

WDM 광패킷스위칭 시스템을 위한 이득 제어 어븀첨가 광섬유증폭기

정회원 양충열*, 황현용*, 홍현하*, 김해근*, 김환우**

A Gain-Clamped Erbium-Doped Fiber Amplifier (GC-EDFA) for WDM Optical Packet Switching System

Choong-Reol Yang*, Hyun-Yong Hwang*, Hyun-Ha Hong*, Hae-geun Kim*, Hwan-Woo, Kim**

Regular Members

요 약

1545~1560nm 파장의 C 대역에서 0.8 nm 간격의 16 채널 WDM 을 갖는 펌프 LD 타입의 2단 이득 제어 어븀 첨가 광섬유증폭기(GC-EDFA)를 버스트 패킷 모드 광스위칭 시스템에 사용하기 위하여 실험을 통해 입증하였다.

ABSTRACT

A dual-stage gain-clamped erbium doped fiber amplifiers (GC-EDFA) having pump laser diode and 16-channel wavelength division multiplexing (WDM) of 0.8 nm spacing in C band of 1545 ~ 1560 nm wavelength was demonstrated to be used in the burst packet mode optical switching system through experimental setup.

I. Introduction

In case some channels are added/dropped from the WDM network due to any reconfiguration or any partial failure of the network, the input signal, which is transmitted to an Erbium-doped fiber amplifier (EDFA) from the WDM network, is varied, which in turn causes an undesirable power transient or output variation^[1] to take place in the surviving channels. Therefore, the WDM packet mode optical switching system, where the traffic input condition is varied in burst, is required to have an amplifier having a characteristic of amplifying the signal constantly irrespective of the transient due to any variation in some channels. A conventional EDFA would experience gain transients for the drop or add of one or more channels due to channel failure or

burst traffic. These transients are reflected the remaining channels and may cause substantial degradation in the system's bit error rate. Transient effects as well as gain fluctuation resulting in degradation of signal quality in the surviving channel should be suppressed and thus a gain clamping is required. This configuration is referred to as the GC-EDFA. the GC-EDFA with dynamic gain and flatness controller that is fast enough to ensure reliable services continuously in the surviving channels, when one or more channels are suddenly dropped or added, as may be experienced when a system reconfiguration or fault interrupts some of the channels. The conventional EDFA, which is generally used in the transmission system, is subject to the characteristic definition that the input signal traffic is kept uniformly or constantly, and it is,

* 한국전자통신연구원(cryang@etri.re.kr)
논문번호: K01056-0206, 접수일자: 2001년 2월 6일

** 충남대학교 전자공학과 교수

therefore, unavailable for the burst packet mode WDM optical switching system. We have demonstrated that the GC-EDFA comprising a double pump laser diode (LD) and an electrical automatic gain controller (AGC) is available for the packet mode optical switching system.

II. Proposed GC-EDFA

The problem of dynamic gain saturation dependent on the input power has been successfully overcome by using either optical or electronic stabilization approach. The optical approach involves establishing lasing action at a desired wavelength to clamp the amplifier's average population inversion. But it may occur the degradation owing to the relaxation oscillations in the laser^[2]. In the electrical control approach, a probe signal is launched to the input through a low splitting ratio coupler, the optical gain is controlled by adjusting the pump power while monitoring the output power of a probe signal through a low splitting ratio coupler at a fixed wavelength and input power. This configuration is referred to as the pump controlled EDFA. A fast pump control EDFA basis rather than dummy link basis does not increase the complexity of the network's EDFA and is simple, less expensive, and well suited to the packet switching system.

Recently, fast pump control method has been available to limit the power excursions of surviving channels against fast power transients. For the packet mode switching system, a fast

pump control circuit in an EDFA with several s of response time was required to maintain the constant power. In the conventional electronic gain control approach, the optical gain is controlled by adjusting the pump power while monitoring the output power of a probe signal at a fixed wavelength and input power through a dual stage pump laser control^[3].

In this paper, we propose and demonstrate a new scheme to protect surviving channels by controlling the pump power while monitoring the total and output power. And we experimentally investigated the power transient of GC-EDFA and demonstrated the feasibility of proposed scheme.

Fig. 1 illustrates a block diagram of proposed gain-clamped EDFA. To achieve both low noise figure and high pump efficiency, it has been recognized that the use of a dual stage configuration has been advantageous. The GC-EDFA removes 1530~1544 nm band by using the amplified spontaneous emission (ASE) filter, and uses 1545~1560 nm band for the purpose of designing 16 WDM channels with a wavelength spacing of 0.8 nm in C band. The maximum strength of the entire input signal in the GC-EDFA is -4 dBm, and the power per channel for 10^{-12} BER (bit error rate) is equivalent to -16 dBm. The GC-EDFA represents a case that the gain is flatted at 20 dB. In this case, during the input power is varied in the range of -4 dBm to -19 dBm, the gain is maintained at 20 ± 0.5 dB in the gain band of 15 nm without any variation in the gain spectrum in C band of 1545~1560 nm, and a

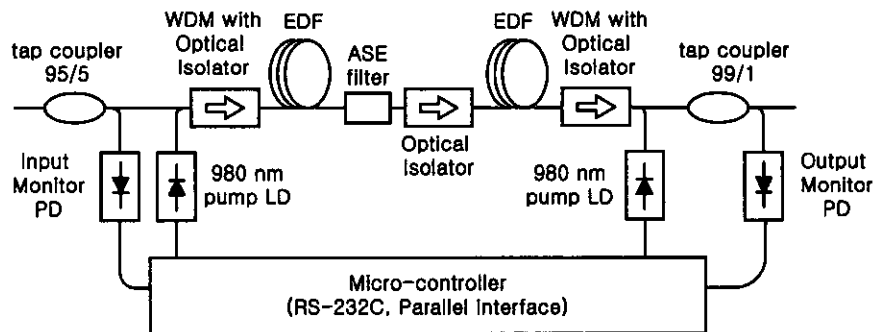


Fig. 1 Structure of dual-stage GC-EDFA using a pump power control method

very flattened gain spectrum of 1~1.2 dB is realized. After tapping the optical signal by using two tap couplers of 95:5 and 99:1 at the input and the output end of the EDFA, respectively, the data monitored through the input monitor PD and the output monitor PD are transmitted to the micro-controller to control the output of the 980 nm pump LD. The input optical signal of the GC-EDFA and the pumping optical signal of the 980 nm pump LD are combined through the WDM with the optical isolator attached thereto, and further are amplified through the EDF. Fig. 2 illustrates the output spectra of the GC-EDFA depending on the input power. As shown in the figure, we can see that the GC-EDFA has a uniform gain of 20 dB even though the input power is varied from -19 dBm to -4 dBm.

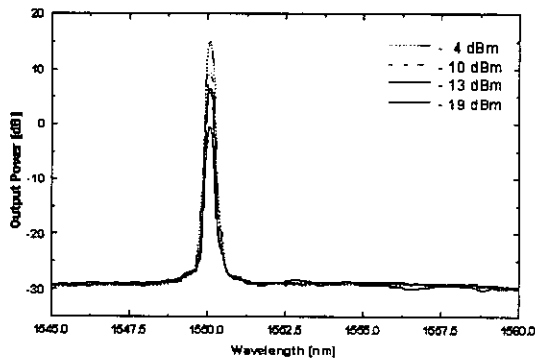


Fig. 2 Output spectra according to the input power

III. Experimental setup

Fig. 3 describes the experimental setup for measuring the dynamic eye pattern and any power transient occurring in other survival channels in case some channels are dropped when the GC-EDFA transmits an optical signal having 16 channels of a uniform size at an interval of 0.8 nm (100 GHz) in the wavelength area of 15 nm from 1545 nm to 1560 nm. A distributed feedback laser diode (DFB-LD) having 2 wavelengths of $\lambda_1 = 1545$ nm and $\lambda_2 = 1555$ nm were used as a light source, and any channel in the network was added/dropped by using acousto optic modulators. We can obtain the eye

pattern on oscilloscope by supplying external modulated data of the 2.5 Gbps using the pulse pattern generator (PPG) at the stage following the DFB-LD 1. In case 15 channels are dropped in the 16 channels WDM networks, and in order to measure the power transient of such one operating channel, the power value of one channel was assigned to the DFB-LD 1, the operating channel, while the power value of other 15 channels was applied to the DFB-LD 2. The two light sources were coupled by using a 3 dB coupler. Then, any output port was connected with the GC-EDFA to amplify the signal, and in order to measure the power transient of the operating channel, an optical bandpass filter having the center wavelength of 1545 nm to transmit was used and it was measured with a sampling oscilloscope (30 GHz), while in the other output ports, the input power to the GC-EDFA was measured by using an optical power meter. In order to measure the power transient taking place in the other operating channel under the worst case, we considered that 15 channels of all 16 channels were dropped and it was modulated with the square wave (~5 MHz) from AOM. Thereby, the power transient of the operating channel was measured.

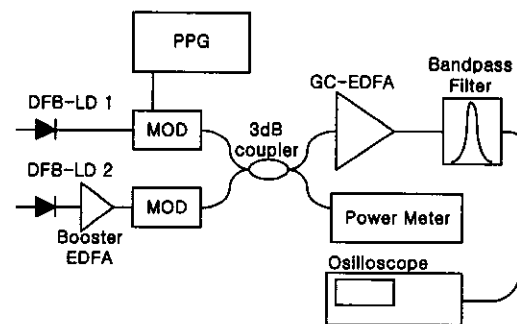


Fig. 3 Experimental setup

IV. Results and Discussion

Fig. 4 and Fig. 5 show the power transient characteristics occurring in the survival channels after the power pass through the GC-EDFA. Fig. 4 depicts the power transient occurring in the survival channels in case 8 channels, 12 channels

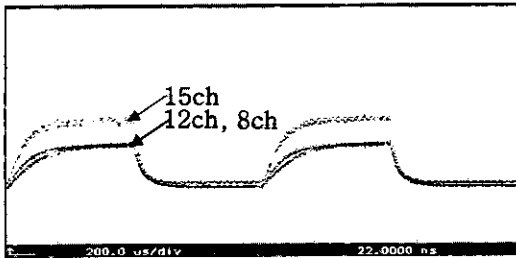


Fig. 4 Power transient characteristics of GC-EDFA due to variation in channel; in case of 1 kHz modulation frequency when 8 chs, 12 chs and 15 chs are dropped, respectively

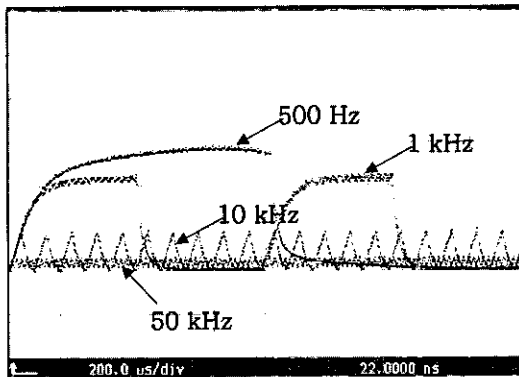
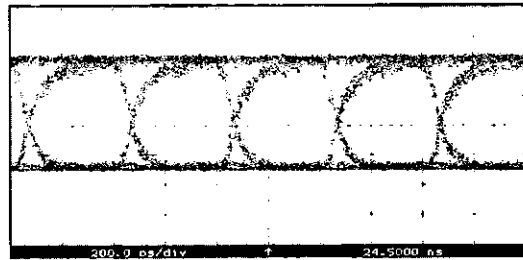


Fig. 5 Power transient characteristics of GC-EDFA due to variation in frequency under the worst case; in case of 500 Hz, 1 kHz, 10 kHz and 50 kHz modulation frequency when 15 chs are dropped

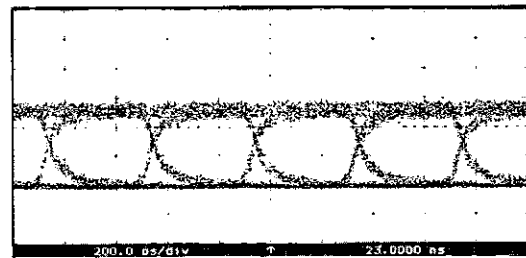
and 15 channels of all 16 channels are dropped respectively. The modulation frequency of AOM is 1 kHz. As shown in this figure, we can see that the more channels are dropped, the larger the power transient occurred. Fig. 5 depicts the power transient occurring in the survival channels depending upon respective variations of the modulation frequency to 500 Hz, 1 kHz, 10 kHz and 50 kHz, respectively. We considered the worst case that the 15 channels of all 16 channels are dropped. As shown in the figure, we can see that the higher the modulation frequency is, that is, the faster add and drop of channels occurred, the slower the reaction time of EDF is, and consequently, the power transient becomes smaller gradually.

The characteristics as shown in Fig. 4 and Fig. 5 illustrate that the GC-EDFA used in the experiment determined a speed that is required for protecting the operating channel through such



(a)

Fig. 6(a) Dynamic eye at 2.5 Gbps data, when 1 ch of all 16 channels is dropped with 5 Hz modulation frequency



(b)

Fig. 6(b) Dynamic eye at 2.5 Gbps data, when 8 chs of all 16 channels are dropped with 10 kHz modulation frequency

characteristics and further responded to the input signal up to about 10 kHz. And, since the EDFA has a slow dynamics (several 10 ms lifetime), in case that the input power was modulated at 10 kHz or over (below 100 μ s), an infinitesimal power transient took place. In this case, the inter-optical packet spacing (the responding speed of a driver) needs to be faster than 100 μ s. Fig. 6 (a) and Fig. 6 (b) illustrate the dynamic eye pattern obtained from 2.5 Gbps data that are supplied from Gbps PPG of experimental setup shown in Fig. 2. The measurements of dynamic characteristic were repeatedly carried out at 5 Hz, 50 Hz, 500 Hz, 10 kHz and 50 kHz while 1 channel, 8 channels and 15 channels of all 16 channels were dropped. Fig. 6 (a) and Fig. 6 (b) depict the power transient measured in case of 1 channel of all 16 channels at 5 Hz and 8 channels of all 16 channels at 10 kHz, respectively. From these test results, we found clamping frequency of the GC-EDFA was less than 5 Hz, and the reaction frequency of

GC-EDFA was greater than 10 kHz. We can see that in case the modulation frequency is below 5 Hz, the power transient never occurs in survival channels because the response time of the operating circuit for the gain clamping of this GC-EDFA is 200 ms. However, we also can see that in case the modulation frequency is over 5 Hz, the power transient occurs in the upper part of the eye pattern because the GC-EDFA fails to clamp any gain. Further, we can see that in case the modulation frequency is 10 kHz and 50 kHz respectively, channels are inserted and extracted fast, that is, the reaction time of EDF be slow, and consequently, the power transient does not occur. Through Fig. 6 (a) and Fig. 6 (b), we can see that the fast the modulation frequency is, i.e., the fast add and drop of channels occurred, the reaction time of EDF becomes slow, and consequently, the power transient does not occur.

V. Conclusion

We have newly designed the gain-clamped EDFA for providing an uniform gain across 16 WDM channels at an intervals of 0.8 nm (100 GHz) in the C band range from 1545 nm to 1560 nm by using the dual stage pump power control method. An experimental setup has been configured to measure the power transient of the GC-EDFA, and the use of the AOM has been to take the effect of insertion and extraction of channels. This experimental setup has confirmed that the EDFA adopting the pump power control method have been available for the optical packet switch system. This study has demonstrated that GC-EDFA overcome the output variation and the power transient impact resulting from the add/drop of any channel, and further had a characteristic of maintaining the optical signal amplification gain uniformly under the burst mode packet traffic, and therefore, it is available for the burst packet mode optical switching system. We have confirmed the feasibility of the GC-EDFA to the packet switching system through the dynamic eye pattern, which was demonstrated at 2.5 Gbps

optical data signal in WDM network. GC-EDFA is also suitable for multiwavelength networks using the internet packet (IP) or packet switching. A further study is being conducted for embodying a pump power driver with a high-speed response time of 100 μ s or over to a high-speed power transient. Currently, the research on the packet mode GC-EDFA considering the WDM burst mode packet traffic is in progress, and the WDM burst packet mode GC-EDFA with the aforementioned characteristics would be available for the packet switching system and the IP-based system as well.

ACKNOWLEDGMENT

This study has been performed as one of the advanced projects under the initiative of Korea Ministry of Information Communication.

Reference

- [1] S.Y. Park, et al., "Doped fiber and pump power of gain-flattened EDFA's," *Electronics Letters*, vol. 32, pp. 2161, 1996.
- [2] G. Luo, et al., "Experimental and Theoretical Analysis of Relaxation-Oscillations and Spectral Hole Burning in All-Optical Gain-Clamped EDFA's for WDM Networks," *J. Lightwave Technol.*, vol. 16, pp. 527, 1998.
- [3] A. K. Srivastava, et al., "Fast-Link Control Protection of Surviving Channels in Multi-wavelength Optical Network," *IEEE Photon. Technol. Lett.*, vol. 9, pp. 1667, 1997.

양 충 열(Choong-reol, Yang)



1979년 3월~1983년 2월 : 건국
대학교 전자공학과
(학사)
1989년 1월~1989년 12월 : 미국
GDLS사 파견 연구원
1986년 7월~1992년 6월 : (주)
현대정공 기술연구소
1992년 6월~현재 : ETRI 광패킷교환팀 선임연구원
1998년 8월 : 충남대학교 대학원 전자공학과(석사)

현재 : 충남대학교 대학원 전자공학과 (박사과정)
<주관심 분야> 광통신, 광패킷스위칭, 광라우터
cryang@etri.re.kr

황 현 용(Hyun-yong, Hwang)



1994년 3월~1998년 2월 : 호서
대학교 전자공학과
(학사)
1998년 3월~2000년 2월 : 숭실
대학교 전자공학과
(석사)
2000년 5월~현재 : ETRI 광패킷
교환팀 연구원

<주관심 분야> 광패킷스위칭

홍 현 하(Hyun-Ha Hong)



1979년 2월 : 광운대학교
전자공학과(학사)
1981년 2월 : 연세대학원
전자공학과(석사)
현재 : KAIST 전기과 박사과정
1985년 10월~현재 : ETRI, 광패
킷교환팀장

<주관심 분야> 광패킷스위칭, 광버스트스위칭, 광라
우터

김 해 근(Hae-geun Kim)

1977년 2월 : 경북대학교 전자공학과 (학사)
1982년 8월 : 연세대학교 전자공학과 (석사)
1994년 8월 : University of South Florida 전기공학과
(박사)
1980년~2001년 8월 현재 : ETRI, 광라우팅연구부장
<주관심 분야> 광통신

김 환 우(Hwan-Woo, Kim)



1977년 2월 : 서울대학교
전자공학과(학사)
1979년 2월 : 한국과학기술원
전기전자공학과(석사)
1988년 6월 : University of
Utah, USA(P.H)
1979년~현재 : 충남대학교
전자공학과 교수

<주관심 분야> 신호처리, 디지털 통신, 이동 통신