

중속형 SOGI 기반 DFIG 풍력터빈의 개선된 회전자 속도 추정  
 누옌안탄, 이동춘  
 영남대학교 전기공학과

Improved Rotor Speed Estimation in DFIG Wind Turbine Systems Based on Cascaded SOGI

Anh Tan Nguyen and Dong-Choon Lee

Department of Electrical Engineering, Yeungnam University

**Abstract-** In this paper, an improved rotor speed estimation in DFIG wind turbine systems based on a cascaded SOGI is proposed. Due to excellent harmonics and DC offset rejection capability of the cascaded SOGI, the accurate rotor speed estimation can be achieved despite the harmonics and sensing offset in DFIG currents. The simulation results have verified the validity of proposed method.

1. INTRODUCTION

The reliability of the doubly-fed induction generator (DFIG)-based wind turbine systems can be enhanced by the uses of sensorless control methods. Based on the rotor current found by the stator-side quantities, i.e., stator voltage and stator current, and the measured rotor current, the rotor position can be estimated [1]-[3]. At the normal conditions, the good estimations can be achieved by the uses of aforementioned methods. Their estimation performances, however, can be adversely affected by the harmonics in the stator and rotor currents under distorted grid conditions. In addition, the current sensing offset is a concern for those methods as well.

Due to excellent harmonic rejection capability, the second-order generalized integrator (SOGI) has attracted a great deal of attention. Based on SOGI, the extraction of the positive- and negative-sequence components of grid voltage for grid synchronization control has been proposed [4]. Recently, the application of SOGI to sensorless control in DFIG wind turbine systems has been suggested [5].

This paper proposes an improved speed estimation in DFIG wind turbine systems using the cascaded SOGI. Since the cascaded SOGI behaves like an adaptive filter, the proposed speed observer can allow the accurate speed estimation in presence of harmonics and sensing offset in DFIG currents. The effectiveness of the proposed method is verified by the simulation results.

2. CONTROL OF DFIG WIND TURBINE SYSTEMS

In the DFIG wind turbine systems, the grid-side converter is responsible for maintaining the DC-link voltage and grid-side power factor at unity. At the same time, the maximum power point tracking (MPPT) algorithm can be carried out by the rotor-side converter [5].

3. PROPOSED SENSORLESS CONTROL METHOD

3.1. Cascaded SOGI Structure

In the cascaded SOGI structure shown in Fig. 1,  $x$  is the input signal,  $x''$  and  $qx''$  are the output signals, which are in quadrature with each other,  $k_1$  and  $k_2$  are damping factors and  $\omega'$  is the tuning frequency.

The transfer functions of the cascaded SOGI are expressed as:

$$G_1(s) = \frac{x''(s)}{x(s)} = \frac{s^2 \omega'^2 k_1 k_2}{s^4 + s^3 \omega'(k_1 + k_2) + s^2 \omega'^2 (2 + k_1 k_2) + s \omega'^3 (k_1 + k_2) + \omega'^4} \quad (1)$$

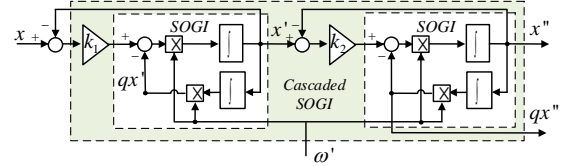


Fig. 1. Cascaded SOGI structure.

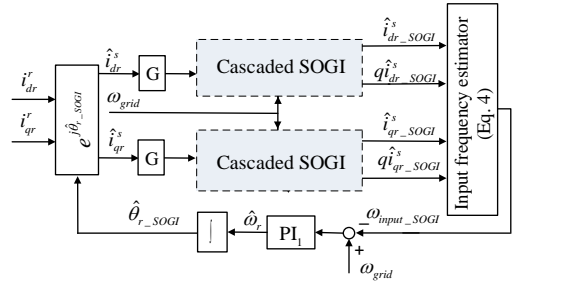


Fig. 2. Cascaded SOGI-based speed observer.

$$G_2(s) = \frac{qx''(s)}{x''(s)} = \frac{s \omega'^3 k_1 k_2}{s^4 + s^3 \omega'(k_1 + k_2) + s^2 \omega'^2 (2 + k_1 k_2) + s \omega'^3 (k_1 + k_2) + \omega'^4} \quad (2)$$

With an additional single SOGI used as pre-filter in the cascaded SOGI structure, the adverse effects of harmonics and DC offset in  $x$  on  $x''$  and  $qx''$  can be significantly mitigated.

3.2. Cascaded SOGI-based Speed Observer

Converting (1) and (2) into a time domain, the amplitude ratio of two cascaded SOGI outputs is obtain as:

$$\frac{|qx''|}{|x''|} = \frac{\omega'}{\omega} \quad (3)$$

where  $\omega$  is the input frequency of cascaded SOGI.

As observed in (3), if the values of  $|qx''|$ ,  $|x''|$  and  $\omega'$  are known, the estimation of  $\omega$  can be achieved. Based on this idea, a cascaded SOGI-based speed observer is proposed as shown in Fig. 2, in which, the frequency  $\omega_{input\_SOGI}$  and the gain  $G$  are expressed as:

$$\omega_{input\_SOGI} = \omega_{grid} \sqrt{\frac{\hat{i}_{dr\_SOGI}^{s2} + \hat{i}_{qr\_SOGI}^{s2}}{(q\hat{i}_{dr\_SOGI}^s)^2 + (q\hat{i}_{qr\_SOGI}^s)^2}} \quad (4)$$

$$G = \sqrt{\frac{\hat{i}_{dr}^{s2} + \hat{i}_{qr}^{s2}}{\hat{i}_{dr\_SOGI}^{s2} + \hat{i}_{qr\_SOGI}^{s2}}} \quad (5)$$

where the  $d$ - $q$  axis rotor current  $\hat{i}_{dqr}^s$  is found by:

$$\hat{i}_{dqr}^s = \frac{\lambda_{dqs}^s - L_s i_{dqs}^s}{L_m} \quad (6)$$

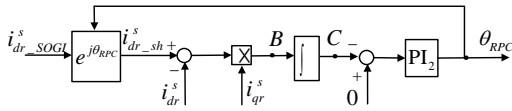


Fig. 3. Rotor position compensator.

TABLE I. Parameters of DFIG

Parameters	Values
Stator frequency	60 Hz
Rated stator power	2 MW
Rated torque	$10.6 \times 10^3$ Nm
Rated line-line stator voltage	690 V
Rated stator current	1670 A

In order to remove the effects of sensing offset in stator currents on estimation performances, the measured stator currents are filtered out through the cascaded SOGI-based filters with the tuning frequency at grid frequency.

### 3.3. Rotor Position Compensator

In the case of the unspecified initial rotor position as well as speed estimation error, it is necessary to add the compensation component  $\theta_{RPC}$  to  $\hat{\theta}_{r\_SOGL}$  in Fig. 4 to correct the estimated rotor position. The component  $\theta_{RPC}$  is output of the rotor position compensator (RPC) as shown in Fig. 3.

In proposed method, the estimated slip angle can be expressed as:

$$\hat{\theta}_{sl} = \theta_{grid} - (\hat{\theta}_{r\_SOGL} + \theta_{RPC}). \quad (7)$$

## 4. SIMULATION RESULTS

Simulations of the proposed method for a 2 MW DFIG wind turbine system are performed in the PSIM platform. The system specifications are listed in Table I.

Figs. 4 and 5 show the system performances with applications of MRAS observer-based method [2] and proposed method under distorted grid conditions and DFIG current disturbance, respectively. For the grid distortion, the 5% fifth- and seventh-order harmonic components are included in each phase of grid voltage. The wind speed profile is shown in Fig. 4(a). Fig. 4(b) shows the speed estimation error  $\Delta\omega_r$ . The estimated slip angle  $\hat{\theta}_{sl}$  is illustrated in Fig. 4(c). Figs. 4(d) and (e) show the stator active power  $P_s$  and reactive power  $Q_s$ , respectively.

As observed from Fig. 4 and 5, the performances of two methods considered are similar under normal operating conditions. Under distorted grid condition and DFIG current disturbance, however, the proposed method gives the superior performances to the existing method. The maximum of  $\Delta\omega_r$  is about 60 rpm for the existing method, whereas it is about 20 rpm for proposed method. When the current disturbance of 115 A is applied at  $t = 4.5$  s, with the existing method, the fluctuations of active power  $\Delta P_s$  and reactive power  $\Delta Q_s$  are about 380 kW and 400 kVAR, as shown in Fig. 4(d) and (e), respectively. On the other hand, with the proposed method,  $\Delta P_s$  and  $\Delta Q_s$  are about 150 kW and 100 kVAR, as shown in Fig. 5(d) and (e), respectively.

## 5. CONCLUSIONS

In this paper, an improved speed estimation in DFIG wind turbine systems has been proposed, based on cascaded SOGI. Due to excellent harmonics and DC offset rejection capability of cascaded SOGI, the good rotor speed estimation can be achieved even under grid voltage distortion and DFIG current disturbances. Under those conditions, the simulation results have shown that the maximum of

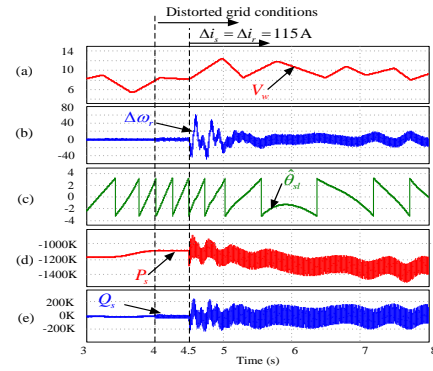


Fig. 4. System performances with MRAS observer-based method [2] under distorted grid conditions and current disturbances. (a) Wind speed (m/s). (b) Speed estimation error (rpm). (c) Estimated slip angle (rad). (d) Stator active power (W). (e) Stator reactive power (VAR).

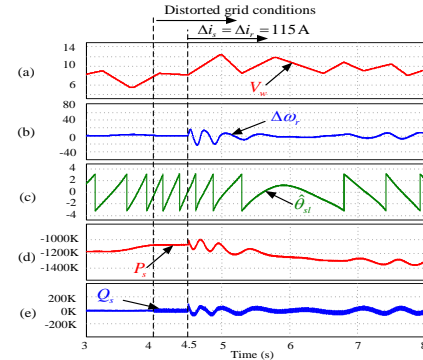


Fig. 5. System performances with proposed method under distorted grid conditions and current disturbances. (a) Wind speed (m/s). (b) Speed estimation error (rpm). (c) Estimated slip angle (rad). (d) Stator active power (W). (e) Stator reactive power (VAR).

$\Delta\omega_r$  for the proposed method is about 20 rpm (about 1.11% of measured speed).

## ACKNOWLEDGMENTS

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