자기력 부상 시스템에서 외란 제거를 위한 자속 궤환 방식에 관한 연구

A study on the Flux Feedback Approach for the Rejection of Dynamic Disturbance Forces in a Magnetically Suspended System

이준호†(한국철도기술연구원), 신경호**(한국철도기술연구원), 이강미**(한국철도기술연구원)
김백현**(한국철도기술연구원), 김종기**(한국철도기술연구원), 김용규**(한국철도기술연구원)

Jun-Ho Lee(Korea Railroad Research Institute),Kyeong-Ho Shin(Korea Railroad Research Institute),
Kang-Mi Lee(Korea Railroad Research Institute),Bak-Hyun Kim(Korea Railroad Research Institute),
Jong-Ki Kim(Korea Railroad Research Institute),Yong-Kyu Kim(Korea Railroad Research Institute)

ABSTRACT

This study is concerned with static and sinusoidal disturbance rejection for a single periodic input disturbance with known period. In the area of active elimination of a disturbance force, the control input should have two different kinds of gains: one is to deliver a stable control and the other is a force component to cancel the external disturbance force. In this paper we employ a simple state feedback control law to make the balance beam stable and employ a linear observer to estimate the states which represent the external disturbance force components. Simulation results verify our proposed control method to reject a static and sinusoidal disturbance force.

1. Introduction

In magnetic bearing systems, one of the interesting research areas is to control the electromagnet force[1],[2] to reject a static external disturbance force or a sinusoidal disturbance force which is produced by rotor mass imbalance or by a certain external disturbance force [3]. Static disturbance force can be rejected by using a simple integrator even if the control law does not have a component for the cancellation of the static disturbance force [4]. However if this static disturbance force is combined with a periodic force which has a certain frequency the simple integrator cannot reject the periodic disturbance term. In order to reject the periodic disturbance force, if the additional analog circuit which can reject the periodic force is not employed, the control law should have a certain component to cancel

out the periodic disturbance force [5],[6].

In this paper we present a method to reject a static and a periodic disturbance force which has a certain frequency. A magnetically suspended balance beam is used as a test rig. Static and sinusoidal disturbance forces are produced by a small fan which is placed on the one side of the balance beam. A state feedback control law is then employed to make the balance beam stable [7].

Since we do not employ any additional analog circuit, the state feedback control law should have a certain component to cancel out the disturbance term [8]. Thus the effective rejection property of the disturbance force by using pure digital control method can be achieved by using a very accurate and high performance estimation or tracking tool for the disturbance force.

In this paper we present a flux feedback approach to produce control current. The control current has two components. One is the gap deviation signal which is sensed by the gap sensor, the other is the flux deviation feedback signal which is expressed by a linear combination of the total force of the magnetic

† Korea Railroad Research Institute
  E-mail : jhlee77@krri.re.kr
  Tel : (031) 460-5040, Fax : (031) 460-5449

** Korea Railroad Research Institute

bearings. At the equilibrium point the total force of the magnetic bearings should be equal to the sum of the state feedback control effort and the estimated disturbance force. The flux deviation feedback signal can be expressed by the sum of the state feedback control effort and the estimated disturbance force. For the tracking of the sensor output we employ a linear observer.

This control method can be applied to a magnetically suspended system such as flywheel energy storage system which has rotational input disturbance and artificial heart pump system supported by electromagnets which has the static and periodic disturbance force.

First, we show the geometrical structure of the balance beam and mathematical model which includes static and sinusoidal disturbance components, and then we present how to design the linear observer. Finally we verify the proposed external disturbance cancellation method by simulation results.

2. Balance Beam Mathematical Model

Fig. 1 shows the geometry of the symmetric balance beam with two horseshoe shaped magnetic bearings. Table 1 shows each parameter of the balance beam system. A small fan is placed on the one side of the balance beam to produce a static and periodic disturbance force.

Table 1. Balance beam parameters

Parameter	Symbol	Value	Units
Angular Position	θ		rad
Half Bearing Span	L_a	1.1412	m
Mass Moment of Inertia about the Pivot Point	J	0.0948	kg m^2
Coil Current in Bearing 1	i_1'		A
Coil Current in Bearing 2	i_2'		A
Coil Resistance	R	0.7	Ω
Coil Inductance	L	0.728	mH
Magnetic Bearing Open Loop Stiffness	K_a	2114	N/m
Actuator Current Gain	K_i	1.074	N/A
Steady Current	i_0	1	A
Steady Gap	g_0	380	μm

Equation of motion of the balance beam is expressed by the second order dynamic equation as:

$$J\ddot{\Theta} + C_z\dot{\Theta} = L_a(f_1 - f_2) + f_d \tag{1}$$

Where C_z is a damping factor of pivot, f_d is the disturbance force and

$$f_1 = \mu_0 A_g N^2 \frac{(i_b + i_1')^2}{(2g_0 + L_a\theta)^2}$$

$$f_2 = \mu_0 A_g N^2 \frac{(i_b + i_2')^2}{(2g_0 - L_a\theta)^2}$$
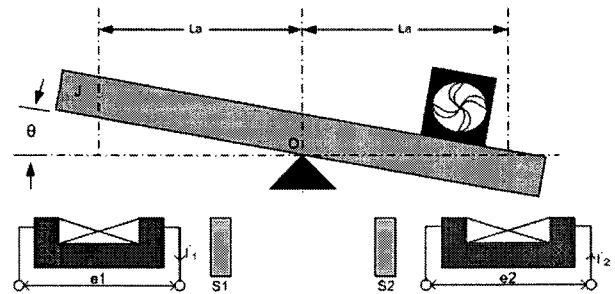
where i_b is the bias current.

Fig 1. Balance beam which has a small fan producing sinusoidal disturbance force

By Linearization at $i_1' = 0$, $i_2' = 0$ and $\theta = 0$ equation (1) becomes

$$\dot{x} = Ax + Bu + D\omega \tag{2}$$

where

$$x = \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix}, \quad A = \begin{bmatrix} 0 & 1 \\ \frac{2k_x L_a^2}{J} & -\frac{C_z}{J} \end{bmatrix}$$

$$B = \begin{bmatrix} 0 \\ \frac{k_i L_a}{J} \end{bmatrix}, \quad D = \begin{bmatrix} 0 \\ \frac{1}{J} \end{bmatrix}$$

$$u = i_1' - i_2'$$

As we see in Eq. (2) we employed a current control method for the rejection of the disturbance force acting on the balance beam. In (2) if we define that ω is a static and periodic disturbance terms ω can be expressed such as:

$$\omega = k_1 + k_2 \sin(\beta t + \phi) + k_3(\beta t + \phi) \tag{3}$$

Where k_1, k_2, and k_3 are the amplitude of the disturbance force and β, ϕ represent a frequency and a phase of the periodic disturbance forces. Thus (3) becomes

$$\omega = C_\omega x_\omega \tag{4}$$

Where

$$x_\omega = \begin{bmatrix} 1 \\ \sin(\beta t + \phi) \\ \cos(\beta t + \phi) \end{bmatrix}$$

$$C_\omega = \begin{bmatrix} k_1 & k_2 & k_3 \end{bmatrix}$$

Let states of the disturbance forces x_w be

$$\dot{x}_\omega = A_\omega x_\omega \tag{5}$$

The solution of (5) is

$$x_\omega = C e^{A_\omega t} \tag{6}$$

The differentiation of (6) yields

$$\dot{x}_\omega = A_\omega C e^{A_\omega t} = \begin{bmatrix} 0 \\ \beta \cos(\beta t + \phi) \\ \beta \sin(\beta t + \phi) \end{bmatrix} \tag{7}$$

From (7) we get A_ω matrix as:

$$A_\omega = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & \beta \\ 0 & -\beta & 0 \end{bmatrix}$$

The augmented state equation is derived by using (2), (4) and (5) with the result

$$\begin{bmatrix} \dot{x} \\ \dot{x}_\omega \end{bmatrix} = \begin{bmatrix} A & D C_\omega \\ 0 & A_\omega \end{bmatrix} \begin{bmatrix} x \\ x_\omega \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u \tag{8}$$

$$y = \begin{bmatrix} I & 0 \end{bmatrix} \begin{bmatrix} x \\ x_\omega \end{bmatrix} \tag{9}$$

3. Design of Disturbance Estimator

How to design a linear observer is already a well known procedure. In this section we simply show the design procedure of a linear observer for output tracking. From (8) and (9) we get the state space equations as:

$$\dot{x} = A_a x + B_a u \tag{10}$$

$$y = C_a x \tag{11}$$

The objective is to make $\lim_{t \to 0} \hat{y} - y = 0$ by using a certain observer feedback gain L. Here $\hat{y}$ represents estimated output. Thus the state space equation of the estimated system by observer (see Fig. 2) is

$$\dot{\hat{x}} = A_a \hat{x} + B_a u - L(\hat{y} - y) \tag{12}$$

$$= A_a \hat{x} + B_a u - L C_a (\hat{x} - x) \tag{13}$$

where $C_a = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \end{bmatrix}$. In this equation the estimator is expressed by the linear state space equations.

4. Synthesis of Feedback Controller

In the previous section we showed the normal design procedure of a linear observer. This linear observer has a limited estimation range due to the balance beam nonlinearity. The purpose of this section is to show an approach using flux to estimate the plant states in the presence of plant nonlinearity. As we see in the fundamental equation flux which is produced by core magnet is proportional to the pole face area as:

$$\Phi = B A_g$$

$$= \frac{\mu_0 N i}{2g} A_g = k_\phi \frac{i}{g} \tag{14}$$

where $k_\phi = \frac{\mu_0 N A_g}{2}$ and $i = i_b + i'$. From (14) we get the current equation as a function of gap displacement and air gap flux.

$$i = \frac{1}{k_\phi} g \Phi \tag{15}$$

In the presence of external disturbance force the role of the electromagnet actuators is to reject the external disturbance force. The following formula should meets the requirement:

$$F_n = F_1 - F_2 = -\hat{F}_d - kx \tag{16}$$

where $\hat{F}_d$ is the estimated disturbance force, F_1 and F_2 are the forces which are produced by the first and the second electromagnets, k is the state feedback gain. Eq. (15) and (16) allow us to drive the control current as a function of the exerted force F_n and air gap flux Φ such as:

$$F_n = k_f \left(\frac{i_1^2}{g_1^2} - \frac{i_2^2}{g_2^2} \right)$$

$$= \frac{k_f}{k_\phi^2} \left(\Phi_1^2 - \Phi_2^2 \right)$$

$$= \frac{k_f}{k_\phi^2} \left[(\phi_b + \phi')^2 - (\phi_b - \phi')^2 \right]$$

$$= \frac{k_f}{k_\phi^2} 4 \phi_b \phi' \tag{17}$$

where $k_f = \frac{\mu_0 N^2 A_g}{4}$, ϕ_b is the bias flux.

Eq. (17) yields

$$\phi' = \frac{F_n k_\phi{}^2}{4 k_f \phi_b}$$ (18)

The air gap flux in each actuator is

$$\Phi_1 = \phi_b + \phi'$$

$$= \phi_b + \frac{F_n k_\phi{}^2}{4 k_f \phi_b}$$ (19)

$$\Phi_2 = \phi_b - \phi'$$

$$= \phi_b - \frac{F_n k_\phi{}^2}{4 k_f \phi_b}$$ (20)

Finally we get the control current from (15), (19) and (20) as:

$$i_1 = \frac{1}{k_\phi} g_1 \Phi_1$$

$$= \frac{1}{k_\phi} g_1 \left(\phi_b + \frac{F_n k_\phi{}^2}{4 k_f \phi_b} \right)$$ (21)

$$i_2 = \frac{1}{k_\phi} g_2 \Phi_2$$

$$= \frac{1}{k_\phi} g_2 \left(\phi_b - \frac{F_n k_\phi{}^2}{4 k_f \phi_b} \right)$$ (22)

In (21) and (22) the exerted force is implemented by the following relation.

$$F_n = -\hat{F}_d - kx$$ (23)

In the above procedure, we showed a synthesis method of the control current in the presence of the external static and sinusoidal disturbance force. The main frame in this procedure is in the equation (14) which represents a linear combination between Φ and $\frac{\dot{i}}{g}$. Based on this linear combination the control current is also expressed by the linear multiplication of Φ and g, and then the final control current is achieved by the modified formula which has the deviated flux feedback components as shown in Eq. (21) and (22). As we have mentioned, Eq. (21) and (22) have the exerted force components that involve estimated disturbance force by the observer and state feedback. Thus, once the observer estimates the external disturbance force $\hat{F}_d$ by substituting (21) and (22) into (2) we can easily check the cancellation of the external disturbance force term. Fig. 2 shows the block diagram for the rejection of the external
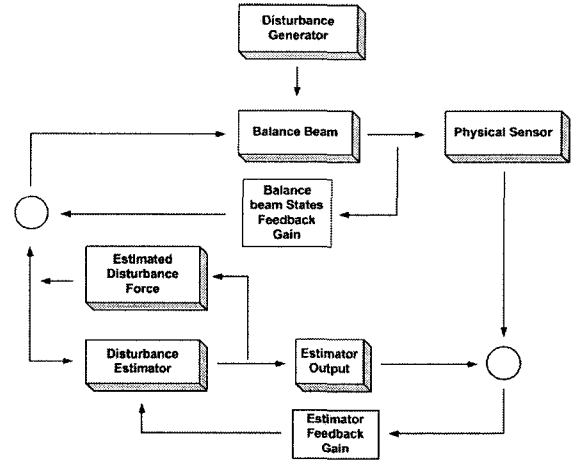
disturbance force using observer.

Fig. 2. Block diagram for disturbance estimation

5. Simulation Results

In this section we show the simulation results of the static and periodic disturbance force rejection. A MATLAB Simulink model was designed based upon the nonlinear electromagnet force equation, which represents a nonlinear simulation. For the simulations we used the following feedback gains and parameters of the static and sinusoidal disturbance.

● State feedback gains:

$$K_p = 5500, \quad K_d = 500$$

● Estimated feedback gains:

$$10^6 \times [0.0008 \quad 0.2006 \quad 0.0983 \quad 2.7952 \quad -0.3089]^T$$

● Disturbance coefficient matrix:

$$C_\omega = [5 \quad 1 \quad 1]$$

● Frequency of the sinusoidal disturbance:

$$\beta = 60 \text{ [rad/sec]}$$

● Phase of the sinusoidal disturbance:

$$\phi = 10 \text{ [rad]}$$

Fig. 3 and 4 show the balance beam gap deviations and the static and sinusoidal disturbance force which has 10[Hz] frequency and 10[rad] phase value produced by small fan. In order to realize the proposed control method we need to know the exact value of the phase in the small fan. Only for the simulation we set 10[rad] for the phase of the periodic function. In Fig. 3 we see good levitation status even if there is the external static and periodic

disturbance force which is acting on the balance beam, shown in Fig. 4. The peak in Fig 3. occurs due to the static disturbance as we see in Fig. 4. and also due to the estimation error of the disturbance force in the transient state of the estimator.
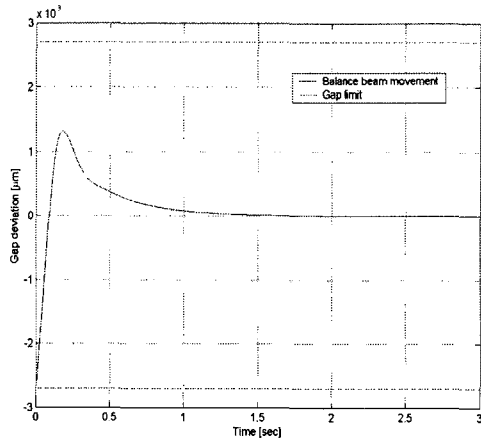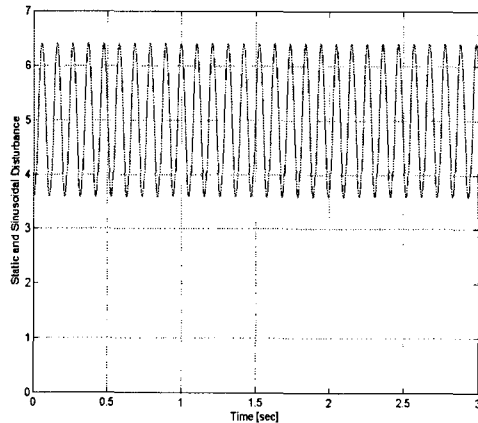

Fig. 3 Gap Deviations

Fig. 4 Static and sinusoidal disturbance

Fig. 5 represents the exerted input force to produce the control current expressed by Eq. (21) and (22). If we compare Fig. 5 with Fig. 4 we see the same amplitude in the static and sinusoidal disturbance force except for the sign convention. This means that the exerted force components included in the control current rejects the disturbance force. Fig. 6 shows the simulation results for the disturbance force estimated by the designed observer. In this figure we see good tracking properties of the estimator after the very short transient state. The transient state has a certain
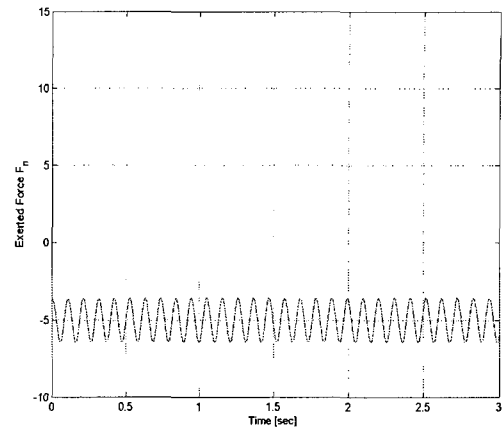

Fig. 5. Exerted input force for current

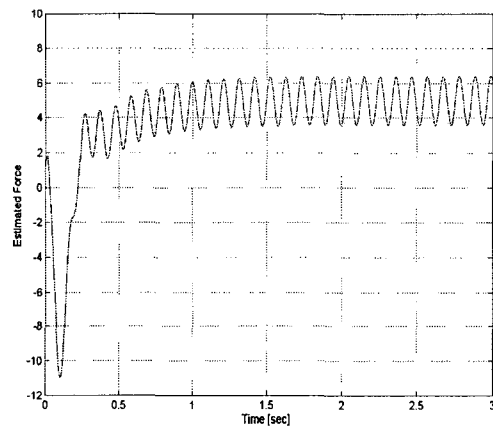unexpected response due to the balance beam

Fig. 6. Estimated disturbance force

nonlinearity and the linear property of the estimator. From the implementation point of view the estimator feedback gains should be selected very carefully to avoid the unexpected transient response mentioned above. It should be mentioned that we do not see the initial transient states in Fig. 5 even if there is the peak transient value in the estimated disturbance force (see Fig. 6). This is because of the wide range scaling in time axis. Fig. 7 is the rescaled plot in time axis and shows the initial transient states. Fig. 8 shows the electromagnets force signal which cancel out the static and sinusoidal disturbance force shown in Fig. 4. Fig. 9 shows the control current to produce electromagnets force. In Fig. 9 we see the initial transient response, but in Fig. 8 initial transient

response is not shown due to the time scale. Initial transient peak of the electromagnet force acting on the balance beam is shown in the rescaled Fig. 7.
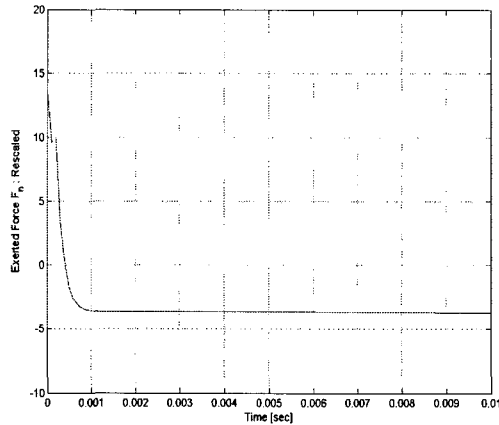

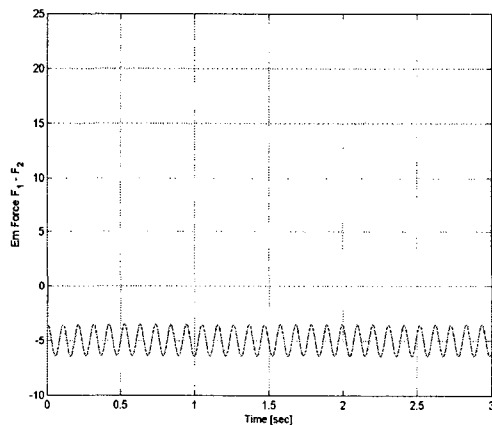Fig. 7 Exerted input force for current (Rescaled)

Fig. 8 Electromagnet actuator force : $F_1 - F_2$

6. CONCLUSIONS

In this paper we proposed a control method using an exerted force and flux feedback to reject a static and periodic disturbance force. A control current formula including exerted force component and deviated flux component was achieved by combining the state feedback control law and the output of the estimator. First we showed the balance beam geometrical scheme and the fundamental equation of motion of the balance beam, and then we showed the synthesis procedure of the control current to cancel out the
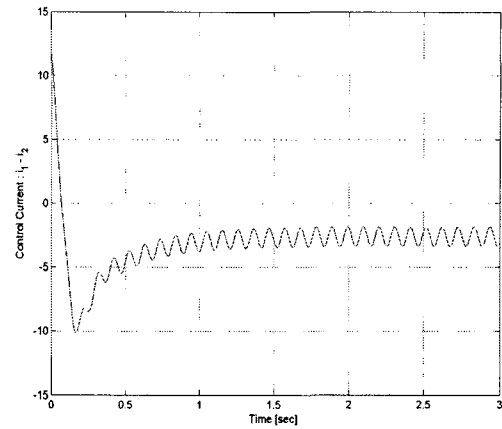

Fig. 9 Control current : $i_1 - i_2$

external disturbance force. Finally the proposed control approach was validated by the simulation results produced by a nonlinear simulation model.

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