

An Effective Method in Analyzing a Class of Bilinear Systems via Taylor Polynomials

(Taylor 다항식에 의한 양선형 시스템의 효과적인 해석법)

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要 約

본 논문에서는 Taylor 다항식에 의해 양 선형 시스템을 해석하는 효과적인 방법이 제안된다. Yang 과 Chen에 의해 유도된 결과는 미지의 상태 벡터가 닫혀진 형태로 구해지지 않고 또한 사용하는 항이 증가할 때 큰 차원의 선형 대수 방정식을 풀어야하는 문제점을 가지고 있지만, 본 논문에서 제안된 방법은 미지의 상태 벡터가 닫혀진 형태로 구해지며 선형 대수 방정식을 풀 필요가 없다.

Abstract

In this paper, an effective method in analyzing a class of bilinear systems via Taylor polynomials is proposed. The result derived by Yang and Chen shows an implicit form for unknown state vector and requires to solve a linear algebraic equation with large dimension when the number of terms used increase. In comparison to the result of Yang and Chen, the method in this paper gives a closed form for unknown state vector and does not need to solve any linear algebraic equation.

I. Introduction

There are many practical systems whose mathematical model may be expressed by bilinear forms, such as nuclear reactor kinetic equations, biological, socio-economic or ecological systems [1]-[3]. The analytic solutions for time-varying bilinear systems cannot be found in general [4]-[5]. However, when those systems are to be analyzed,

it is desirable to obtain an analytic solution since numerical solution cannot be compared with analytic one [6].

Recently, as a method of approximately finding an analytic solution of time-varying bilinear systems, series expansion approach has been proposed [7]-[11]. Yang and Chen [11] employed the Taylor polynomials in analyzing a class of time-varying bilinear systems as a better alternative than Block pulse functions [7], Walsh functions [8], or Chebyshev polynomials [9]-[10]. According to the results in Yang and Chen [11], the final linear algebraic equation for the analysis of a class of time-varying bilinear systems is given by

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$$H - [X(0), 0, \dots, 0] = F^* \begin{bmatrix} H_1 \\ H_2 \\ \vdots \\ H_n \end{bmatrix} P_{(r \times r)} + B^* \begin{bmatrix} u_1 \\ u_2 \\ \vdots \\ u_n \end{bmatrix} P_{(r \times r)} \quad (1)$$

Here H and H_i , $i = 1, 2, \dots, n$, are unknown matrices to be determined, $P_{(r \times r)}$ is the operational matrix of integration for the Taylor polynomials. Also, the i -th row and j -th column of an $n \times n$ matrix, F^* , and $n \times q$ matrix, B^* , is an $1 \times r$ coefficient vector of the Taylor series expansion for the function in the i -th row and j -th column of an $n \times n$ matrix $F(t)$ and an $n \times q$ matrix $B(t)$, respectively. In (1), it should be observed that (1) is not a closed form for unknown matrices H and H_i , $i = 1, 2, \dots, n$, and the dimension of (1) is larger when the number r of terms used increase. Therefore it becomes very difficult to solve (1).

In this paper, we suggest another approach of analyzing a class of time-varying bilinear systems via Taylor polynomials in order to overcome the problems indicated above. The final form derived in this paper is a closed form for unknown state vector and does not include the operation of solving any linear algebraic equation. Also, the inverse of a matrix is not needed in this approach.

II. Problem Formulation

When a function $f(t)$ is expanded at the neighborhood of $t = 0$ as

$$f(t) = \sum_{i=0}^{\infty} f_i \phi_i(t), \quad t \in [0, 1] \quad (2)$$

where

$$\phi_i(t) = t^i, \quad f_i = \frac{1}{i!} \left(\frac{d^i f(t)}{dt^i} \right) \Big|_{t=0} \quad (3)$$

$\phi_i(t)$, for $i=0, 1, \dots$, are called Taylor series basis functions [12]. Also a linear combination of finite number of Taylor series basis functions is called a Taylor polynomial. For an integer $m \geq 1$, let $\phi_{(m)}(t)$ be the m -vector function defined by

$$\phi_{(m)}^T(t) = [\phi_0(t), \phi_1(t), \dots, \phi_{m-1}(t)], \quad t \in [0, 1] \quad (4)$$

where the superscript T denotes the transpose.

Let us consider the following bilinear systems

$$\dot{x}(t) = A(t)x(t) + B(t)u(t) + \sum_{k=1}^r N_k(t)x(t)u_k(t)$$

$x(0)$ = initial vector

(5)

where $x(t)$ is an n -dimensional state vector, $u(t)$ is an r -dimensional input vector, and $A(t)$, $B(t)$, and $N_k(t)$ are $n \times n$, $n \times r$, and $n \times n$ time-varying matrices, respectively, and $U_k(t)$, $k=1, 2, \dots, r$, are the entries of $u(t)$.

Now let the matrix $F(t)$ be defined as

$$F(t) = A(t) + \sum_{k=1}^r N_k(t)u_k(t) \quad (6)$$

Then eqn. (5) becomes

$$\dot{x}(t) = F(t)x(t) + B(t)u(t) \quad (7)$$

For each $i, j=1, 2, \dots, n$, let $f_{ij}(t)$ be the entry in i -th row and j -th column of $F(t)$ and for each $i=1, 2, \dots, n$ and $k=1, 2, \dots, r$, let $b_{ik}(t)$ be the element in i -th row and k -th column of $B(t)$. Assume that all elements of $F(t)$, $B(t)$ and $U(t)$ are analytic in the time interval $[0, 1]$; then the expansions of those elements via m Taylor series basis functions are as follows:

$$f_{ij}(t) \cong [f_{ij0} \ f_{ij1} \ \dots \ f_{ij,m-1}] \phi_{(m)}(t) \quad (8)$$

$$b_{ik}(t) \cong [b_{ik0} \ b_{ik1} \ \dots \ b_{ik,m-1}] \phi_{(m)}(t) \quad (9)$$

and

$$u_k(t) \cong [u_{k0} \ u_{k1} \ \dots \ u_{k,m-1}] \phi_{(m)}(t) \quad (10)$$

Then, for each $i=1, 2, \dots, n$, the expansion of $x_i(t)$ via m Taylor series basis functions is expressed as

$$x_i(t) \cong [x_{i0} \ x_{i1} \ \dots \ x_{i,m-1}] \phi_{(m)}(t) \quad (11)$$

Our problem is to find a formula, which has an explicit form and does not require the inverse of large matrix, for the vectors $[x_{i0} \ x_{i1} \ \dots \ x_{i,m-1}]$ for $i=1, 2, \dots, n$, in terms of $x(0)$, $[f_{ij0} \ f_{ij1} \ \dots \ f_{ij,m-1}]$ for $ij=1, 2, \dots, n$, $[b_{ik0} \ b_{ik1} \ \dots \ b_{ik,m-1}]$ for $i=1, 2, \dots, n$ and $k=1, 2, \dots, r$, and $[u_{k0} \ u_{k1} \ \dots \ u_{k,m-1}]$ for $k=1, 2, \dots, r$.

III. A Mathematical Preliminary

Let there be given the following two arbitrary

Taylor polynomials each of which is a linear combination of m Taylor series basis functions;

$$p(t) = p_0 \phi_0(t) + p_1 \phi_1(t) + \dots + p_{m-1} \phi_{m-1}(t) \cong p^T \phi_m(t) \tag{12}$$

and

$$q(t) = q_0 \phi_0(t) + q_1 \phi_1(t) + \dots + q_{m-1} \phi_{m-1}(t) \cong q^T \phi_m(t) \tag{13}$$

where

$$p^T = [p_0 \ p_1 \ \dots \ p_{m-1}] \tag{14}$$

$$q^T = [q_0 \ q_1 \ \dots \ q_{m-1}] \tag{15}$$

Then, the multiplication of two functions $p(t)$ and $q(t)$ can be approximated as a linear combination of m Taylor series basis functions as follows [12]:

$$p(t) q(t) \cong h^T \phi_m(t) \tag{16}$$

where

$$h = \begin{bmatrix} p_0 \ q_0 \\ p_0 \ q_1 + p_1 \ q_0 \\ \vdots \\ p_0 \ q_{i-1} + p_1 \ q_{i-2} + \dots + p_{i-1} \ q_0 \\ \vdots \\ p_0 \ q_{m-1} + p_1 \ q_{m-2} + \dots + p_{m-1} \ q_0 \end{bmatrix} \quad \text{i-th row} \tag{17}$$

In the above, the terms whose indices are higher than $\phi_{m-1}(t)$ are neglected. The validity of this approximation [12] is confirmed by the fact that the considered time domain is $t \in [0, 1)$.

Now, the integration of the multiplication of two functions $p(t)$ and $q(t)$ can be approximated as a linear combination of m Taylor series basis functions via the operational matrix for forward integration [12], $P_{(m \times m)}$, as

$$\int_0^1 p(s) q(s) ds \cong \int_0^1 h^T \phi_m(s) ds \cong h^T P_{(m \times m)} \phi_m(t) \tag{18}$$

where

$$P_{(m \times m)} = \begin{bmatrix} 0 & 1 & 0 & 0 & \dots & 0 & 0 & 0 \\ 0 & 0 & \frac{1}{2} & 0 & \dots & 0 & 0 & 0 \\ 0 & 0 & 0 & \frac{1}{3} & \dots & 0 & 0 & 0 \\ \vdots & & & & & & & \\ 0 & 0 & 0 & 0 & \dots & 0 & \frac{1}{m-2} & 0 \\ 0 & 0 & 0 & 0 & \dots & 0 & 0 & \frac{1}{m-1} \\ 0 & 0 & 0 & 0 & \dots & 0 & 0 & 0 \end{bmatrix} \tag{19}$$

We can show that when $p(t)$ is a $n \times n$ matrix and $q(t)$ is a $n \times 1$ vector, $\int_0^1 P(s) q(s) ds$ can be similarly expressed.

Theorem 1.

Let there be given $F(t)$ and $\underline{x}(t)$ as in (7). For each $i, j=1, 2, \dots, n$, let $f_{ij}(t)$ be the entry in i -th row and j -th column of $F(t)$. And let the expansions of $f_{ij}(t)$ and the each function of $\underline{x}(t)$ be expressed as follows:

$$f_{ij}(t) \cong [f_{i,0} \ f_{i,1} \ \dots \ f_{i,m-1}] \phi_m(t), \quad i, j = 1, 2, \dots, n \tag{20}$$

$$x_i(t) \cong [x_{i,0} \ x_{i,1} \ \dots \ x_{i,m-1}] \phi_m(t), \quad i = 1, 2, \dots, n \tag{21}$$

Also, let us define the $n \times n$ matrix F_w and $n \times 1$ vector \underline{X}_w as

$$F_w = \begin{bmatrix} f_{11w} & f_{12w} & \dots & f_{1nw} \\ f_{21w} & f_{22w} & \dots & f_{2nw} \\ \vdots & \vdots & & \vdots \\ f_{n1w} & f_{n2w} & \dots & f_{nnw} \end{bmatrix} \tag{22}$$

$$\underline{X}_w = \begin{bmatrix} x_{1w} \\ x_{2w} \\ \vdots \\ x_{nw} \end{bmatrix} \tag{23}$$

where $w=0, 1, \dots, m-1$.

Then

$$\int_0^1 F(s) \underline{X}(s) ds \cong [F_0 \underline{X}_0, \ F_0 \underline{X}_1 + F_1 \underline{X}_0, \ \dots, \ F_0 \underline{X}_{m-1} + F_1 \underline{X}_{m-2} + \dots + F_{m-1} \underline{X}_0] P_{(m \times m)} \phi_m(t) \tag{24}$$

Proof:

Since

$$F(t) \underline{X}(t) = \begin{bmatrix} f_{11}(t) & f_{12}(t) & \dots & f_{1n}(t) \\ f_{21}(t) & f_{22}(t) & \dots & f_{2n}(t) \\ \vdots & \vdots & & \vdots \\ f_{n1}(t) & f_{n2}(t) & \dots & f_{nn}(t) \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_n(t) \end{bmatrix} = \begin{bmatrix} f_{11}(t)x_1(t) + f_{12}(t)x_2(t) + \dots + f_{1n}(t)x_n(t) \\ f_{21}(t)x_1(t) + f_{22}(t)x_2(t) + \dots + f_{2n}(t)x_n(t) \\ \vdots \\ f_{n1}(t)x_1(t) + f_{n2}(t)x_2(t) + \dots + f_{nn}(t)x_n(t) \end{bmatrix} \tag{25}$$

the integration of $F(t) \underline{x}(t)$ is

$$\int_0^t F(s) \underline{x}(s) ds = \begin{bmatrix} \int_0^t \{f_{11}(s) x_1(s) + f_{12}(s) x_2(s) + \dots + f_{1n}(s) x_n(s)\} ds \\ \int_0^t \{f_{21}(s) x_1(s) + f_{22}(s) x_2(s) + \dots + f_{2n}(s) x_n(s)\} ds \\ \vdots \\ \int_0^t \{f_{n1}(s) x_1(s) + f_{n2}(s) x_2(s) + \dots + f_{nn}(s) x_n(s)\} ds \end{bmatrix} \quad (26)$$

The first row in (26) is

$$\begin{aligned} & \int_0^t \{f_{11}(s) x_1(s) + f_{12}(s) x_2(s) + \dots + f_{1n}(s) x_n(s)\} ds \\ &= \int_0^t f_{11}(s) x_1(s) ds + \int_0^t f_{12}(s) x_2(s) ds + \dots \\ &+ \int_0^t f_{1n}(s) x_n(s) ds \end{aligned} \quad (27)$$

By using the property of (16) and (18), we obtain

$$\begin{aligned} & \int_0^t f_{11}(s) x_1(s) ds \\ & \cong [f_{110} X_{10}, f_{110} X_{11} + f_{111} X_{10}, \dots, \\ & \quad f_{110} X_{1,m-1} + f_{111} X_{1,m-2} + \dots + f_{11,m-1} X_{10}] \\ & \quad P_{(m \times m)} \phi_m(t) \\ & \int_0^t f_{12}(s) x_2(s) ds \\ & \cong [f_{120} X_{20}, f_{120} X_{21} + f_{121} X_{20}, \dots, \\ & \quad f_{120} X_{2,m-1} + f_{121} X_{2,m-2} + \dots + f_{12,m-1} X_{20}] \\ & \quad P_{(m \times m)} \phi_m(t) \\ & \quad \vdots \\ & \int_0^t f_{1n}(s) x_n(s) ds \\ & \cong [f_{1n0} X_{n0}, f_{1n0} X_{n1} + f_{1n1} X_{n0}, \dots, \\ & \quad f_{1n0} X_{n,m-1} + f_{1n1} X_{n,m-2} + \dots + f_{1n,m-1} X_{n0}] \\ & \quad P_{(m \times m)} \phi_m(t) \end{aligned} \quad (28)$$

Therefore the first row in (26) becomes

$$\begin{aligned} & \int_0^t \{f_{11}(s) x_1(s) + f_{12}(s) x_2(s) + \dots + f_{1n}(s) x_n(s)\} ds \\ & \cong [f_{10} X_0, f_{10} X_1 + f_{11} X_0, \dots, \\ & \quad f_{10} X_{m-1} + f_{11} X_{m-2} + \dots + f_{1,m-1} X_0] P_{(m \times m)} \phi_m(t) \end{aligned} \quad (29)$$

where

$$f_{1w} = [f_{11w} \ f_{12w} \ \dots \ f_{1nw}] \quad (30)$$

$$X_w^T = [x_{1w} \ x_{2w} \ \dots \ x_{nw}] \quad (31)$$

for $w=0,1, \dots, m-1$.

Similarly, for the second row in (26), we can obtain easily

$$\begin{aligned} & \int_0^t \{f_{21}(s) x_1(s) + f_{22}(s) x_2(s) + \dots + f_{2n}(s) x_n(s)\} ds \\ & \cong [f_{20} X_0, f_{20} X_1 + f_{21} X_0, \dots, f_{20} X_{m-1} + f_{21} X_{m-2} + \dots \\ & \quad + f_{2,m-1} X_0] P_{(m \times m)} \phi_m(t) \end{aligned} \quad (32)$$

Where

$$f_{2w} = [f_{21w} \ f_{22w} \ \dots \ f_{2nw}] \quad (33)$$

$$X_w^T = [x_{1w} \ x_{2w} \ \dots \ x_{nw}] \quad (34)$$

for $w=0,1, \dots, m-1$, and so forth.

Finally, we find

$$\begin{aligned} & \int_0^t F(s) x(s) ds \\ &= \begin{bmatrix} f_{10} X_0 \ f_{10} X_1 + f_{11} X_0 \ \dots \ f_{10} X_{m-1} + f_{11} X_{m-2} + \dots \\ f_{20} X_0 \ f_{20} X_1 + f_{21} X_0 \ \dots \ f_{20} X_{m-1} + f_{21} X_{m-2} + \dots \\ \vdots \\ f_{n0} X_0 \ f_{n0} X_1 + f_{n1} X_0 \ \dots \ f_{n0} X_{m-1} + f_{n1} X_{m-2} + \dots \\ + f_{1,m-1} X_0 \\ + f_{2,m-1} X_0 \\ \vdots \\ + f_{n,m-1} X_0 \end{bmatrix} P_{(m \times m)} \phi_m(t) \end{aligned} \quad (35)$$

Since

$$\begin{bmatrix} f_{1w} \\ f_{2w} \\ \vdots \\ f_{nw} \end{bmatrix} = \begin{bmatrix} f_{11w} \ f_{12w} \ \dots \ f_{1nw} \\ f_{21w} \ f_{22w} \ \dots \ f_{2nw} \\ \vdots \\ f_{n1w} \ f_{n2w} \ \dots \ f_{nnw} \end{bmatrix} = F_w \quad (36)$$

we obtain

$$\begin{aligned} & \int_0^t F(s) x(s) ds \\ & \cong [F_0 X_0 \ F_0 X_1 + F_1 X_0 \ \dots \ F_0 X_{m-1} + F_1 X_{m-2} + \dots + \\ & \quad F_{m-1} X_0] P_{(m \times m)} \phi_m(t) \end{aligned} \quad (37)$$

This completes the proof.

Q.E.D

IV. Main Result

In the following, we present an efficient method in analyzing a class of bilinear systems via Taylor polynomials.

Theorem 2.

For the system in (7), let the entries of $F(t)$,

$B(t)$, and $\underline{U}(t)$ be expanded as in (8), (9), and (10), respectively, and let the expansion of each function of $\underline{x}(t)$ via m Taylor series basis functions be expressed as

$$x_i(t) \cong [x_{i0} x_{i1} \dots x_{i,m-1}] \phi_m(t), i=1, 2, \dots, n \quad (38)$$

Also, let us define the $n \times r$ matrix B_w and $r \times 1$ vector \underline{U}_w as

$$B_w = \begin{bmatrix} b_{11w} & b_{12w} & \dots & b_{1rw} \\ b_{21w} & b_{22w} & \dots & b_{2rw} \\ \vdots & \vdots & \ddots & \vdots \\ b_{n1w} & b_{n2w} & \dots & b_{nrw} \end{bmatrix} \quad (39)$$

$$\underline{U}_w = \begin{bmatrix} u_{1w} \\ u_{2w} \\ \vdots \\ u_{rw} \end{bmatrix}, \text{ for } w=0, 1, \dots, m-1 \quad (40)$$

Then the $1 \times m$ vectors $[x_{i0} \ x_{i1} \ \dots \ x_{i,m-1}]$ for $i=1, 2, \dots, n$ can be obtained by

$$\begin{aligned} \underline{X}_0 &= \underline{x}(0) \\ X_{v-1} - 0 &= \frac{F_0 \underline{X}_{v-2} + F_1 \underline{X}_{v-3} + \dots + F_{v-2} \underline{X}_0}{v-1} + \\ &\quad \frac{B_0 \underline{U}_{v-2} + B_1 \underline{U}_{v-3} + \dots + B_{v-2} \underline{U}_0}{v-1} \end{aligned} \quad (41)$$

where $v=2, 3, \dots, m$ and the $n \times n$ matrix F_w and $n \times 1$ vector \underline{X}_w were given in (22) and (23).

Proof:

Integrating (7) from 0 to t , we obtain

$$\underline{x}(t) - \underline{x}(0) = \int_0^t F(s) \underline{x}(s) ds + \int_0^t B(s) \underline{u}(s) ds \quad (42)$$

Expanding the state variables with m Taylor series basis functions

$$\begin{aligned} \underline{x}(t) &= \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_n(t) \end{bmatrix} \\ &\cong \begin{bmatrix} x_{10} \phi_0(t) + x_{11} \phi_1(t) + \dots + x_{1,m-1} \phi_{m-1}(t) \\ x_{20} \phi_0(t) + x_{21} \phi_1(t) + \dots + x_{2,m-1} \phi_{m-1}(t) \\ \vdots \\ x_{n0} \phi_0(t) + x_{n1} \phi_1(t) + \dots + x_{n,m-1} \phi_{m-1}(t) \end{bmatrix} \\ &= \underline{X}_0 \phi_0(t) + \underline{X}_1 \phi_1(t) + \dots + \underline{X}_{m-1} \phi_{m-1}(t) \end{aligned} \quad (43)$$

is obtained.

Substituting this into (42) and using the Theorem 1, we have

$$\begin{aligned} &[\underline{X}_0 \ \underline{X}_1 \ \dots \ \underline{X}_{m-1}] \phi_m(t) - [\underline{x}(0) \ 0 \ \dots \ 0] \phi_m(t) \\ &= [F_0 \underline{X}_0 \ F_0 \underline{X}_1 + F_1 \underline{X}_0 \ \dots \ F_0 \underline{X}_{m-1} + F_1 \underline{X}_{m-2} + \dots \\ &\quad + F_{m-1} \underline{X}_0] \underline{P}_{(m \times m)}, \phi_m(t) \\ &\quad + [B_0 \underline{U}_0 \ B_0 \underline{U}_1 + B_1 \underline{U}_0 \ \dots \ B_0 \underline{U}_{m-1} + B_1 \underline{U}_{m-2} + \dots \\ &\quad + B_{m-1} \underline{U}_0] \underline{P}_{(m \times m)}, \phi_m(t) \end{aligned} \quad (44)$$

Since (44) should be satisfied for all t in the time interval $t \in [0, 1)$,

$$\begin{aligned} &[\underline{X}_0 \ \underline{X}_1 \ \dots \ \underline{X}_{m-1}] - [\underline{x}(0) \ 0 \ \dots \ 0] \\ &= [F_0 \underline{X}_0 \ F_0 \underline{X}_1 + F_1 \underline{X}_0 \ \dots \ F_0 \underline{X}_{m-1} + F_1 \underline{X}_{m-2} + \dots \\ &\quad + F_{m-1} \underline{X}_0] \underline{P}_{(m \times m)} \\ &\quad + [B_0 \underline{U}_0 \ B_0 \underline{U}_1 + B_1 \underline{U}_0 \ \dots \ B_0 \underline{U}_{m-1} + B_1 \underline{U}_{m-2} + \dots \\ &\quad + B_{m-1} \underline{U}_0] \underline{P}_{(m \times m)} \end{aligned} \quad (45)$$

in obtained.

Substituting (19) into (45), we find

$$\begin{aligned} &[\underline{X}_0 \ \underline{X}_1 \ \dots \ \underline{X}_{m-1}] - [\underline{x}(0) \ 0 \ \dots \ 0] \\ &= \begin{bmatrix} 0 \ F_0 \underline{X}_0 \ \frac{F_0 \underline{X}_1 + F_1 \underline{X}_0}{2} \ \dots \\ F_0 \underline{X}_{m-2} + F_1 \underline{X}_{m-3} + \dots + F_{m-2} \underline{X}_0 \\ \vdots \\ 0 \ B_0 \underline{U}_0 \ \frac{B_0 \underline{U}_1 + B_1 \underline{U}_0}{2} \ \dots \\ B_0 \underline{U}_{m-2} + B_1 \underline{U}_{m-3} + \dots + B_{m-2} \underline{U}_0 \end{bmatrix} \quad (46) \end{aligned}$$

By equating each column in (46), we can obtain finally

$$\begin{aligned} \underline{X}_0 - \underline{x}(0) &= 0 \\ \underline{X}_1 - 0 &= F_0 \underline{X}_0 + B_0 \underline{U}_0 \\ \underline{X}_2 - 0 &= \frac{F_0 \underline{X}_1 + F_1 \underline{X}_0}{2} + \frac{B_0 \underline{U}_1 + B_1 \underline{U}_0}{2} \\ &\quad \vdots \\ \text{vth} \\ \text{column} \ \underline{X}_{v-1} - 0 &= \frac{F_0 \underline{X}_{v-2} + F_1 \underline{X}_{v-3} + \dots + F_{v-2} \underline{X}_0}{v-1} + \\ &\quad \frac{B_0 \underline{U}_{v-2} + B_1 \underline{U}_{v-3} + \dots + B_{v-2} \underline{U}_0}{v-1} \\ &\quad \vdots \\ \underline{X}_{m-1} - 0 &= \frac{F_0 \underline{X}_{m-2} + F_1 \underline{X}_{m-3} + \dots + F_{m-2} \underline{X}_0}{m-1} + \\ &\quad \frac{B_0 \underline{U}_{m-2} + B_1 \underline{U}_{m-3} + \dots + B_{m-2} \underline{U}_0}{m-1} \end{aligned} \quad (47)$$

This completes the proof.

Q.E.D

V. An Example

Let us consider the following bilinear system [11]

$$\dot{x}(t) = A(t)x(t) + B(t)u(t) + \sum_{k=1}^2 N_k(t)x(t)u_k(t), \quad x(0) = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

where

$$A(t) = \begin{bmatrix} -4 & 1 \\ -20 & 3t \end{bmatrix}, \quad B(t) = \begin{bmatrix} t & 1 \\ 1 & t \end{bmatrix},$$

$$N_1(t) = \begin{bmatrix} 2 & 3t \\ -5 & t \end{bmatrix}$$

$$N_2(t) = \begin{bmatrix} 3 & -1 \\ 20 & -5t \end{bmatrix}, \quad u(t) = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

From (6), we obtain the matrix F(t) as

$$F(t) = \begin{bmatrix} -1 & 0 \\ 0 & -2t \end{bmatrix}$$

When the number of terms used is 4, we obtain

$$f_{11}(t) = [-1 \ 0 \ 0 \ 0] \phi_0(t)$$

$$f_{12}(t) = f_{21}(t) = [0 \ 0 \ 0 \ 0] \phi_4(t)$$

$$f_{22}(t) = [0 \ -2 \ 0 \ 0] \phi_4(t)$$

Therefore

$$F_0 = \begin{bmatrix} -1 & 0 \\ 0 & 0 \end{bmatrix}, \quad F_1 = \begin{bmatrix} 0 & 0 \\ 0 & -2 \end{bmatrix},$$

$$F_2 = F_3 = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$$

Similarly, we have

$$b_{11}(t) = b_{22}(t) = [0 \ 1 \ 0 \ 0] \phi_0(t)$$

$$b_{12}(t) = b_{21}(t) = [1 \ 0 \ 0 \ 0] \phi_0(t)$$

$$u_1(t) = [0 \ 0 \ 0 \ 0] \phi_0(t)$$

$$u_2(t) = [1 \ 0 \ 0 \ 0] \phi_0(t)$$

Thus

$$B_0 = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, \quad B_1 = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, \quad B_2 = B_3 = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$$

$$U_0 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad U_1 = U_2 = U_3 = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

Equation (47) yields

$$X_0 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$X_1 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$X_2 = \begin{bmatrix} -\frac{1}{2} \\ -\frac{1}{2} \end{bmatrix}$$

$$X_3 = \begin{bmatrix} \frac{1}{6} \\ 0 \end{bmatrix}$$

Therefore an approximate solution of this bilinear system is

$$x(t) = \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} \approx \begin{bmatrix} 0 \\ 1 \end{bmatrix} \phi_0(t) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} \phi_1(t) + \begin{bmatrix} -\frac{1}{2} \\ -\frac{1}{2} \end{bmatrix} \phi_2(t) + \begin{bmatrix} \frac{1}{6} \\ 0 \end{bmatrix} \phi_3(t) = \begin{bmatrix} t - \frac{1}{2}t^2 + \frac{1}{6}t^3 \\ 1 - \frac{1}{2}t^2 \end{bmatrix}$$

It is remarked that while an approach presented in this paper is extremely simpler than that of [11], the solution obtained by (47) is equal to that of [11].

V. Concluding Remarks

In this paper, we have proposed an effective way of analyzing a class of bilinear systems via Taylor polynomials. The approach suggested in this paper gives a closed form for unknown state vector and does not need to solve any linear algebraic equation, while the method based upon the product and coefficient matrix derived by Yang and Chen is an implicit form for unknown state vector and requires to solve a linear algebraic equation with large dimension when the number of terms employed increase.

References

- [1] C. Bruni, G. Dipillo, and G. Koch, "Bilinear systems: an appealing class of "nearly linear" systems in theory and applications," *IEEE Trans. Auto. Contr.*, vol. AC-19, no. 4, pp. 334-348, August 1974.
- [2] R.R. Mohler, "Biological modeling with variable compartmental structure," *IEEE Trans. Auto. Contr.*, vol. AC-19, no. 6, pp. 922-926, December 1974.
- [3] R.R. Mohler, *Bilinear Control Processes: With Applications to Engineering, Ecology and Medicine*, Academic Press, New York, 1973.
- [4] B. Friedland, *Control System Design*, McGraw-Hill, 1986.
- [5] C.D. McGillem and G.R. Cooper, *Continuous and Discrete Signal and System Analysis*, CBS College Publishing, 1984.
- [6] S.J. Farlow, *Partial Differential Equations for Scientists and Engineers*, John Wiley & Sons, 1982.
- [7] B. Cheng and N.S. Hsu, "Analysis and parameter estimation of bilinear systems via block-pulse functions," *Int. J. Contr.*, vol. 36, no. 1, pp. 53-65, 1982.
- [8] K.R. Palanisamy and V.P. Arunachalam, "Analysis of bilinear systems via single-term walsh series," *Int. J. Contr.*, vol. 41, no. 2, pp. 541-547, 1985.
- [9] J.H. Chou and I.R. Horng, "Shifted-chebyshev series analysis and identification of time-varying bilinear systems," *Int. J. Contr.*, vol. 43, no. 1, pp. 129-137, 1986.
- [10] T.T. Lee and S.C. Tsay, "Analysis of linear time-varying systems and bilinear systems via shifted Chebyshev polynomials of the second kind," *Int. J. Systems Sci.*, vol. 17, no. 12, pp. 1757-1766, 1986.
- [11] L.F. Yang and C.K. Chen, "Analysis of bilinear systems via Taylor series," *Int. J. Systems Sci.*, vol. 18, no. 4, pp. 641-648, 1987.
- [12] P.D. Sparis and S.G. Mouroutsos, "Analysis and optimal control of time-varying linear systems via Taylor series," *Int. J. Contr.*, vol. 41, no. 3, pp. 831-842, 1985. *

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