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A study on narrowband interference rejection in PN spread spectrum communication systems

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Abstract

In this paper, a improved whitening filter for narrowband interference rejection in PN spread spectrum communication systems is proposed. This whitening filter has the feedback from the output to the input of adaptive transversal filter and is superior to conventional LMS whitening filter in respect to SNR and power spectrum since the calculation of tap weight is based on maximum likelihood estimation.

1. Introduction

The performance of a direct sequence spread spectrum receiver may be enhanced considerably by using linear mean square (LMS) estimation techniques for suppression of narrowband interference. This technique is conceptionally accomplished by implementing digital whitening filter hm that is equivalent to whitening the received signal (transmitted signal plus noise plus jammer). Since interference as well as signal is whitened the performance using LMS estimation is somecases inferier to that obtained using other electronic rejection circuitry. So new suppression filter suggested by the form of feedback in order to whiten interference alone.

2. System model and signal format

The system considered is a BPSK spread spectrum system shown in Fig. 1 and it is

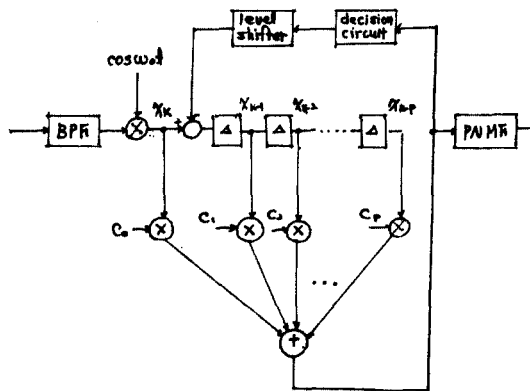


Fig. 1 system model

assumed that the narrowband interference is multiple CW interference and narrowband gaussian interference and SS modulation is accomplished by direct sequence (PN) method. In a narrowband interference environment the received wideband signal can be expressed as

$$x_k = S_k + \mathcal{I}_k + n_k$$

$$= U P_k + \sum_{m=1}^M C_m \cos(2\pi f_m k \Delta t + \Phi_m) + n_k \quad (1)$$

where

- $\mathcal{I}_k$  narrowband interference
- $n_k$  zero mean white Gaussian noise with a variance  $\sigma^2$
- $U$  information symbol
- $P_k$  modulated PN sequence
- $C_m$  amplitude of narrowband jamming
- $f_m$  sampling frequency of narrowband jam-

ing  
 $\Phi_n$  phase of the narrowband jamming distributed uniformly over the range  $(-\pi, \pi)$   
 $\Delta t$  sampling interval

### 3. Coefficient Calculation

If there is no error in decision circuit of Fig.1, the equations to determine the coefficients is expressed as

$$\begin{aligned} \hat{E}_k &= \sum_{l=1}^p C_l E_{k-l} = C_p C_r \\ &= \sum_{l=1}^p C_l (\chi_{k-l} - d_{k-l}) \\ &= \sum_{l=1}^p C_l (M_{k-l} + \delta_{k-l}) \end{aligned} \quad (2)$$

where

$\{d_k\}$  independent random binary data sequences

$$C_p = [C_1, \dots, C_p]^T = [C_{p-1}, C_p] \quad (3)$$

$$\chi_p = [\chi_1, \dots, \chi_{p-1}] = [\chi_{p-1}, \chi_p] \quad (4)$$

We choose that minimizes the sum of squares of the deviation  $\hat{E}_k - E_k$ , that is the mean squares of the

$$J = E\{(\hat{E}_k - E_k)^2\}$$

The resulting estimate obtained by setting  $\frac{\partial J}{\partial C} = 0$  is

$$C_p^* = R_p^{-1} U_p \quad (5)$$

where

$$R_p = [r_{ij}]_{i,j=1,\dots,p} = \begin{bmatrix} R_{p-1} & J r_{p-1} \\ r_{p-1}^T & r_0 \end{bmatrix} \quad (6)$$

$$= E(\chi_p \chi_p^T) \quad (7)$$

$$U_p = [u_1, u_2, \dots, u_p]^T = [E(\chi_1^2), \dots, E(\chi_p^2)]$$

The coefficients  $\{C_p^*\}$  can be recursively computed by using a form of the Levinson algorithm.

### 4. Performance

#### A. Power Spectra

Power Spectra for whitening filter with feedback is shown in Fig. 2. In this case, the performance of a direct sequence (DS)

spread spectrum receiver is enhanced by using this filters assuming the number of tones  $M$  is chosen to 100 and the order of the whitening filter is 4.

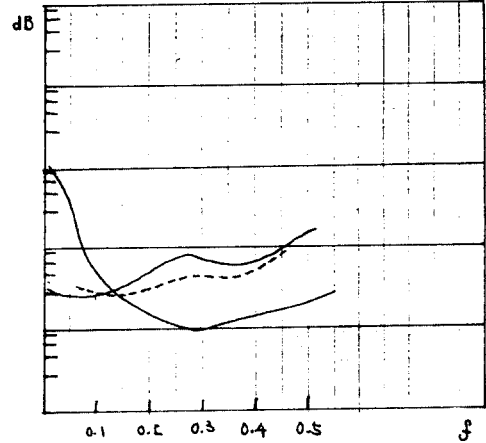


Fig.2 Power spectrum; SNR+8dB,  $C_m=0.6$ , the number of tones is chosen to be 50.

#### B. Signal-to- Noise ratio

In SS system of Fig. 1, the correlator input is

$$E_k = \sum_{n=0}^M C_n (\chi_n - d_n) \quad (8)$$

assuming  $d_{k-n} = d_{k-n}$  in ideal case, then

$$\begin{aligned} S &= \sum_{k=1}^L E_k P_k + P_k \\ &= \sum_{k=1}^L \sum_{n=0}^M C_n P_k (P_{k-n} + \delta_{k-n} - P_{k-n}) + P_k \end{aligned} \quad (9)$$

The mean of  $S$  is

$$E(S) = L \quad (10)$$

and

$$\begin{aligned} E(S^2) &= E\left[\sum_{k=1}^L \sum_{m=0}^M \sum_{j=1}^L \sum_{i=0}^M C_n C_j P_k P_i \delta_{k-n} \delta_{i-j}\right] \\ &= \sum_{k=1}^L \sum_{n=0}^M \sum_{j=1}^L \sum_{i=0}^M C_n C_j E(P_k P_i) E(\delta_{k-n} \delta_{i-j}) \\ &= L \sigma^2 \sum_{n=1}^M C_n^2 \end{aligned} \quad (11)$$

since

$$E(P_k P_i) = \delta_{k-i} = \begin{cases} 1, & k=i \\ 0, & \text{otherwise} \end{cases} \quad (12)$$

$$E[\delta_{k-n} \delta_{i-j}] = \begin{cases} 1, & n=j=0 \\ \delta_{n-j}, & n>0, j>0, i=k \\ 0, & \text{otherwise} \end{cases} \quad (13)$$

$$E(\xi_{j-n}) = 0 \quad (14)$$

$$\begin{aligned} E(\xi_{j-n} \xi_{i-j}) &= \frac{1}{L} \sum_{n=1}^L Q_n^2 \cos(2\pi n \Delta f M (j-n) + \sigma^2 \delta_{j-n}) \\ &= M P_N \Delta f \Delta t + \sigma^2 \delta_{j-n} \end{aligned} \quad (15)$$

The variance of  $S$  is

$$\begin{aligned} V(S) &= E(S^2) - L^2 \\ &= \sum_{i=1}^L \sum_{j=1}^L \sum_{k=1}^L \sum_{n=1}^L Q_n C_j E(P_k P_i) E(\xi_k \xi_j) \\ &\quad + E(P_k P_i) - L^2 \\ &= L P_N \Delta f \Delta t M + L \sigma^2 \sum_{n=1}^L Q_n^2 \end{aligned} \quad (16)$$

Therefore

$$SNR = \frac{E^2(S)}{V(S)} = \frac{L}{\sigma^2 \sum_{n=1}^L Q_n^2 + L P_N \Delta f \Delta t M} \quad (17)$$

The SNR of conventional SS spread spectrum system is

$$SNR' = \frac{L}{\frac{1}{2} L b_w^2 + \sigma^2} \quad (18)$$

in the system without feedback

$$SNR'' = \frac{L}{(1 + \sigma^2) \sum_{n=1}^L Q_n^2 + M P_N \Delta f \Delta t} \quad (19)$$

since

$$E_k = \sum_{n=0}^N Q_n X_{k-n} \quad (20)$$

Hence the performance is improved and plotted in Fig. 3.

#### 4. Conclusion

This paper has proposed a feedback whitening filters that has the feedback from the output to the input of adaptive transversal filter and showed the SNR representation of this system. This filter can effectively suppress narrowband jamming signals and has satisfactory performance regard to SNR and Power spectrum. Although the error probability has not derived in this paper, it is also expected to be improved because of whitening only jamming component.

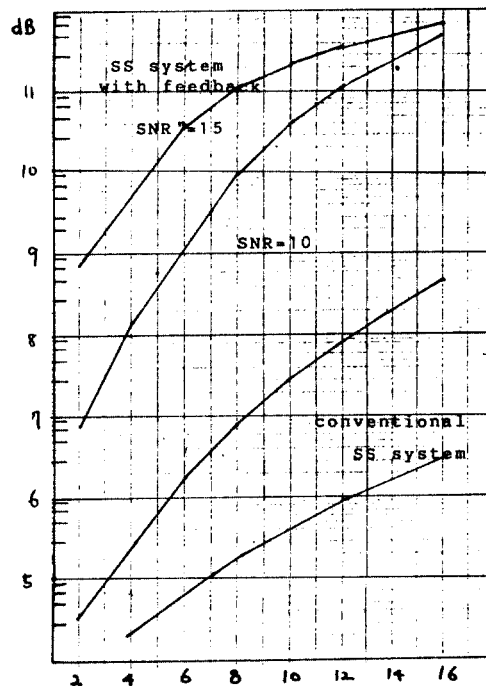


Fig.3 SNR comparison. (SNR versus The number of taps )

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