

論 文
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목표물 위치추정을 위한 칼만 추적필터

Kalman Tracking Filter for Estimating Target Position

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요 약

정속도, 직선운동 모델의 간략형 칼만 필터(Simplified Kalman Filter-SKF)를 이용하여 운동하고 있는 목표물의 위치를 추적하며, 목표물의 기동이 탐지자(Detector)에 의해 탐지될 때 최소자승 입력평가자(Least Square Input Estimator)에 의해 추정된 기동입력치를 SKF의 추정치에 보정해 줌으로써 위치추정의 정도를 높일 수 있음을 보인다. 목표물의 기동 여부를 탐지하는 탐지자로 공산비검정(Likelihood Ratio Test)를 이용하였으며 컴퓨터 시뮬레이션을 통해 본 방법의 유효성을 밝힌다.

Abstract

By using a least-square input estimator and likelihood ratio technique, a tracking problem is presented. A Kalman tracking filter based on constant-velocity, straight-line model is used to track a target and the filtered estimate is updated using an input estimate when a maneuver is detected. Track residuals at each scan are sensed by a detector to guard against unexpected corrections of the filter. The simulation results show there are significant improvements using the scheme presented.

1. Introduction

The estimation problem of a moving object finds a wide application in such areas as air traffic control, weapons systems, space aircrafts, and range ships. In the above areas of application, a continuous tracking of an airborne object may be desired. In order to provide continuously the most reliable knowledge of

the locations of airborne targets, many different tracking filters, such as the Kalman filter, the α - β filter, the Wiener filter, and a simple extrapolator, have been considered since in the early 1960's Kalman¹⁾ first introduced the idea of the recursive filtering and there have been many advances in the development of sophisticated digital filtering algorithms for tracking airborne targets.

Earlier work on the maneuvering target tracking problem includes Singer's generalized tracking model.²⁾ The generalized model tracks a maneuvering target fairly well provided

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the so-called maneuver parameters are appropriately chosen, but if the target is not maneuvering, the tracker degrades in performance compared to the tracker based on a constant-velocity model. McAuray and Delinger⁴ have shown that there are significant improvements in the tracking capability when using a maneuvering detector to two parallel models. Thorp¹¹ showed that a weighted combination of two Kalman filter estimates in response to a detected maneuver. Moose et al.¹⁵ combined the generalized model of Singer with the adaptive Semi-Markov maneuver model. Another technique, described by Chan et al.³, use a least square estimator to estimate a target's acceleration input and updates the output of the baseline tracker, i.e., the predicted estimate, by the input estimate if the detection is declared.

By incorporating the input estimator with an update of the filtered estimate to provide some improvement in position accuracy, a maneuvering target problem is implemented. Measurements of target position are made in sensor coordinates and then filtering is performed in the same frame. We continue to use the constant-velocity, straight-line tracker for estimating positions. When a bias develops in the residual sequence due to the target deviation from the assumed motion, updating of the filtered estimate is performed to remove the bias. Whenever the estimate is updated, the error covariance increases. So, in order to guard against unexpected update of the estimate while the target doesn't maneuver, the likelihood ratio test is used to monitor the occurrence of maneuver at each time.

2. Modeling

A plot of the target and sensor geometry

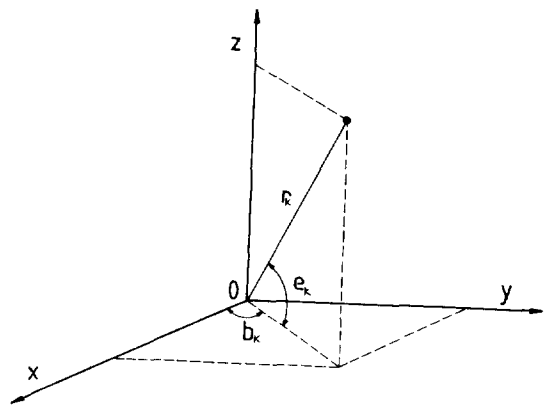


Fig. 1. Target geometry at time t_k : r_k , b_k , and e_k denote range, bearing, and elevation, respectively. The sensor is assumed at the origin.

is shown in Fig. 1.

The selection of spherical coordinates (r, b, e) rather than Cartesian coordinates (x, y, z) for our target and sensor modeling is due to the fact that the measurement error covariance becomes diagonal. The true target motion is a nonlinear coupled differential equation in the range, bearing, and elevation variables but the approximation of target motion by a linear system can be found. An approximate spherical dynamic⁵ of the target can be represented in the matrix form as

$$X_{k-1} = A_k X_k + B_k U_k + G_k W_k, \tag{1}$$

where

$$X_k = \begin{bmatrix} r_k \\ \dot{r}_k \\ b_k \\ \dot{b}_k \\ e_k \\ \dot{e}_k \end{bmatrix}, \quad A_k = \begin{bmatrix} 1 & T & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & T & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & T \\ 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix},$$

$$B_k = \begin{bmatrix} T^2/2 & 0 & 0 \\ T & 0 & 0 \\ 0 & T^2/2d_1 & 0 \\ 0 & T/d_1 & 0 \\ 0 & 0 & T^2/2d_2 \\ 0 & 0 & T/d_2 \end{bmatrix}, \quad G_k = \begin{bmatrix} 1 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 1 \end{bmatrix},$$

and

X_k = state vector at time t_k

A_k = state transition matrix

$U_k = [u_r(k) : u_b(k) : u_e(k)]^T$ deterministic input vector

$W_k = [w_r(k) : w_b(k) : w_e(k)]^T$ noise on the state

T = sampling period

$d_1 = \hat{r}_{k/k} \cos(\hat{\theta}_{k/k})$

$d_2 = \hat{r}_{k/k}$ with $\hat{r}_{k/k}$, $\hat{\theta}_{k/k}$ provided by the filtered estimate $\hat{X}_{k/k}$.

The observation equation can be written as

$$Z_{k+1} = H_{k+1}X_{k+1} + V_{k+1},$$

where

$$H_k = \begin{pmatrix} 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 \end{pmatrix} \quad (2)$$

$Z_k = [z_r(k) : z_b(k) : z_e(k)]^T$ measurement at time t_k

$V_k = [v_r(k) : v_b(k) : v_e(k)]^T$ measurement noise.

$\{W_k\}$ and $\{V_k\}$ are assumed as independent zero mean, white noises with

$$E\{W_k W_n^T\} = Q_k \delta_{kn} \quad (3)$$

$$E\{V_k V_n^T\} = R_k \delta_{kn} \quad (4)$$

$$E\{W_k V_n^T\} = 0$$

for all $k, n=0, 1, \dots$, where δ_{kn} is the Kronecker delta. It is further assumed that the initial state is independent of the two noises.

3. Kalman Filter Equations

By dropping the time indices of the constant matrices A_k , G_k , and H_k for convenience, the Kalman filter equations are given by

$$\hat{X}_{k+1/k+1} = \hat{X}_{k+1/k} + K_{k+1}[Z_{k+1} - H\hat{X}_{k+1/k}] \quad (5)$$

$$\hat{X}_{k+1/k} = A\hat{X}_{k/k} + B_k U_k \quad (6)$$

$$K_{k+1} = P_{k+1/k} H^T [H P_{k+1/k} H^T + R_{k+1}]^{-1}$$

$$P_{k+1/k} = A P_{k/k} A^T + G Q_k G^T$$

$$P_{k+1/k+1} = [I - K_{k+1} H] P_{k+1/k}. \quad (7)$$

In order to start the recursive filtering operation, the Kalman filter equations should be initialized. Assuming that after the first two measurements, i.e., Z_1 and Z_2 , are received, the optimum state vector $\hat{X}_{2/2}$ can be initialized as in (8)

$$\hat{X}_{2/2} = \begin{pmatrix} \hat{r}_{2/2} \\ \hat{\theta}_{2/2} \\ \hat{b}_{2/2} \\ \hat{b}_{2/2} \\ \hat{e}_{2/2} \\ \hat{e}_{2/2} \end{pmatrix} = \begin{pmatrix} z_r(2) \\ [z_r(2) - z_r(1)]/T \\ z_b(2) \\ [z_b(2) - z_b(1)]/T \\ z_e(2) \\ [z_e(2) - z_e(1)]/T \end{pmatrix}. \quad (8)$$

The corresponding covariance of the errors in the optimum estimate, as shown in Appendix A in detail, is

$$P_{2/2} = S_1 R_1 S_1^T + S_2 R_2 S_2^T,$$

where

$$S_1 = \begin{pmatrix} 0 & 0 & 0 \\ 1/T & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 1/T & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 1/T \end{pmatrix}, \quad S_2 = \begin{pmatrix} 1 & 0 & 0 \\ 1/T & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 1/T & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 1/T \end{pmatrix}$$

and R_1 , R_2 are obtained from (4).

Now U_k is also unknown but will be estimated in the sequel. If the target is not maneuvering, the target motion can be well modeled by simplifying the maneuvering model, i.e., $U_k = 0$ in (1). The filter that uses the simplified model is called the Simplified Kalman Filter (SKF). The filtered estimate $\hat{X}_{k+1/k+1}$ can be expressed by using $\hat{X}_{k/k}$ and Z_{k+1} :

$$\begin{aligned} \hat{X}_{k+1/k+1} &= A\hat{X}_{k/k} + B_k U_k + K_{k+1}[Z_{k+1} - H(A\hat{X}_{k/k} \\ &\quad + B_k U_k)] \\ &= D_{k+1} A\hat{X}_{k/k} + K_{k+1} Z_{k+1} + D_{k+1} B_k U_k \end{aligned} \quad (9)$$

where $D_{k+1} = [I - K_{k+1} H]$.

Using the similar idea to the innovations process,²⁾ a sequence is defined by use of the

filtered estimate

$$\tilde{Z}_k = Z_k - H\hat{X}_{k/k} \tag{10}$$

This residual sequence $\{\tilde{Z}_k\}$ has the important properties^{2,8)} that it is a zero mean white Gaussian noise if the initial state and the two noises are Gaussian. Its covariance becomes

$$E\{\tilde{Z}_k\tilde{Z}_k^T\} = HP_{k/k}H^T + R_k = \omega_k \tag{11}$$

4. Estimator of the input³⁾

As shown in the previous section, the Kalman filter equation requires an U_k , which is unknown, but should be estimated. When $U_k = 0$ in(9), we denote the estimate of the SKF by $\bar{X}_{k/k}$. Suppose that prior to time t_k no maneuvers occur such that $\hat{X}_{k/k} = \bar{X}_{k/k}$ and the target now undergoes a maneuver with a sequence of inputs $U_k, U_{k+1}, \dots, U_{k+n-1}$. The Kalman filter(9), which is linear, will continue to give, at times $t_{k+1}, t_{k+2}, \dots, t_{k+n}$, the estimates:

$$\begin{aligned} \hat{X}_{k+1/k+1} &= D_{k+1}A\hat{X}_{k/k} + K_{k+1}Z_{k+1} + D_{k+1}B_kU_k \\ &= \bar{X}_{k+1/k+1} + D_{k+1}B_kU_k \\ \hat{X}_{k+2/k+2} &= D_{k+2}A(D_{k+1}A\hat{X}_{k/k} + K_{k+1}Z_{k+1} + D_{k+1} \\ &\quad B_kU_k) + K_{k+2}Z_{k+2} + D_{k+2}B_{k+1}U_{k+1} \\ &= \bar{X}_{k+2/k+2} + D_{k+2}AD_{k+1}B_kU_k + D_{k+2} \\ &\quad B_{k+1}U_{k+1} \\ \hat{X}_{k+n/k+n} &= \bar{X}_{k+n/k+n} + \sum_{j=0}^{n-2} \left[\prod_{i=0}^j (D_{k+n-i}A) D_{k+n-1-j} \right. \\ &\quad \left. B_{k+n-2-j}U_{k+n-2-j} \right] + D_{k+n}B_{k+n-1}U_{k+n-1} \end{aligned} \tag{12}$$

Equation(12) gives us an insight that the bias developed due to target maneuvers will be removed by an addition of a correction term to the estimate of the SKF. Now we make one approximation to estimate the unknown deterministic input. The target moves under a constant acceleration, i.e., $U_{k-n} = U$ for $n=0,1,\dots,m-1$. Even though U_{k+n} are not constant over the interval, the estimator will give the best constant estimate for the different inputs in the least square sense. Equation(12) can be rewritten

as

$$\hat{X}_{k+n/k+n} = \bar{X}_{k+n/k+n} + \left\{ \sum_{j=0}^{n-2} \left[\prod_{i=0}^j (D_{k+n-i}A) D_{k+n-1-j} B_{k+n-2-j} \right] + D_{k+n}B_{k+n-1} \right\} U \tag{13}$$

Let

$$\bar{Z}_{k+n} = Z_{k+n} - H\bar{X}_{k+n/k+n} \tag{14}$$

and from(10)

$$\tilde{Z}_{k+n} = Z_{k+n} - H\hat{X}_{k+n/k+n}, \tag{15}$$

then from (13), (14), and(15)

$$\begin{aligned} \bar{Z}_{k+n} &= H(\hat{X}_{k+n/k+n} - \bar{X}_{k+n/k+n}) + \tilde{Z}_{k+n} \\ &= H\left\{ \sum_{j=0}^{n-2} \left[\prod_{i=0}^j (D_{k+n-i}A) D_{k+n-1-j} B_{k+n-2-j} \right] \right. \\ &\quad \left. + D_{k+n}B_{k+n-1} \right\} U + \tilde{Z}_{k+n}. \end{aligned} \tag{16}$$

The matrix form of(16) for $n=1, 2, \dots, m$, namely for moving data window m , is given by

$$Y = FU + e, \tag{17}$$

where

$$Y = \begin{pmatrix} \bar{Z}_{k+1} \\ \bar{Z}_{k+2} \\ \vdots \\ \bar{Z}_{k+m} \end{pmatrix}, \quad F = \begin{pmatrix} HD_{k+1}B_k \\ H(D_{k+2}AD_{k+1}B_k + D_{k+2}B_{k+1}) \\ \vdots \\ H\left\{ \sum_{j=0}^{m-2} \left[\prod_{i=0}^j (D_{k+m-i}A) D_{k+m-1-j} B_{k+m-2-j} \right] + D_{k+m}B_{k+m-1} \right\} \end{pmatrix}$$

and

$$e = \begin{pmatrix} \tilde{Z}_{k+1} \\ \tilde{Z}_{k+2} \\ \vdots \\ \tilde{Z}_{k+m} \end{pmatrix}.$$

Y and e are both $3m \times 1$ vectors and F is a $3m \times 3$ matrix. Since R_k is diagonal for all time, the $3m \times 3m$ covariance matrix for e is found to be the diagonal matrix M as

$$M = E\{ee^T\} = \begin{bmatrix} \omega_{k+1} & & & & 0 \\ & \omega_{k+2} & & & \\ & & \cdot & & \\ & & & \cdot & \\ 0 & & & & \omega_{k+m} \end{bmatrix} \quad (18)$$

The unbiased least square estimate^{(10), (14)} of U is

$$\hat{U} = (F^T M^{-1} F)^{-1} (F^T M^{-1} Y) \quad (19)$$

and its error covariance is

$$L = (F^T M^{-1} F)^{-1} \quad (20)$$

As long as detection occurs, the estimate of the tracker is updated by the optimum input estimate \hat{U} through

$$\bar{X}_{k+m/k+m} = \bar{X}_{k+m/k+m} + C_{k+m} \hat{U}, \quad (21)$$

$$\text{where } C_{k+m} = \sum_{j=0}^{m-2} \left[\prod_{i=0}^j (D_{k+m-i} A) D_{k+m-1-j} B_{k+m-2-j} \right] + D_{k+m} B_{k+m-1}$$

and $\bar{X}_{k+m/k+m}$ is the updated estimate of X_{k+m} .

The brief configuration of the input estimator is shown in Fig. 2.

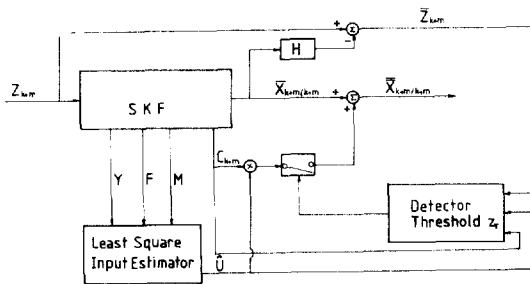


Fig. 2. Input estimator at t_{k+m} .

As we expect, the correction through the equation (21) will not only remove the bias de-

veloped in $\bar{X}_{k+m/k+m}$ but also increase the covariance of estimation errors. So, the update of the estimate should be performed when maneuvers are detected. While the target doesn't maneuver, nonzero \hat{U} is due to the effects of the two noises. Such small \hat{U} should be ignored. Using the results obtained in the Appendix B, the following covariance matrix with the positive definite $C_{k+m} L C_{k+m}^T$ term is obtained:

$$E\{ (X_{k+m} - \bar{X}_{k+m/k+m}) (X_{k+m} - \bar{X}_{k+m/k+m})^T \} = P_{k+m/k+m} + C_{k+m} L C_{k+m}^T \quad (22)$$

5. Detection of the maneuver

Detection of a target requires the choice of a threshold and a moving data window. These quantities are chosen by considering tradeoffs among the probability of false alarm P_F , the probability of detection P_D , savings of computation time, and accuracy of the least square estimate. Now we assume that all the statistics related to our detection problem are Gaussian. As shown in the previous section, if the detector correctly detects the maneuver, the estimate of the SKF is updated by the input estimate \hat{U} . Our detection scheme is that the norm of \hat{U} , $\|\hat{U}\|_\infty$, is first found and the optimum test for the corresponding component of $\|\hat{U}\|$ is made. The idea behind this detection scheme is that if $\|\hat{U}\|$ is small, the actual U is not only small but also \bar{Z}_k is small. Let the subscript x denote vectors or scalars corresponding to $\|\hat{U}\|$. Surely, x will represent one of the following: r (ange), b (earing or e (levation). Detection of the bias at time t_{k+m} based on the multiple observation reduces to the following hypothesis test:

$$H_0 : \text{no maneuver occurs} : r_k = (\bar{Z}_k)_x \quad (23)$$

$$H_1 : \text{maneuver occurs} : r_k = (\bar{Z}_k)_x - (C_k \hat{U}_k)_x \quad \text{for } k = q+1, \dots, q+m \quad (24)$$

where \hat{U}_k is the optimum estimate of U at t_k . From the above expressions, r , c , and v are def-

ined as

$$\mathbf{r} = \begin{bmatrix} r_{q+1} \\ r_{q+2} \\ \vdots \\ r_{q+m} \end{bmatrix}, \quad \mathbf{c} = \begin{bmatrix} -C_{q+1} \hat{U}_{q+1} \\ -C_{q+2} \hat{U}_{q+2} \\ \vdots \\ -C_{q+m} \hat{U}_{q+m} \end{bmatrix}_x$$

and $\mathbf{v} = \begin{bmatrix} \bar{Z}_{q+1} \\ \bar{Z}_{q+2} \\ \vdots \\ \bar{Z}_{q+m} \end{bmatrix}_x$

which are all $m \times 1$ vectors. It has always been known that the measurement error covariance \mathbf{R}_k is uncoupled. One valid assumption can be made that $\mathbf{R}_k = \mathbf{R}$ for all time. If we suppose that the target is in a well-defined track long enough before any maneuver occurs so that the steady state value of $P_{q/q}$ can be chosen to compute ω_q in (11), then we might take, to a good approximation, the $m \times m$ covariance matrix of \mathbf{v} as

$$E\{\mathbf{v}\mathbf{v}^T\} = \begin{bmatrix} \omega_q & & & 0 \\ & \ddots & & \\ & & \ddots & \\ 0 & & & \omega_q \end{bmatrix}_x = \omega. \quad (25)$$

Then the likelihood ratio and likelihood ratio test are

$$L(\mathbf{r}) = \text{EXP}[-1/2(\mathbf{r}-\mathbf{c})^T \omega^{-1}(\mathbf{r}-\mathbf{c}) + 1/2\mathbf{r}^T \omega^{-1} \mathbf{r}] \quad (26)$$

and

$$L(\mathbf{r}) \begin{matrix} > \\ < \end{matrix} \lambda, \quad (27)$$

repeatively and where λ is the threshold of the test. Taking logarithms on the both sides of (26), (27) and substituting (26) into (27), equation (27) can be written

$$\mathbf{c}^T \omega^{-1} \mathbf{r} \begin{matrix} > \\ < \end{matrix} \begin{matrix} H_1 \\ H_0 \end{matrix} \ln \lambda + 1/2 \mathbf{c}^T \omega^{-1} \mathbf{c}. \quad (28)$$

If a normalized scalar sufficient statistic is defined as

$$A(\mathbf{r}) = (\mathbf{c}^T \omega^{-1} \mathbf{r}) / (\mathbf{c}^T \omega^{-1} \mathbf{c})^{1/2}, \quad (29)$$

the likelihood ratio test thus becomes

$$A(\mathbf{r}) \begin{matrix} > \\ < \end{matrix} \begin{matrix} H_1 \\ H_0 \end{matrix} \ln \lambda (\mathbf{c}^T \omega^{-1} \mathbf{c})^{-1/2} + 1/2 (\mathbf{c}^T \omega^{-1} \mathbf{c})^{1/2} = Z_T \quad (30)$$

The detection operation therefore consists in determining $A(\mathbf{r})$ and comparing it with the threshold Z_T . If $A(\mathbf{r})$ exceeds Z_T , it is decided that H_1 is the true hypothesis; otherwise, it is decided that H_0 . We now can obtain the two expressions about P_F and P_D from the conditional densities of the sufficient statistic (29) conditioned upon the hypothesis H_0 and H_1 :

$$P_F = \int_{Z_T}^{\infty} f_A | H_1 (A | H_0) dA = \text{erf}[-Z_T] \quad (31)$$

$$P_D = 1 - \int_{-\infty}^{Z_T} f_A | H_0 (A | H_0) dA = 1 - \text{erf}[Z_T - (\mathbf{c}^T \omega^{-1} \mathbf{c})^{1/2}] \quad (32)$$

In using the Neyman-Pearson criterion^{13,14} subject to the constraint

$$P_F = \alpha, \quad (33)$$

the threshold Z_T and the probability of detection P_D can be obtained from (31), (32), and (33). However, it should be noted that the P_D depends not only upon the threshold Z_T but also \mathbf{c} , i.e., \mathbf{U}_x so that P_D will increase with increases

in U_x . Thus, this means that we need to specify the lower bound of P_D by using the minimum U_x that must be detected because the computed P_D for the given value α is exact only for one sample period. If any U_x greater than $(U_x)_{min}$ is observed, it will give a P_D higher than the lower bound. So, the steady-state value of $P_{q/q}$ and K_q are used since the estimation and detection is needed most when the gain of the SKF is small.

It has been mentioned earlier that the number m is a design parameter of the detector and estimator. Essentially, the m most recent residuals are examined to determine whether they differ significantly from the statistical description of their values that assumes no maneuvers. The number m greater than one will not only increase the accuracy of the input estimate but also prevent failure declarations due to a single unacceptable measurement. On the other hand, it is inappropriate to use a large number m since this will decrease the sensitivity to maneuver occurrence as time progresses, along with an increase in the computation time. Hence, we might choose $m=5$ as the reasonable number of data points in the detection of a maneuver.^{3,4,5,10}

6. Simulation and Results

The tracking scheme presented was implemented on a VAX/VMS computer using simulated data. For purposes of comparison, the estimate of the SKF without the maneuver detector was implemented in addition to that of the Kalman filter described in section 4, called the Modified Kalman Filter(MKF). Target trajectories generated in sensor coordinates are shown in Fig.3 through Fig.5. The following statistics are used:

$$R = \text{diag}(\sigma_r^2, \sigma_b^2, \sigma_e^2)$$

where $\sigma_r = 0.0183\text{km}$ and $\sigma_b = \sigma_e = 0.003\text{rad}$

$$Q = \text{diag}(\sigma_r^2, \sigma_b^2, \sigma_e^2)$$

where $\sigma_r = 0.183\text{m}$ and $\sigma_b = \sigma_e = 0.03\text{mrad}$

$T = 2\text{ sec}$

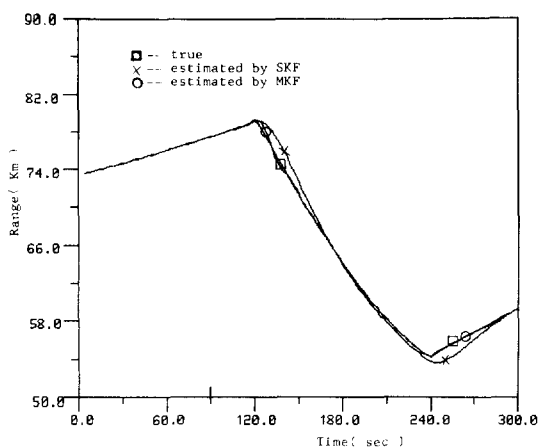


Fig.3. Tracker performance in range coordinate.

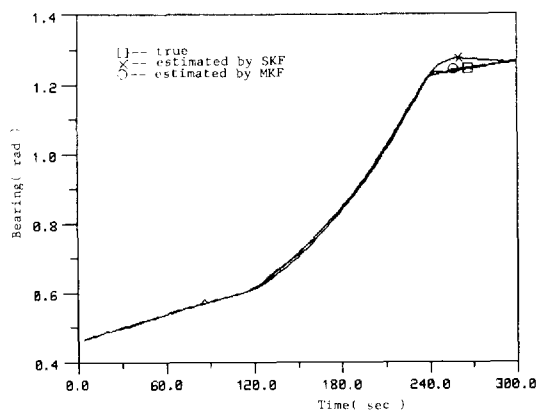


Fig.4. Tracker performance in bearing coordinate.

and the simulated scenario is as follows:

The target initially flying at 0.1km/sec in speed is on a constant course for 112 seconds. At time $t=112\text{ sec}$ it begins to maneuver with acceleration input $+0.02\text{km/sec}^2$, at $t=120\text{sec}$ commences a fast turn, and at $t=126\text{ sec}$ completes its maneuvers. The target keeps a straight-line track at 0.38km/sec in speed until it again starts its maneuvers with -0.02km/sec^2 at $t=232\text{ sec}$, making another fast turn at $t=240\text{ sec}$, and finally completing its maneuvers at $t=246\text{ sec}$.

For our experiment, the probability of false alarm, P_F , was given by $P_F = 2 \times 10^{-3}$ which led to $Z_T = 2.87$. The filter was near steady state at $t=64\text{ sec}$ after the SKF was first put into

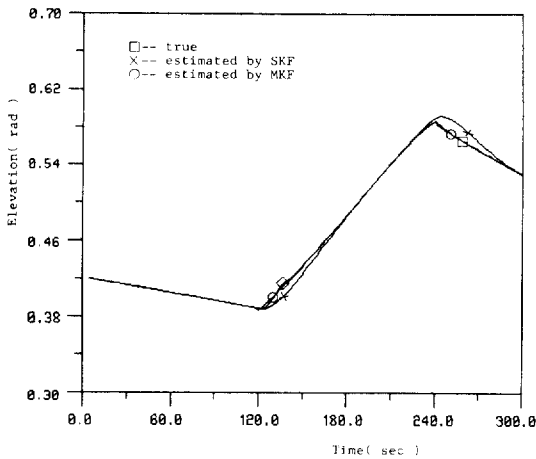


Fig.5. Tracker performance in elevation coordinate

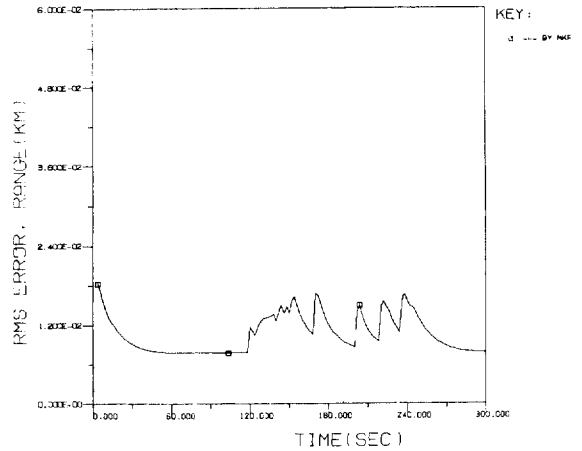


Fig.6. rms error in range coordinate.

operation.

In Fig.3 through Fig.5, we have shown the position estimates of the MKF and the SKF. The results shown in Fig.3 through Fig.5 show the MKF's superior tracking performance.

In order to give a good comparison, we computed the sum square residual errors; which are the sum square of the differences between the true and estimated values. Table 1 gives the sum square residual errors. It is clear that the residual errors are quite small, especially for the MKF and a small difference as shown indicates the satisfactory performance of the MKF. The next simulation, described by Fig.6 through Fig.8, shows the rms error in the MKF position estimate. It becomes apparent that during detecting maneuvers the rms error increases but decreases during constant-speed, straight-line flight because the filter is settled by the detector. This simulation also shows that the rms estimation errors are kept below

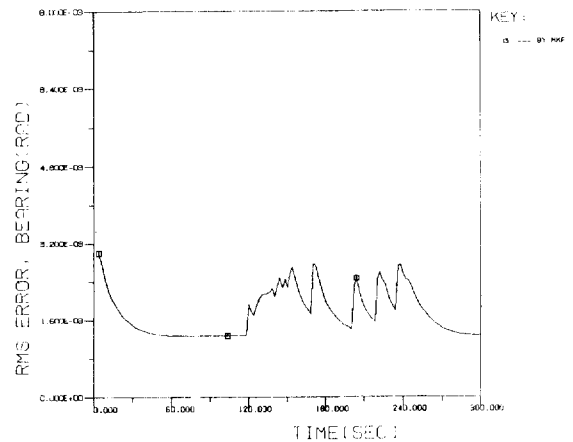


Fig.7. rms error in bearing coordinate.

Table 1. Sum square residual errors between the true and estimated values.

	SKF	MKF
Range	0.726914×10^2	0.262592×10^1
Bearing	0.204657×10^{-1}	0.964886×10^{-3}
Elevation	0.415042×10^{-2}	0.127342×10^{-3}

the inherent measurement errors. This means that the MKF is able to maintain track with good tracking accuracy.

7. Conclusions

In this paper we have presented a tracking scheme which has given a good estimate of a target position in three dimensional space. This tracking scheme gives two advantages in addition to the several advantages which the scheme suggested by Chan et al. does:

First, By incorporating the least square input es-

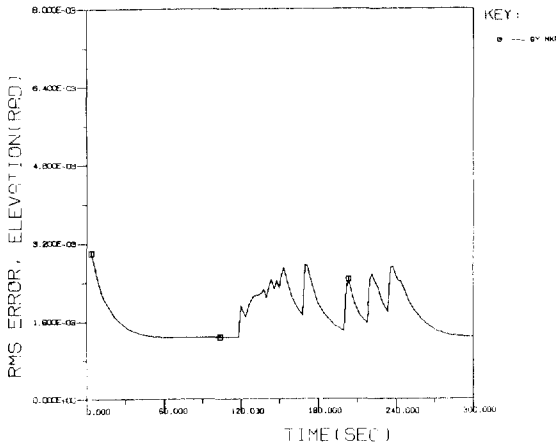


Fig.8. rms error in elevation coordinate.

timator and a detector, our baseline tracker produces a filtered estimate instead of a predicted estimate to ensure increased accuracy and the filtered estimate is updated when maneuvers are detected. Second, a detector based on multiple observation residuals are used to detect maneuvers to avoid maneuver declarations due to a single unacceptable observation.

Simulations show that the tracking scheme presented here can give a realistic solution to tracking problems for maneuvering or non-maneuvering targets. In particular, if a typical target trajectories are known in practical applications, we can recommend reasonable choices of the threshold to improve the overall performance.

Appendix A

Initialization of the optimum state vector and the corresponding covariance of a tracking filter may be often taken while the maneuvers of the target are not well known. Therefore, we start the initialization under assumption of constant-velocity, straight-line flight and with approximate equations(A-1) which provide satisfactory results for the majority of applications

$$\begin{aligned} \dot{r}_2 &\approx (r_2 - r_1)/T \\ \dot{b}_2 &\approx (b_2 - b_1)/T \end{aligned} \quad (\text{A-1})$$

$$\dot{e}_2 \approx (e_2 - e_1)/T.$$

Equation(8) can be rewritten as

$$\begin{aligned} \hat{r}_{2/2} &= \hat{r}_2 + v_r(2) \\ \hat{r}_{2/2} &= \hat{r}_2 + [v_r(2) - v_r(1)]/T \\ \hat{b}_{2/2} &= b_2 + v_b \\ \hat{b}_{2/2} &= \hat{b}_2 + [v_b(2) - v_b(1)]/T \\ \hat{e}_{2/2} &= e_2 + v_e(2) \\ \dot{e}_{2/2} &= \dot{e}_2 + [v_e(2) - v_e(1)]/T. \end{aligned} \quad (\text{A-2})$$

Expression in the matrix form of (A-2) becomes

$$X_2 - \hat{X}_{2/2} = S_1 V_1 - S_2 V_2,$$

$$S_1 = \begin{bmatrix} 0 & 0 & 0 \\ 1/T & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 1/T & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 1/T \end{bmatrix} \quad \text{and} \quad S_2 = \begin{bmatrix} 1 & 0 & 0 \\ 1/T & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 1/T & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 1/T \end{bmatrix}.$$

Hence,

$$\begin{aligned} P_{2/2} &= E\{(X_2 - \hat{X}_{2/2})(X_2 - \hat{X}_{2/2})^T\} \\ &= S_1 R_1 S_1^T + S_2 R_2 S_2^T. \end{aligned}$$

Appendix B

The least square estimate \hat{U} is unbiased because e is also a zero mean, white sequence. Taking the expectation in both (21) and (13) with the time index change, we get

$$\begin{aligned} E\{\bar{X}_{k+m/k+m}\} &= E\{\bar{X}_{k+m/k+m}\} + C_{k+m} E\{\hat{U}\} \\ &= E\{\bar{X}_{k+m/k+m}\} + C_{k+m} E\{U\} \\ &= E\{\hat{X}_{k+m/k+m}\}. \end{aligned}$$

Next we derive the error covariance in (22). From (13) and (21),

$$\bar{X}_{k+m/k+m} = \hat{X}_{k+m/k+m} + C_{k+m} (\hat{U} - U).$$

From (17), (19), and (20), we see that

$$\hat{U} - U = LF^T M^{-1} e.$$

Then,

$$X_{k+m} - \bar{X}_{k+m/k+m} = (X_{k+m} - \hat{X}_{k+m/k+m}) - C_{k+m}$$

$LF^TM^{-1}e$.

Hence,

$$\begin{aligned} E\{ (X_{k+m} - \bar{X}_{k+m/k+m}) (X_{k+m} - \bar{X}_{k+m/k+m})^T \} \\ = P_{k+m} + C_{k+m} L F^T M^{-1} E \{ e e^T \} (C_{k+m} L F^T M^{-1})^T \\ = P_{k+m} + C_{k+m} L C_{k+m}^T \end{aligned}$$

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