suppression and high efficiency of the resulting oscillator circuit.

III. Oscillator Design

Figure 4 shows the layout of the 5.5 GHz oscillator incorporated with the high-Q helical resonator. The negative resistance was generated from the transistor with a shunt stub added to its source terminal. A small signal oscillation condition at 5.5 GHz was met by tuning the output matching stubs of the circuit. The final circuit was fabricated by etching a metal layer on the substrate using a standard photo/mask etching technique and soldering an Agilent ATF-36077 packaged pHEMPT transistor. Lumped elements are completely removed for the planarity and easier fabrication of the circuit.

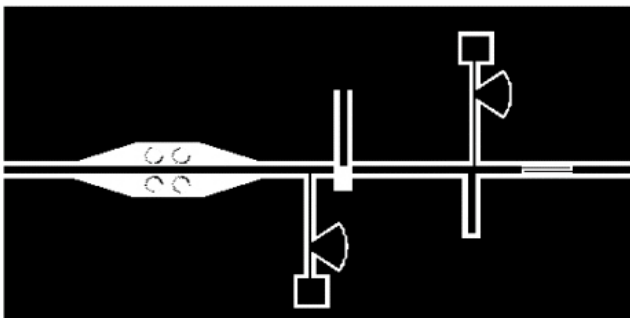


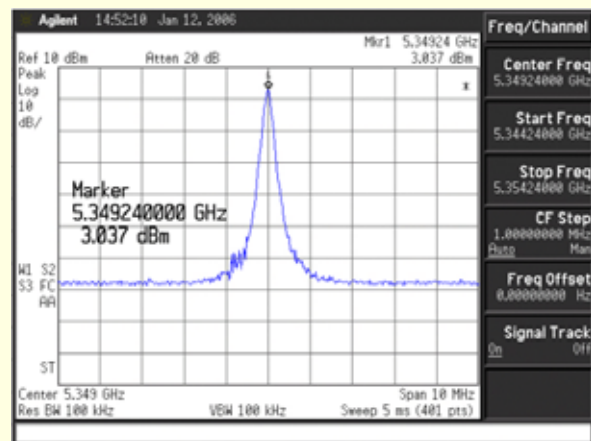
Fig. 4. Layout of the oscillator circuit.

IV. Measurement Results

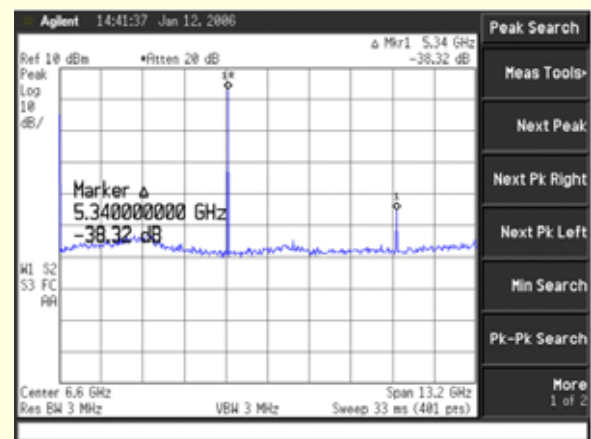
A typical output spectrum of the fabricated oscillator, measured using an Agilent 8565E spectrum analyzer, is shown in Fig. 5. The oscillator showed stable oscillation at a frequency of 5.3 GHz, which is similar to the expected result from the simulation. The output power of the fundamental oscillating signal was measured as 3.037 dBm with the second harmonic

suppression of -38.32 dBc at the bias condition of $V_{gs} = -0.2$ V and $V_{ds} = 1.5$ V. The phase noise was measured as -91 dBc/Hz and -112 dBc/Hz at the offsets of 100 kHz and 1 MHz, respectively.

In Fig. 6 for comparison, we show the measured fundamental output spectrum and harmonic performance of the reference CPW oscillator without any PHR structure. Other parts of the PHR-based oscillator, with the exception of the resonator component, were designed to be perfectly identical to the reference oscillator. The same bias condition, layout, and transistor were also used in both oscillators. The output power of the reference oscillator, oscillating at 5.31 GHz, was measured as 3.1 dBm with a 32.51 dB rejection of the second harmonic. The phase noise was measured as -81 dBc/Hz and -101.5 dBc/Hz at the offsets of 100 kHz and 1 MHz, respectively. These results constitute a 6 dB reduction in second harmonic suppression and a phase noise improvement of 10.5 dB at a 1 MHz offset of the newly developed oscillator when compared to those of a conventional CPW oscillator without PHR structure.



(a)



(b)

Fig. 5. (a) Typical output spectrum and (b) harmonic characteristic of the fabricated oscillator ($V_{ds} = 1.5$ V and $V_{gs} = -0.2$ V).

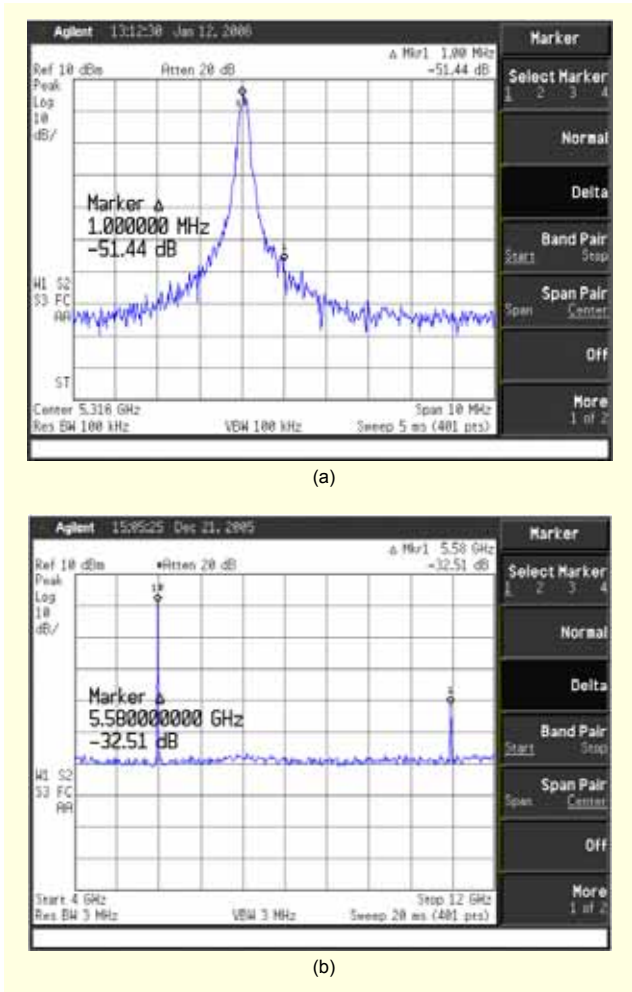


Fig. 6. Measured output spectrums of the reference oscillator without the PHR structure: (a) fundamental output power spectrum and (b) harmonic characteristic.

V. Conclusion

We designed and fabricated a novel compact planar microwave oscillator based on a PHR. The introduction of the printed helical resonator was verified to be effective in reducing the phase noise and in enhancing the harmonic performance of an oscillator circuit without any use of a bulky 3-D resonator or additional compensation circuit. The fully planar and compact nature of the circuit shows perfect compatibility with the low cost monolithic application of the chip.

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