# Soft-Switching Boost Chopper Type DC-DC Power Converter with a Single Auxiliary Passive Resonant Snubber.

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Abstract - This paper presents boost and buck and buck-boost DC-DC converter circuit topologies of high-frequency soft switching transition PWM chopper type DC-DC high power converters with a single auxiliary passive resonant snubber. In the proposed boost power converter circuits operating under a principle of ZCS turn-on and ZVS turn-off commutation schemes, the capacitor and inductor in the auxiliary passive resonant circuit works as the lossless resonant snubber. In addition to this, the switching voltage and current peak stresses as well as EMI and RFI noises can be basically reduced by this single passive resonant snubber. Moreover, it is proved that converter circuit topologies with a passive resonant snubber are capable of solving some problems of the conventional hard switching PWM processing based on high-ferquency pulse modulation operation principle. The simulation results of this converter are discussed as compared with the experimental ones. The effectiveness of this power converter with a single passive resonant snubber is verified by the 5 kW experimental breadboad set up.

Keywords; DC-DC Converters, Time ratio controlled chopper, A single passive resonant snubber, Soft switching PWM, Extended converter topologies

# I. Introduction

In recent years, semiconductor switched-mode power conversion circuits and systems have been begun to be effectively applied for new energy applications such as solar photovoltaic generations, fuel cell generation and secondary battery energy storage and super capacitor energy storage, with the great advances in power semiconductor switching devices and their peripheral technologies. The active high power devices and power modules such as MOSFETs, IGBTs, SITs and MCTs are generally introduced for high performance power conversion circuits with PWM control scheme. However, in high-frequency PWM applications,

their dynamic and static performances are not more suitable and acceptable for high power converter operating under a principle hard-switching PWM. Because their transient switching operation in turn-on and off modes causes high di/dt and high dv/dt electrical stresses, the power semiconductor devices have high peak voltage stress or peak current stress due to parasitic parameters related on switching surges. On the other hand, soft-switching PWM techniques have been introduced for chopper and inverter in order to solve these significant problems, which are based upon the switching commutation operating schemes to turn on and off under zero current or zero voltage principle. To improve these semiconductors power devices related switching problems, a variety of circuit topologies of soft-switching PWM chopper type and inverter type DC-DC converters have been proposed on the basis of passive lossless snubber and active resonant snubber. In this paper, in the first place, a new single-switch auxiliary passive resonant snubber-assisted boost PWM chopper type DC-DC converter, together with its extended buck type and buck-boost type converter topologies, which can operate in a single lossless soft-switching PWM scheme auxiliary resonant active power switches is proposed for high power applications. In the second place, the operation principle and performance evaluations on the basis of computer simulation results are presented along with its concerned converter topologies. A 5 kw experimental breadboad setup circuit using a single-switch auxiliary passive resonant snubber assisted boost chopper DC-DC converter is built and tested and its experimental performances are illustrated here in.

#### II. CIRCUIT DESCRIPTION

A practical passive soft switching circuit topology of boost PWM chopper type DC-DC converter with a single auxiliary passive resonant snubber is shown in Fig.1. This

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boost chopper type DC-DC converter circuit using a single IGBT, which can achieve soft switching transition commutation modes, is composed of DC load side energy feed back type resonant lossless snubber with passive power switches and passive resonant components. In this circuit, an inductor L<sub>1</sub> in series with a main active power switch S, two resonant capacitors C1 and C2. Because the pulse modulated active power switch S with a single passive resonant lossless snubber can completely achieve soft-switching high-frequency turn-on and off points in spite of duty cycle control, which is based on zero-current turn-on due to resonant in L<sub>1</sub> and zero-voltage turn-off due to resonant capacitor C1. In addition, this boost type DC-DC power converter circuit operation is based on soft switched PWM strategy by a single active power switch. This generic soft switching snubber approach is possible to use all types of conventional PWM chopper type DC-DC converters.

Since two capacitors of the resonant snubber circuit act as lossless snubbers, power losses and EMI noise in the soft commutation of non isolation link PWM boost converter circuit can be actually reduced including peak voltage and peak current stress related ratings.

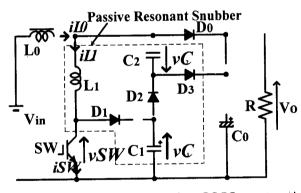


Fig.1. A proposed single-switch PWM boost DC-DC converter with passive resonant snubber.

# III. PRINCIPLE OF OPERATION

The boost soft switching PWM chopper type DC-DC power converter with a single passive resonant snubber which includes  $L_0$ , SW,  $D_0$  and  $C_0$ . Auxiliary passive resonant snubber circuit includes  $C_1$ ,  $C_2$ ,  $L_1$ ,  $D_1$ ,  $D_2$  and  $D_3$ . Because the capacitor  $C_0$  for the output voltage smoothing as well as the inductor  $L_0$  for input current smoothing are too large constants as compared with resonant LC constants and soft commutation time is extremely short, it is assumed that the current flowing through  $L_0$  and the output voltage Vo are

considered as a constant value during the soft commutation transition processing process. In order to describe the working principle, the operating voltage and current waveforms in steady state are depicted in Fig.2 and Fig.3, shows the equivalent circuits of this converter. The operation of this converter is described below for its equivalent circuits show in Fig.3.

**Mode 1:** At mode  $1(t_0 < t < t_1)$ , when the active power switch SW in the boost PWM DC-DC power converter shown in Fig.1 is in off-state, stored energy during the previous conduction interval in the boost DC inductor  $L_0$  is transferred to the load through the blocking diode  $D_0$ . Then, the capacitor  $C_1$  is charged with the polarity depicted in Fig.1 up to the output charged to the output DC voltage Vo, but the capacitor  $C_2$  is not charged. When the active power switch SW is turned on at time  $t_0$ , the current of the blocking diode  $D_0$  is linearly decreased by the inductor  $L_1$ . On the other hand, the current flowing through the inductor  $L_1$  increases linearly from zero.

**Mode 2:** At mode  $2(t_1 < t < t_2)$ , when the current of the diode  $D_0$  becomes zero,  $D_2$  begins to conduct. The resonant inductor  $L_1$  and the capacitors  $C_1$  and  $C_2$ , start to resonate, and the energy stored in  $C_1$  is discharged through  $C_1$ - $C_2$ - $L_1$ -SW loop as shown in Fig.3, producing a sinusoidal resonant current. As a result, the current through the inductor  $L_1$  contains both the current of the inductor  $L_0$  and the resonant current.

**Mode 3:** At mode  $3(t_2 < t < t_3)$ , when the voltage across the capacitor  $C_1$  is equal to zero, the diode  $D_1$  begins to conducts so  $L_1$  and  $C_2$  resonate through  $D_1$  and  $D_2$ . The voltage across the capacitor  $C_2$  still increase in accordance with the decrease of the inductor  $L_1$  current. When the resonant current is completed, the energy stored into the capacitor  $C_1$  is transferred to the capacitor  $C_2$ .

**Mode 4:** At mode  $4(t_3 < t < t_4)$ , when current through the inductor  $L_1$  is equal to the energy storage inductor  $L_0$ , the turn-on commutation process is completed and the voltages across the  $C_1$  and  $C_2$  are kept constant. The converter circuit works as a conventional boost chopper type converter circuit.

**Mode 5:** At mode  $5(t_4 < t < t_5)$ , as soon as the active power switch SW is turned off at time  $t_4$ , the current through the main power switch is completely commutated to  $C_1$  through  $D_1$ . As the voltage across the capacitor  $C_1$  increases to the output voltage Vo due to this charging current.

**Mode 6:** At mode  $6(t_5 < t < t_6)$ , when the diode  $D_3$  is turned on, at time  $t_5$ , the energy stored into  $C_2$  is discharged through the diode  $D_3$ .

Mode 7: At mode  $7(t_6 < t < t_7)$ , when the voltage across  $C_1$  reaches the output voltage Vo of this power converter, the diode  $D_2$  is turned on at time  $t_6$ . In addition, the current flowing through the inductor  $L_1$  is transferred to the load through  $D_2$  and  $D_3$ . Thus, the current through  $L_1$  decreases continuously.

**Mode 8:** At mode  $8(t_7 < t < t_8)$ , when the current flowing through the inductor  $L_1$  becomes zero, the current through the inductor  $L_1$  is kept constant to zero, until the active power switch SW is turned on.

**Mode 9:** At mode  $9(t_8 < t)$ , when the voltage across the capacitor  $C_2$  goes to zero at time  $t_8$ , another operation mode starts. The current flowing through in the capacitor  $C_2$  is commutated to the diode  $D_0$ .

The turn-on di/dt stress and turn-off dv/dt stress are both limited by the circuit constant  $L_1$  and  $C_1$ , respectively. The switching power loss of the power device:IGBT and EMI noise are greatly reduced due to a soft commutation PWM strategy. At the turn-on commutation switching process, the energy stored into  $C_1$  is transferred to  $C_2$ . At the turn-off switching process, the energy stored into  $C_2$  and  $L_1$  are transferred to the load. This kind of a single passive resonant lossless snubber assisted-converter circuit has a high efficient operation under a stable soft commutation strategy,

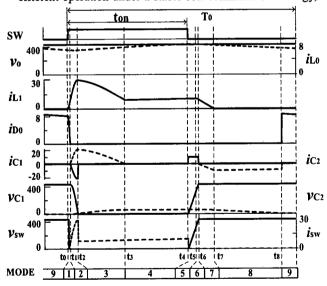


Fig.2 Voltage and current operating waveforms

and the actual efficiency converter results in a high value in spite of PWM-based voltage regulation.

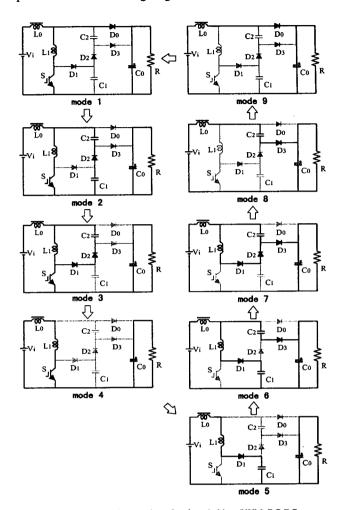


Fig.3 The operating modes of soft switching PWM DC-DC converter with a single passive resonant snubber and equivalent circuits.

## IV. EXPERIMENTAL RESULTS AND DISCUSSIONS

The soft switching boost PWM chopper type DC-DC power converter with auxiliary passive resonant snubber using IGBT is designed and simulated. The design circuit parameters and specifications of this converter shown in Table 1. When the duty cycle D as a control variable is 0.5, voltage and current switching waveforms of this power converter are respectively illustrated in Fig.4 and Fig.5. The soft-switching operation of the boost PWM completely converter with a single passive resonant snubber is able to be

achieved at both ZCS turn on and ZVS turn off. Fig.6 shows comparative output power vs. efficiency characteristics for hard switching and soft switching.

Table 1 Design specifications and Experimental Parameters.

Circuit parameters	$L_1 = 3 \mu H$
	$C_1 = 10 \text{ nF}$
	$C_2 = 500 \text{ nF}$
Input voltage	V <sub>in</sub> = 200 V
Switching frequency	f = 20  kHz
Duty cycle	D = 0.5
Load resistor	$R = 33\Omega$
Boost inductor	$Lo = 500 \mu H$
DC filter capacitor	$Co = 2000 \mu F$

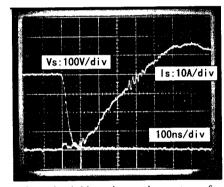


Fig.4 Experimental switching voltage and current waveforms of the active power switch S at turned on (under 5 kW output).

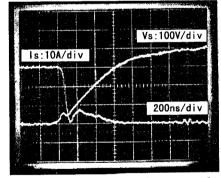


Fig.5 Experimental switching voltage and current waveforms of the active power switch S at turned off (under 5 kW output).

# V. SIMULATION RESULTS

The relation between the duty cycle vs. output voltage characteristics show in Fig.7. In addition, in Fig.7, peak voltage across switch SW is represented because peak voltage is equal to output voltage. Under control variable duty cycle, dv/dt and di/dt dynamic stress characteristics are shown in Fig.8 and Fig.9.

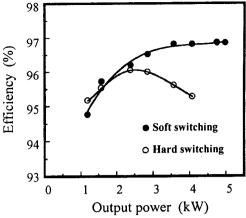


Fig. 6 Output power vs. Efficiency characteristics.

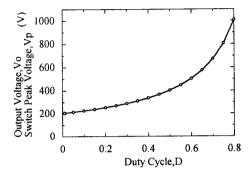


Fig.7 Output voltage Vo and Switch peak voltage Vp, Duty cycle characteristics

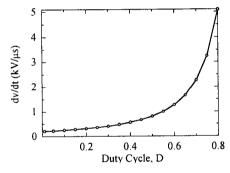


Fig.8. dv/dt characteristics at turn off.

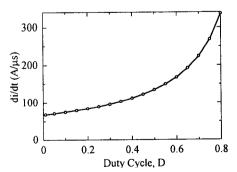
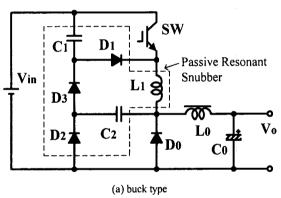


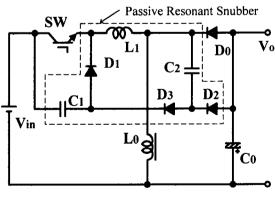
Fig.9. di/dt characteristics at turn on.

### VI. EXTENDED CONVERTER TOPOLOGY

The passive resonant lossless snubber-assisted boost chopper type DC-DC converter circuit using a single active power switch have two different soft switching operation modes; turn-on and turn-off commutation schemes. Both of them have a current loop to transfer the energy stored in the inductor or capacitor in the resonant snubber. In the turn-on point, an inductor in series with power switching device limits effectively the turn-on related di/dt stress and a capacitor in parallel with the inductor to transfer the energy. In the turn-off instant, a capacitor in parallel with switching device restrains the turn-off related dv/dt stress and a capacitor in series with the capacitor to transfer the energy.

Using the general principle mentioned above, the energy recovery-based passive resonant lossless snubber circuit assisted buck and boost-buck DC-DC power converter topologies are respectively shown in Fig.10, and the basic principle of operation is similar to the principle of passive resonant snubber-assisted the boost power converter discussed above.





(b) boost-buck type
Fig.10 Extended converter family.

#### VII. CONCLUSIONS

In this paper, 10kW soft switching auxiliary resonant snubber assisted PWM chopper type boost DC-DC power converter using a single IGBT was proposed for new energy interfaced power conditioner. Its steady-state operation principle was described on the basis of the equivalent circuits of this DC-DC converter and steady-state performance evaluations of the proposed power converter circuit illustrated and discussed on the basis of the simulation and practical experiment as a function of duty cycle variable. The soft switching PWM power converter topology with a single passive resonant snubber can be expected to reduce the switching power losses, switching surge related EMI noise. The operating characteristics of this soft switching PWM chopper type in spite of duty cycle control scheme power converters are practically verified on the basis experimental and simulation results. The proposed power converters have the following salient unique features.

- (i). This boost converter with lossless resonant snubber can operate at a soft switching commutation without auxiliary active switches.
- (ii). The active power switch operation can completely achieve zero-current turn-on and zero-volt turn-off schemes.
- (iii). The capacitors for the passive resonant circuit act as lossless snubbers.
- (iv). This power converter can operate under constant switching frequency.

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